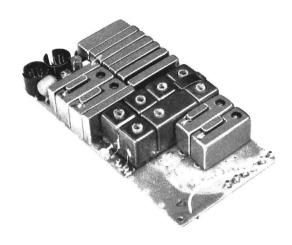
MASTR Personal Series

PROGRESS LINE

PE MODELS

66 - 88 MHz, RECEIVER MODELS 4ER87B10, 11, 12 & 13



SPECIFICATIONS *

Type Number ER-87-B

Audio Output (EIA) 500 milliwatts at less than 5% distortion

Channel Spacing 20 kHz

Sensitivity

12 dB SINAD (EIA Method) 0.35 μV

20 dB Quieting Method $0.40~\mu V$

Selectivity

EIA Two-Signal -60 dB

Spurious Response -70 dB

Image -80 dB

Intermodulation (EIA) -70 dB

Audio Response Within +2 and -10 dB of a standard 6 dB per octave de-emphasis curve from 300

to 3000 Hz (1000 Hz reference)

Modulation Acceptance ±6.5 kHz

Squelch Sensitivity

Critical Squelch 0.20 µV

Maximum Squelch Greater than 20 dB Quieting

Maximum Frequency Spread (MHz)

The specific of the specific o

^{*}These specifications are intended primarily for the use of the serviceman. Refer to the appropriate Specification Sheet for the complete specifications.

TABLE OF CONTENTS

SPECIFICATIONS	Cover
DESCRIPTION	1
CIRCUIT ANALYSIS	1
Oscillator Modules A314, A315 & A318 thru A323 Compensator A313 Front End A336/A337 Crystal Filters A318 & A319/A320 & A321 IF Amplifiers A304 & A306 Limiter A327/A328 & Discriminator A308/A322 Audio Amplifier A309/A310 Audio PA A311 Squelch A312	1 2 3 3 5 5 5 5
MAINTENANCE	
ALIGNMENT PROCEDURE	9
TEST PROCEDURES	10
OUTLINE DIAGRAM	12
SCHEMATIC DIAGRAM	
FRONT END	13
RECEIVER BOARD	13
PARTS LIST & PRODUCTION CHANGES	14
TROUBLESHOOTING PROCEDURE	15
ILLUSTRATIONS	
Figure 1 - Receiver Block Diagram Figure 2 - Typical Oscillator Circuit Figure 3 - Typical Compensator Circuit Figure 4 - Typical IF Amplifier Circuit Figure 5 - Typical Limiter Circuit Figure 6 - Typical Audio Amplifier Circuit Figure 7 - Typical Audio PA Circuit Figure 8 - Typical Squelch Circuit	1 2 3 4 4 5 6 6

- WARNING -

No one should be permitted to handle any portion of the equipment that is supplied with high voltage; or to connect any external apparatus to the units while the units are supplied with power. KEEP AWAY FROM LIVE CIRCUITS!

ADDENDUM #1 TO LBI4664C, LBI4639E, LBI4642D, LBI4689B and LBI30188A

TABLE OF CONTENTS -	
DESCRIPTION	1
OUTLINE DIAGRAM	2
SCHEMATIC DIAGRAM	3
PARTS LIST AND PRODUCTION CHANGES	4

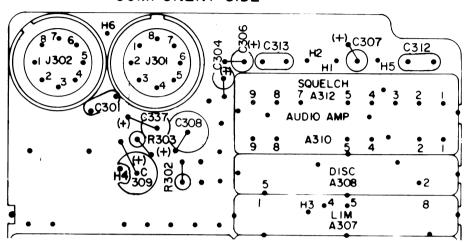
This addendum describes the optional Mobile Detector boards used in Vehicular Repeater applications. Detector boards 19D417493G4,5 are similar to PE receiver boards except that PA module A311, C302, C305, C345 and R301 are not used. Also R306 (3R151P103J) is added.

The Mobile Detector monitors the mobile radio transmit frequencies to determine if a second vehicular repeater is repeating portable-to-base station transmissions. If a portable-to-base transmission is in progress, the mobile detector prevents the Vehicular Reapeter from becoming the priority unit.

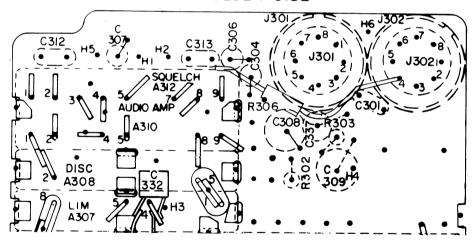
GENERAL ELECTRIC COMPANY+ MOBILE COMMUNICATIONS DIVISION WORLD HEADQUARTERS+LYNCHBURG, VIRGINIA 24502 U.S.A.



COMPONENT SIDE

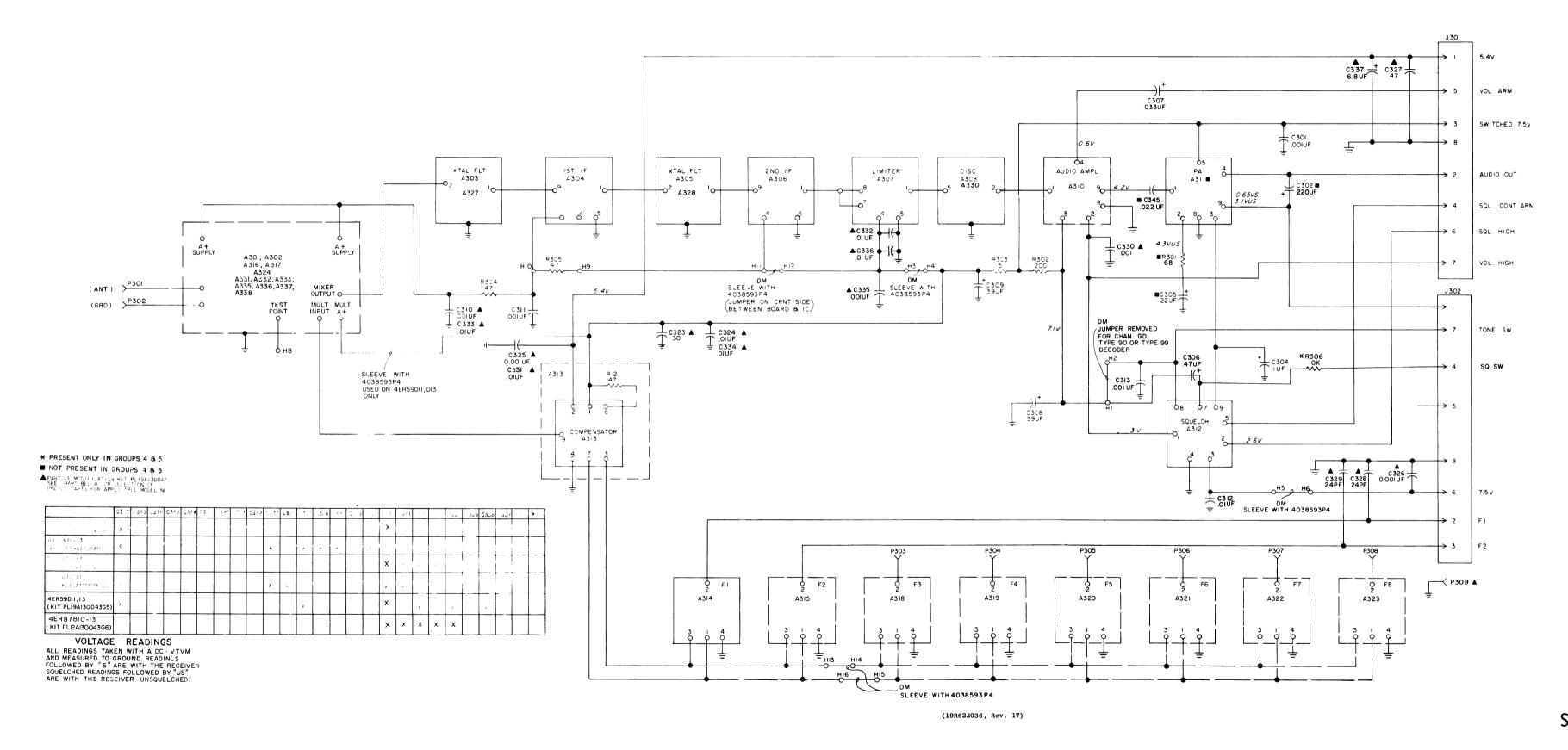


SOLDER SIDE



OUTLINE DIAGRAM

MOBILE DETECTOR (TYPICAL VIEW)



SEE APPLICABLE PRODUCTION CHANGE SHEETS IN INSTRUCTION BOCK SECTION DEALING WITH THIS UNIT, FOR DESTRIPTION OF CHANGES UNDER EACH REVISION LETTER

THIS ELEM DIAG APPLIES TO MCDEL NO REV LETTER

PLI9D417493G1 G
PLI9D417493G2 F
PLI9D417493G3 F
PLI9AI30043G1 A
PLI9AI30043G1 A
PLI9AI30043G3 A
PLI9AI30043G6 A
PLI9AI30043G6 B
PLI9BAI30043G6 A
PLI9BAI30043G6 A
PLI9BAI30043G6 A
PLI9BAI30043G6 A
PLI9BAI30043G6 A
PLI9BAI30043G6 A

ALL RESISTORS ARE 1/8 WATT UNLESS OTHERWISE SPECIFIED AND RESISTOR VALUES IN OHMS UNLESS FOLLOWED BY K-1000 OHMS ON MEG - 1,000,000 OHMS CAPACITOR VALUES IN PICOFFRADS IGUAL TO MICROMICROFARADS JUNLESS FOLLOWED BY UF- MICROFARADS. INDUCTANCE VALUES IN MICROFILENRYS UNLESS FOLLOWED BY MH- MILLIHENRYS OR H-HENRYS.

IN ORDER TO RETAIN RATED EQUIPMENT PERFORMANCE, REPLACEMENT OF ANY SERVICE PART SHOULD BE MADE ONLY WITH A COMPONENT HAVING THE SPECIFICATIONS SHOWN ON THE PARTS LIST FOR THAT PART.

SCHEMATIC DIAGRAM

MOBILE DETECTOR (TYPICAL VIEW)

PARTS LIST

LBI 30036C

150.8-174 MHz RECEIVER MODEL 4ER59D11 STANDARD MODEL 4ER59D13 CHANNEL GUAR

SYMBOL	GE PART NO.	DESCRIPTION	
A338		FRONT END 19C317295G10 (Replaces 19C317295G8)	
A16		MULTIPLIER/MIXER 19C327738G1	
Al		MIXER BOARD 19D424746G1	
C1 and C2	19A116114P2045	Ceramic: 30 pf ±5%, 100 VDCW; temp coef -80 PPM.	
C3	5495323P12	Ceramic: 0.001 µf +100% -20%, 75 VDCW.	
C4	19A116114P3036	Ceramic: 15 pf $\pm 5\%$, 100 VDCW; temp coef -150 PPM.	
C5	5495323P12	Ceramic: 0.001 µf +100% -20%, 75 VDCW.	
C6	19Al16114P10073	Ceramic: 180 pf ±10%, 100 VDCW; temp coef -3300 PPM.	
C7 and C8	19A116114P3036	Ceramic: 15 pf $\pm 5\%$, 100 VDCW; temp coef -150 PPM.	
C9	5495323P12	Ceramic: 0.001 µf +100% -20%, 75 VDCW.	
C10	19A116114P1	Ceramic: 1 pf ±5%, 100 VDCW; temp coef 0 PPM.	
C11	19A116114P2020	Ceramic: 6 pf ±5%, 100 VDCW; temp coef -80 PPM.	
C12	19All6114P3056	Ceramic: 56 pf ±5%, 100 VDCW; temp coef -150 PPM.	
C13	19A116114P2060	Ceramic: 75 pf ±5%, 100 VDCW; temp coef -80 PPM.	
C14	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie 8121 SPECIAL.	
		DIODES AND RECTIFIERS	l
CR1	19All6925Pl	Silicon.	
L1	19B226718G1	Coil. Includes:	İ
	19B209436Pl	Tuning slug.	ĺ
L2 and L3	19A130474P1	Coil.	
L4	19A130473P1	Coil.	
L5	19B226750G1	Coil.	
L6	19C32O379G11	Coil. Includes:	
	198209436P1	Tuning slug.	
Q1	19A115910P1	Silicon, NPN; sim to Type 2N3904.	
Q2	19A116960P1	N Type, field effect; sim to Type 2N4416.	
R1	3R151P182J	Composition: 1.8K ohms ±5%, 1/8 w.	
R1 R2	3R151P1823	Composition: 390 ohms ±10%, 1/8 w.	
R2 R3	3R151P511J	Composition: 510 ohms ±5%, 1/8 w.	
R4	3R151P102J	Composition: 1K ohms ±5%, 1/8 w.	
R5	3R151P682J	Composition: 6.8K ohms ±5%, 1/8 w.	

OL GE PART NO.	DESCRIPTION	SYMBOL	GE PART NO.	DESCRIPTION
	FILTER BOARD 19C327735G1	C306	5491674P27	Tantalum: .47 µf ±20%, 35 VDCW; sim to Sprague Type 162D.
		C307	5491674P31	Tantalum: .033 μf ±20%, 35 VDCW; sim to
19A134162P3	Variable, ceramic: approx 3.5 to 20 pf; sim to Erie Style 513-001.	C308 and	5491674P30	Sprague Type 162D. Tantalum: 39 µf ±20%, 10 VDCW; sim to Sprague Type 162D.
19A116114P1	Ceramic: 1 pf ±10%, 100 VDCW; temp coef 0 PPM.		5495323P12	Ceramic: .001 µf +100% -20%, 75 VDCW.
19A134162P3	Variable, ceramic: approx 3.5 to 20 pf; sim to Erie Style 513-001.	C312	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie
19A116114P4	Ceramic: 1.5 pf ±5%, 100 VDCW; temp coef 0 PPM.	C313	5495323P12	8121 SPECIAL. Ceramic: .001 µf +100% -20%, 75 VDCW.
19A134162P3	Variable, ceramic: approx 3.5 to 20 pr; sim to Erie Style 513-001.	C314*	5495323P12	Ceramic: .001 µf +100% -20%, 75 VDCW. Deleted by REV D.
		C345*	19A116192P6	Ceramic: $0.022 \mu f \pm 20\%$, 50 VDCW; sim to Erie 8131-M050-W5R-223M. Added by REV G.
19A116192P1	Ceramic: 0.01 μ f $\pm 20\%$, 50 VDCw; sim to Erie 8121 SPECIAL.			JACKS AND RECEPTACLES
	INDUCTORS	J301	19A116122P1	Feed-thru: sim to Warren Co 1-B-2994-4.
19B216441G3		and J302		
19C311727P1	Tuning slug.		i	
19B216441G12	Helical resonator. (Part of Z30). Includes:	P301 thru	19A115834P4	Contact, electrical: sim to AMP 2-332070-9.
19C311727P1	Tuning slug.	P308		
1		R301*	3R151P680J	Composition: 68 ohms ±5%, 1/8 w.
13031172771	Tuning Stug.			In REV C & earlier:
	HELICAL RESONATORS		3R151P101J	Composition: 100 ohms ±5%, 1/8 w.
		R302	3R151P201J	Composition: 200 ohms ±5%, 1/8 w.
	Consists of L8 and 19C327717G1 can.			Composition: 15 ohms ±5%, 1/8 w. Composition: 47 ohms ±5%, 1/8 w.
		and R305	SRIGIPHIO	Composition. 41 ones 20%, 1/0 %.
	RECEIVER BOARD 19D417493G1			RECEIVER KIT 19A130043G5
19C304824G1	Crystal Filter.			
	In REV B & earlier:	į.	1	Ceramic: .001 \(\mu f + 100\% - 20\% \), 75 VDCW. Ceramic: .001 \(\mu f + 100\% - 20\% \), 75 VDCW.
19C304516G3	Crystal Filter.	C323	5495323P12	Ceramic: $.001 \mu f + 100\% - 20\%$, 75 VDCW. Added by
19C311879G3	lst IF Amplifier.	C222	19411619201	REV A. Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie
19C304824G1	Crystal Filter.			8:21 SPECIAL. Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie
19C311879G4	2nd IF Amplifier.			8121 SPECIAL.
19C311876G4	Limiter.			Ceramic: 0.01 μ f $\pm 20\%$, 50 VDCW; sim to Erie 8121 SPECIAL.
19C304504G3	Discriminator.	C337	5491674P39	Tantalum: 6.8 μ f $\pm 20\%$, 15 VDCW; sim to Sprague Type 162D.
19C311878G2	Audio Amplifier. Deleted by REV G.	C338*	19A116192P1	Ceramic: 0.01 μ f ±20%, 50 VDCW; sim to Erie 8121 SPECIAL. Deleted by REV B.
19C311995G4	Audio Amplifier. (Includes Tone Filter).			MISCELLANEOUS
1000110001			19B216316P1	Insulator. (Used with J301 & J302).
19C311877G4	PA. In REV E & earlier:		19B216316P1	
	PA.		19B216316P1	ASSOCIATED PARTS
19C311877G4	PA. In REV E & earlier:	A314	19B216316P1	ASSOCIATED PARTS
19C311877G4	PA. In REV E & earlier: PA.	and A315,		ASSOCIATED PARTS
19C311877G4 19C311877G2 19C311880G4	PA. In REV E & earlier: PA. Squelch.	and A315, A318 thru	19B216316P1 4EG28A11	ASSOCIATED PARTS
19C311877G4 19C311877G2 19C311880G4	PA. In REV E & earlier: PA. Squelch. Compensator.	and A315, A318		ASSOCIATED PARTS
19C311877G4 19C311877G2 19C311880G4 19C320061G1	PA. In REV E & earlier: PA. Squelch. Compensator.	and A315, A318 thru A323	4EG28A11	ASSOCIATED PARTS
19C311877G4 19C311877G2 19C311880G4 19C320061G1 5495323P12	PA. In REV E & earlier: PA. Squelch. Compensator.	and A315, A318 thru		ASSOCIATED PARTS
19C311877G4 19C311877G2 19C311880G4 19C320061G1 5495323P12 19A116178P7	PA. In REV E & earlier: PA. Squelch. Compensator.	and A315, A318 thru A323	4EG28A11	ASSOCIATED PARTS
	19A134162P3 19A116114P1 19A134162P3 19A116114P4 19A134162P3 19A116192P1 19B216441G3 19C311727P1 19B216441G17 19C311727P1 19B216441G17 19C311727P1 19C311727P1 19C311727P1 19C311727P1 19C311727P1 19C311727P1	FILTER BOARD 18C3277736G1	Filter Board Code Code	PILTER ROADD 19032777361 C306 S491674P27 S491674P20 S4916192P1 Tuning slug.

*COMPONENTS ADDED, DELETED OR CHANGED BY PRODUCTION CHANGES

PRODUCTION CHANGES

Changes in the equipment to improve performance or to simplify circuits are identified by a "Revision Letter", which is stamped after the model number of the unit. The revision stamped on the unit includes all previous revisions. Refer to the Parts List for descriptions of parts affected by these revisions.

REV. A - Receiver Board 19D417493G1

To improve audio sensitivity. Changed R301.

REV. B - To improve critical squelch operation. Changed C312.

REV. C - To improve producibility. Changed A303.

- REV. D To improve audio sensitivity.
- REV. E To improve frequency response.
 Added C345 to be used with CG receivers.
- REV. A Receiver Kit 19A130043G5

To improve IF filtering. Added C330.

- REV. B To standardize assemblies.

 Deleted C338 and added five washers.
- REV. F Receiver Board 19D417493G1

To improve audio quality. Changed A311.

DESCRIPTION

Receiver Type ER-87-B models are single conversion, superheterodyne FM receivers for operation on the 66-88 MHz bands. The complete receiver mounts on a single printed wiring board, and utilizes both discrete components and Integrated Circuit modules. The application of each model receiver is shown in the following chart:

MODEL NUMBER	FREQ. RANGE	NO. OF FREQ.	TONE OPTION
4ER87B10	66-76 MHz	1 thru 8	
4ER87B11	75-88 MHz	1 thru 8	
4ER87B12	66-76 MHz	1 thru 8	Channel Guard
4ER87B13	75-88 MHz	1 thru 8	Channel Guard

References to symbol numbers mentioned in the following text are found on the Schematic Diagram, Outline Diagram and Parts List (see Table of Contents). The typical circuit diagrams used in the text are

representative of the circuits used in the Integrated Circuit modules. A block diagram of the receiver is shown in Figure 1.

Supply voltage for the receiver includes a continuous regulated 5.4 Volts for the compensator module, a continuous 7.5 Volts for the squelch module, and a switched 7.5 Volts for the remaining receiver stages.

CIRCUIT ANALYSIS

OSCILLATOR MODULES A314, A315 and A318 THRU A323

Oscillator Modules A314, A315 and A318 thru A323 (4EG28A30/31) consist of a crystal-controlled Colpitts oscillator (see Figure 2). An entire oscillator module is contained in a metal can with the receiver operating frequency printed on the top. The crystal frequency ranges from 17.2 - 22.2 MHz, and the crystal frequency is multiplied 5 times.

The oscillator frequency is temperature compensated to provide instant frequency compensation, with a frequency stability of $\pm .0002\%$ from 0°C to $+55^{\circ}\text{C}$

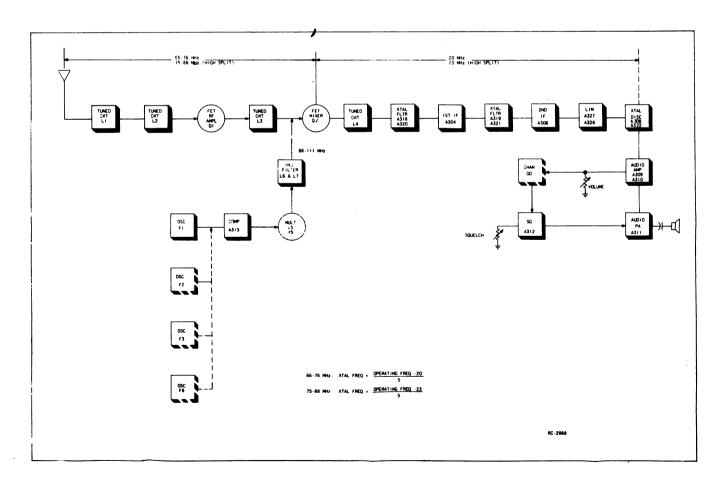


Figure 1 - Block Diagram

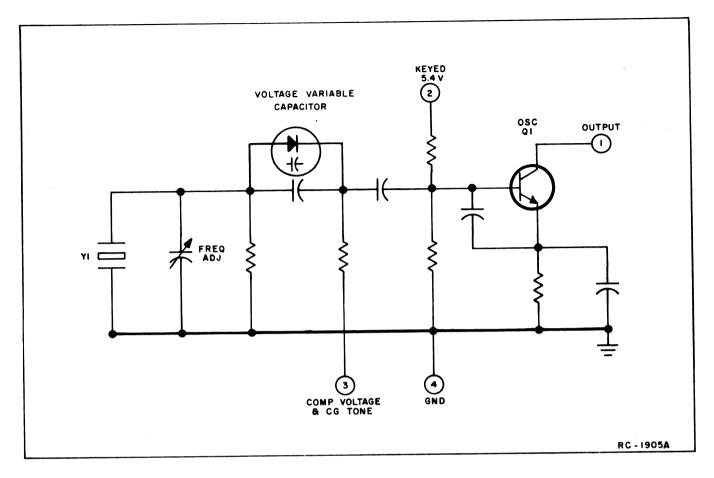


Figure 2 - Typical Oscillator Circuit

and $\pm .0005\%$ from -30° C to $+60^{\circ}$ C. The temperature compensation network contained in compensator module A313.

In single frequency receivers, a jumper from H10 to H11 on the system board connects the oscillator module to the continuous 5.4 Volt supply voltage. The oscillator output is applied to compensator A313.

In two-frequency receivers, an additional oscillator module is mounted on the receiver board. The single-frequency supply jumper is removed, and the proper frequency is selected by connecting the 5.4 Volts to the selected oscillator module through frequency selector switch S1 on the control unit.

-NOTE-

All oscillator modules are individually compensated at the factory and cannot be repaired in the field. Any attempt to remove the oscillator cover will void the warranty.

COMPENSATOR A313

Compensator module A313 contains a buffer-amplifier stage, and the temperature compensation network for the oscillator similar to the compensator used in the transmitter (see Figure 3).

RF from the oscillator is coupled through a DC blocking capacitor to the base of Ql. The output of Ql connects to multiplier coil Ll on the multiplier assembly.

In the compensation network, the regulated 5.4 Volts at Pin 2 is applied to a thermistor-compensated voltage divider. The output at Pin 3 (2.35 Volts measured with a VTVM) is applied to Pin 3 and to the varactor in the oscillator module. At temperatures below -10°C, the compensated voltage increases to maintain the proper voltage on the oscillator voltage-variable capacitor.

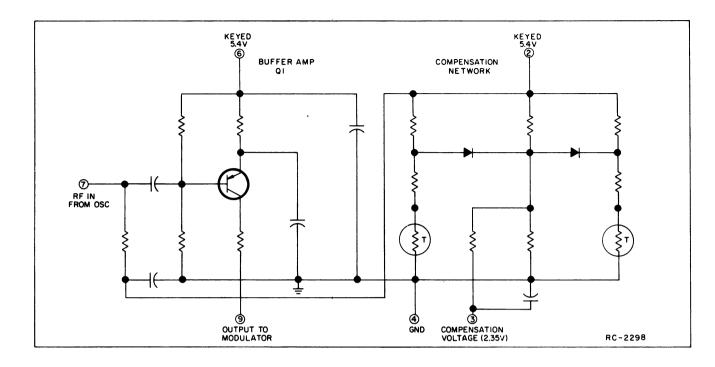


Figure 3 - Typical Compensator Circuit

-SERVICE NOTE-

An abnormally low VTVM reading (or no reading) at Pin 3 may indicate a short or leakage path in the oscillator. This can be checked by unsoldering Pin 2, raising it off of the printed board and taking another reading. If this reading is normal, the problem is in the oscillator module. If the reading remains low (or zero), the problem is in the compensator.

FRONT END A336/A337

The receiver front end consists of three tuned RF coils, a Field Effect Transistor (FET) RF amplifier stage, a multiplier stage and a FET mixer stage. The RF signal from the antenna is coupled to a tap on RF coil L1. RF from L1 is coupled to L2. A tap on L2 is connected to the gate of RF amplifier Q1. The output of Q1 developed across tuned circuit C5/C19, C6/C20 and L3 is applied to the gate of mixer Q2.

The output of compensator module A313 is applied to a tap on L5 in the multiplier assembly. Multiplier coil L5 is tuned to five times the crystal frequency. The injection frequency output from L5 is coupled through bandpass filter L6, L7, C11/C22, C12, C14/C23 and C15 to the gate of mixer Q2. High side injection is used.

The RF signal from RF amplifier Q1 and the high side injection frequency from the multiplier applied to the gate of mixer Q2 produces an IF of 20/23 MHz on the drain of Q2. The 20/23 MHz IF is coupled through the mixer drain tuned circuit L4, C7 and C8 to the input of crystal filter A318/A320 or A319/A321.

CRYSTAL FILTERS A318 & A319/A320 & A321

Filter A318/A320 follows the receiver front end and its output is applied to the 1st IF amplifier module. Filter A319/A321 follows the IF amplifier module. The two Crystal Filters provide the major selectivity for the receiver. A318 & A319/A320 & A321 provides a minimum of 85-dB stopband attenuation.

IF AMPS A304 & A306

An IF amplifier module follows each of the crystal filters, and contains the resistor-matching networks for the filters. A typical IF amplifier circuit is shown in Figure 4.

Each of the IF amplifier modules consists of three R-C coupled amplifier stages that are DC series-connected for reduced drain. The two IF modules provide a total gain of approximately 85 dB.

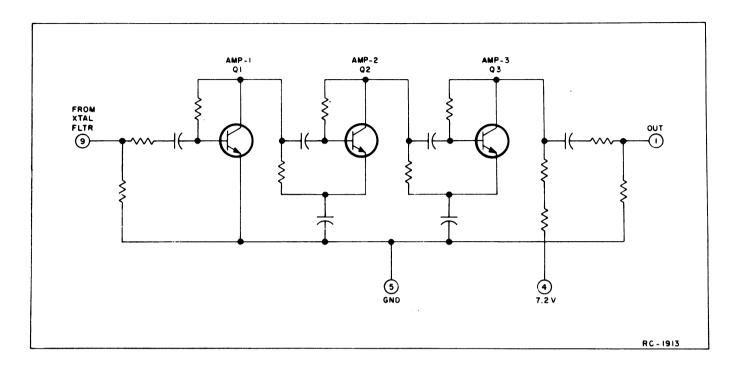


Figure 4 - Typical IF Amplifier Circuit

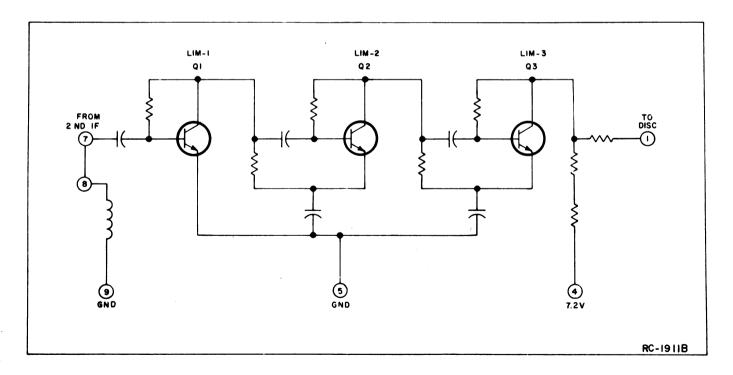


Figure 5 - Typical Limiter Circuit

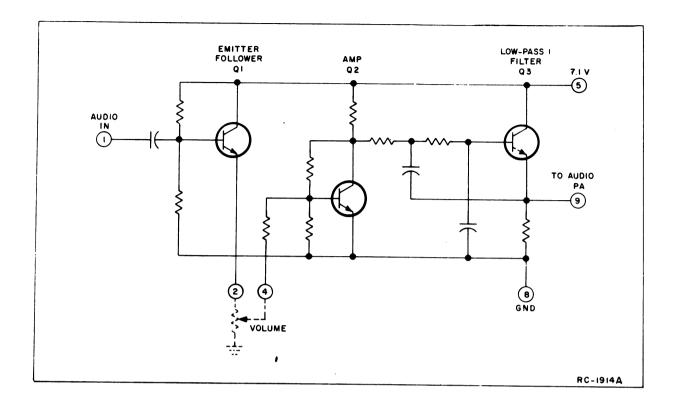


Figure 6 - Typical Audio Amplifier Circuit

LIMITER A327/A328 & DISCRIMINATOR A308/A322

Limiter A327/A328 consists of three R-C coupled limiter stages that are DC series connected for reduced drain. The Limiter module also provides some gain. The output of the Limiter is applied to the discriminator. A typical Limiter circuit is shown in Figure 5.

The receiver uses a 20/23 MHz, fixed-tuned crystal discriminator (A308/A322) to recover the audio from the IF signal. The discriminator output is applied to the audio amplifier module.

AUDIO AMPLIFIER A309/A310

Audio and noise from the discriminator is applied to audio amplifier module A309 (A310 in Channel Guard applications). A typical audio amplifier circuit is shown in Figure 6.

Audio and noise is applied to the base of Q1. This stage operates as an emitter-follower for matching the impedance of the discriminator to the amplifier stage (Q2) and the VOLUME control. The output of Q1 connects from Pin 2 to the base of amplifier Q2 (Pin 4) through the VOLUME control. The output of Q1 is also applied to the input of the Squelch module.

Following amplifier Q2 is an active low-pass filter (Q3). Audio from the filter is connected from Pin 9 to the Audio PA module. In audio amplifier module A323, an active high-pass filter is added in series with the low-pass filter to provide the required tone frequency roll-off.

AUDIO PA A311

When the receiver is quieted by a signal, audio from the active filter is connected to Pin 1 of audio PA module A311, and then to the base of amplifier Q1. Q1 feeds the audio signal to the base of Q2, which drives PA transistors Q4 and Q5. A typical audio PA circuit is shown in Figure 7.

PA transistors Q4 and Q5 operate as complementary emitter-followers, providing a 500 milliwatt output into an 8-ohm load. Audio from Pin 9 is coupled through capacitor C302 on the receiver board to the loud-speaker.

SQUELCH A312

Noise from audio amplifier A309/A310 operates the squelch circuit. A typical squelch circuit is shown in Figure 8.

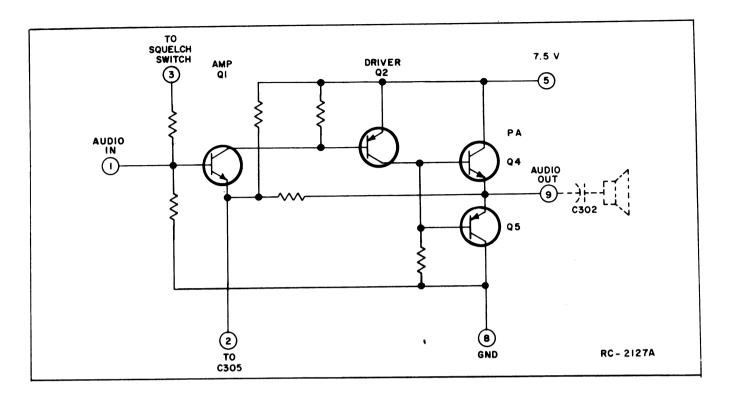


Figure 7 - Typical Audio PA Circuit

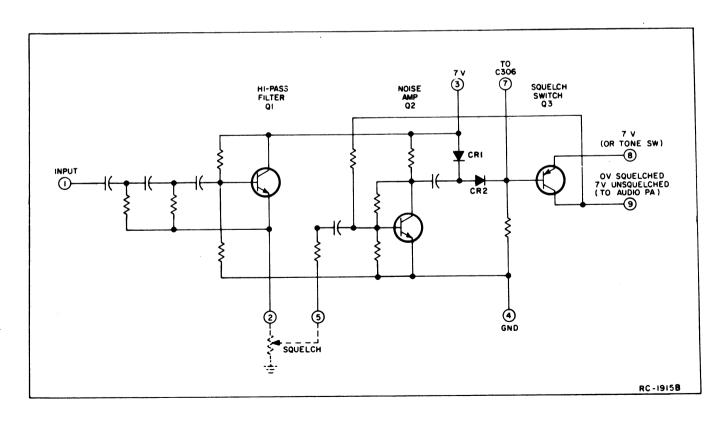


Figure 8 - Typical Squelch Circuit

When no carrier is present in the receiver, the noise output of active high-pass filter Q1 is coupled to the base of noise amplifier Q2 through SQUELCH control R708. R708 controls the gain of the noise amplifier.

The output of noise amplifier Q2 is detected by diodes CR1 and CR2, and the resultant positive voltage turns off the PNP squelch switch Q3. In standard radios, the emitter of Q3 is connected to +7 Volts by means of a jumper from H1 to H2. When noise turns off Q3, its collector drops to ground potential. As the collector of Q3 is connected to the base of amplifier Q1

in the audio PA module, turning off Q3 also turns off Q1, keeping the audio PA turned off.

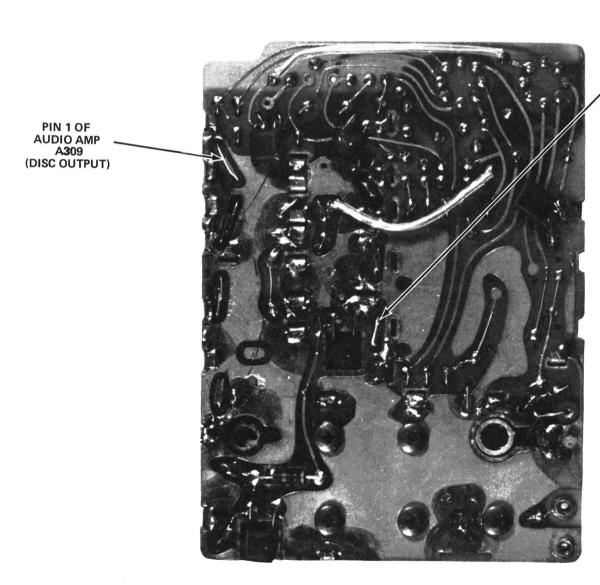
When the receiver is quieted by a signal, squelch switch Q3 turns on. This applies +7 Volts to the base of amplifier Q1 in the audio PA module, turning the audio PA circuit on so that sound is heard at the speaker.

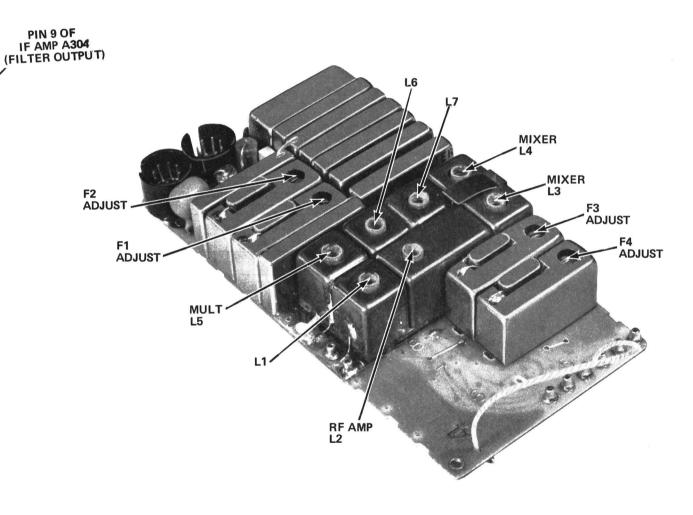
In tone decoder applications, the 7-Volt jumper from H1 to H2 is removed. The emitter of squelch switch Q3 is connected to 7.5 Volts by a DC switch on the decoder.

			•
			•
		•	
		•	

SOLDER SIDE

COMPONENT SIDE





RECEIVER ALIGNMENT

EQUIPMENT

- 1. A 20/23 MHz signal source (GE IF Generator Model 4EX9AlO or equivalent) and a 66-88 MHz source connected to Antenna Switch J702 by Receiver Test Cable 19C317633G1.
- 2. GE Test Amplifier Model 4EX16A10 and RF probe 19C311370G1, or equivalent RF voltmeter.
- 3. Distortion Analyzer or AC-VTVM.
- 4. Oscilloscope

PRELIMINARY CHECKS AND ADJUSTMENTS

- 1. In multi-frequency receivers where the maximum frequency spacing is less than one MHz, align the receiver on the Fl channel. Where the frequency spacing is more than one MHz, align the receiver on the center frequency.
- 2. Set the slugs in L1 thru L3 and L5 thru L7 to the bottom of the coil form for frequencies in the high end of the band. Set the slugs near the top of the coil form for frequencies near the low end of the band.
- 3. Connect the Distortion Analyzer or AC-VTVM across the speaker leads.

RECEIVER ALIGNMENT

STEP	TUNING CONTROL	PROCEDURE	
1.	L1 thru L3 L5 thru L7	Apply an on-frequency signal to J702 and adjust L1, L2, L3, L5, L6 and L7 for maximum quieting sensitivity.	
2.	Mixer L4	Modulate the signal generator with a 20-Hz sawtooth wave and a level of ±20 kHz deviation. Connect the output of the RF detector probe to the vertical input of an oscilloscope. Adjust the signal level until the bandpass is displayed on the scope when the detector probe input is connected to the input of the first IF amplifier A304-9. Adjust L4 for maximum flatness of display. NOTE— Set scope vertical sensitivity to maximum DC sensitivity and sweep rate to 2 ms/cm triggered by the sawtooth wave.	
3.	L1 thru L3 L5 thru L7	With a 10 dB quieting signal level, readjust L1, L2, L3, L5, L6 and L7 for maximum quieting.	
		FREQUENCY ADJUSTMENT	
4.	While applying an on-frequency signal to J702, loosely couple a 20-MHz signal to the Mixer. Adjust the Oscillator trimmer for a zero beat frequency between the two signals. Alternate Method: Apply a strong 20/23 MHz signal to the Mixer. Measure the output of the Discriminator with a DC-VT at Pin 1 of A309/A310. Note the reading. Next, remove the 20/23 MHz signal and apply a strong on-frequency signal to J702. Then tune the oscillator trimmer(s) for the meter reading obtained at Pin 1 of A309/A310.		

ALIGNMENT PROCEDURE

66—88 MHz RECEIVER MODELS 4ER87B10, 11, 12 & 13

TEST PROCEDURES

These Test Procedures are designed to help you to service a receiver that is operating --- but not properly. The problems encountered could be low power, poor sensitivity, distortion, and low gain. By following the sequence of test steps starting with Step 1, the defect can be quickly localized.

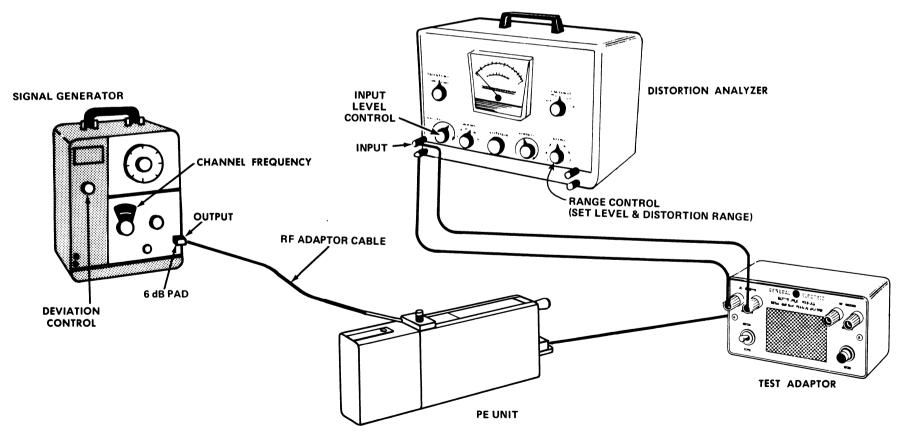
Once the defective stage is pin-pointed, refer to the "Service Check" listed to correct the problem. Additional corrective measures are included in the Troubleshooting Procedure. Before starting with the Receiver Test Procedures, be sure the receiver is tuned and aligned to the proper operating frequency.

TEST EQUIPMENT REQUIRED

- Distortion Analyzer similar to: Heath IM-12
- Signal Generator similar to: Measurements M-800
- 6-dB attenuation pad
- Test Adaptor Model 4EX12A10
- RF Adaptor Cable 19C317633G1

PRELIMINARY ADJUSTMENTS

- 1. Connect the test equipment to the receiver as shown for all steps of the receiver Test Procedure.
- 2. Turn the SQUELCH control fully clockwise for all steps of the Test Procedure.
- 3. Turn on all of the equipment and let it warm up for 20 minutes.



STEP 1

AUDIO POWER OUTPUT AND DISTORTION TEST PROCEDURE

Measure Audio Power output as follows:

- A. Connect a 1,000-microvolt test signal modulated by 1,000 hertz ±3.0 kHz deviation to the Antenna Switch J702.
- B. Set the Volume Control for a 500 milliwatt output (2 volts RMS).
- C. Make distortion measurements according to manufacturer's instructions. Reading should be less than 5%-10% (5% is typical). If the receiver sensitivity is to be measured, leave all controls and equipment as they are.

SERVICE CHECK

If the distortion is more than 5%, or maximum audio output is less than 0.5 watt, make the following checks:

- D. Battery voltage---low voltage will cause
 distortion. (Refer to Receiver Schematic
 Diagram for voltages.)
- E. Audio Gain (Refer to Receiver Trouble-shooting Procedure).

STEP 2

USABLE SENSITIVITY (12 dB SINAD) TEST PROCEDURE

If STEP 1 checks out properly, measure the receiver sensitivity as follows:

- A. Apply a 1000-microvolt, on-frequency signal modulated by 1000 Hz with 3.0-kHz deviation to J702.
- B. Place the RANGE switch on the Distortion Analyzer in the 200 to 2000-Hz distortion range position (1000-Hz filter in the circuit). Tune the filter for minimum reading or null on the lowest possible scale (100%, 30%, etc.)
- C. Place the RANGE switch to the SET LEVEL position (filter out of the circuit) and adjust the input LEVEL control for a +2 dB reading on a mid range (30%).
- D. While reducing the signal generator output, switch the RANGE control from SET LEVEL to the distortion range until a 12-dB difference (+2 dB to -10 dB) is obtained between the SET LEVEL and distortion range positions (filter out and filter in).

- E. The 12-dB difference (Signal plus Noise and Distortion to noise plus distortion ratio) is the "usable" sensitivity level. The sensitivity should be less than rated 12 dB SINAD specification with an audio output of at least 250 milliwatts.
- F. Leave all controls as they are and all equipment connected if the Modulation Acceptance Bandwidth test is to be performed.

SERVICE CHECK

If the sensitivity level is more than rated 12 dB SINAD, check the alignment of the RF stages as directed in the Alignment Procedure, and make the gain measurements as shown on the Troubleshooting Procedure.

STEP 3

MODULATION ACCEPTANCE BANDWIDTH (IF BANDWIDTH)

TEST PROCEDURE

If STEPS 1 and 2 check out properly measure the bandwidth as follows:

- A. Set the Signal Generator output for twice the microvolt reading obtained in the 12-dB SINAD measurement.
- B. Set the RANGE control on the Distortion Analyzer in the SET LEVEL position (1000-Hz filter out of the circuit), and adjust the input LEVEL control for a +2 dB reading on the 30% range.
- C. While increasing the deviation of the Signal Generator, switch the RANGE control from SET LEVEL to distortion range until a 12-dB difference is obtained between the SET LEVEL and distortion range readings (from +2 dB to -10 dB).
- D. The deviation control reading for the 12-dB difference is the Modulation Acceptance Bandwidth of the receiver. It should be more than ±7 kHz (but less than ±9 kHz).

SERVICE CHECK

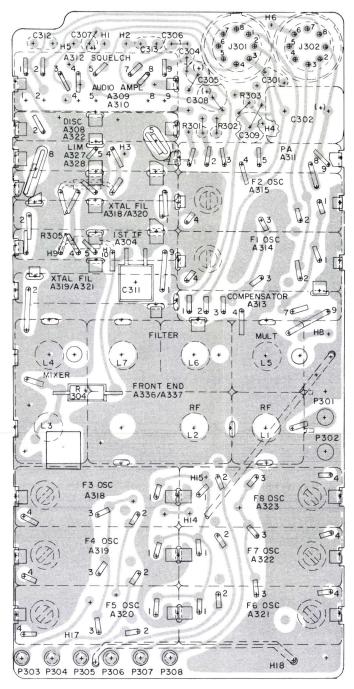
If the Modulation Acceptance Bandwidth test does not indicate the proper width, make gain measurements as shown on the Receiver Troubleshooting Procedure.

RECEIVER BOARD

*OUT A318/A320 IST. IF 8305 A304 HIO4 5 4 4H9 * XTAL FILTER +OUT A319/A321

COMPONENT SIDE

(19D416896, Sh. 2, Rev. 7) (19D416896, Sh. 3, Rev. 6)



SOLDER SIDE

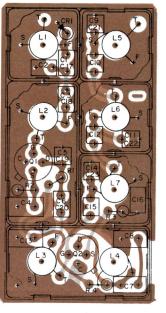
(19D416896, Sh. 2, Rev. 6)

(19D424225, Rev. 0)

RUNS ON SOLDER SIDE RUNS ON BOTH SIDES RUNS ON COMPONENT SIDE

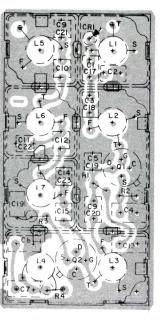
RECEIVER FRONT END

COMPONENT SIDE



(19D423391, Sh. 2, Rev. 0) (19D423391, Sh. 3, Rev. 0)

SOLDER SIDE



(19D423391, Sh. 2, Rev. 0)



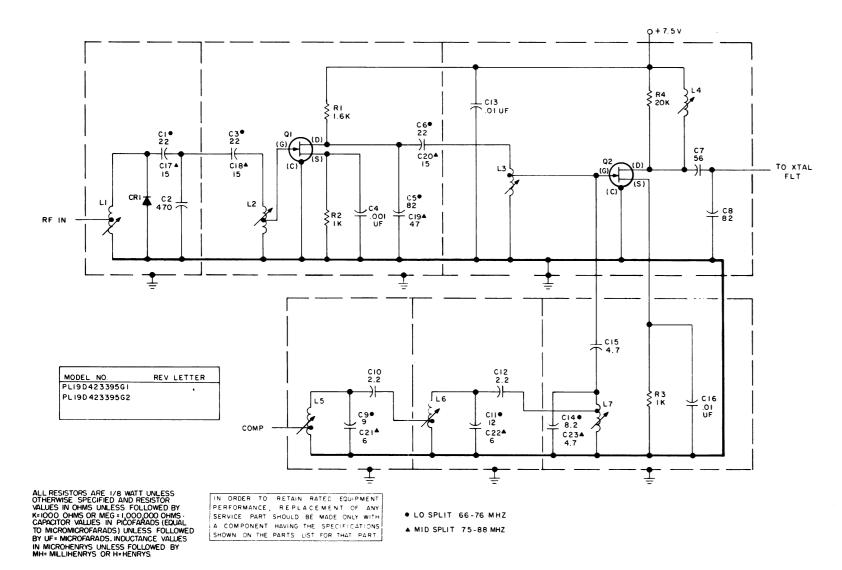
LEAD IDENTIFICATION FOR QI 8 Q2-VIEW FROM LEAD END (19C327318, Rev. 1)

OUTLINE DIAGRAM

66—88 MHz RECEIVER MODELS 4ER87B10, 11, 12 & 13

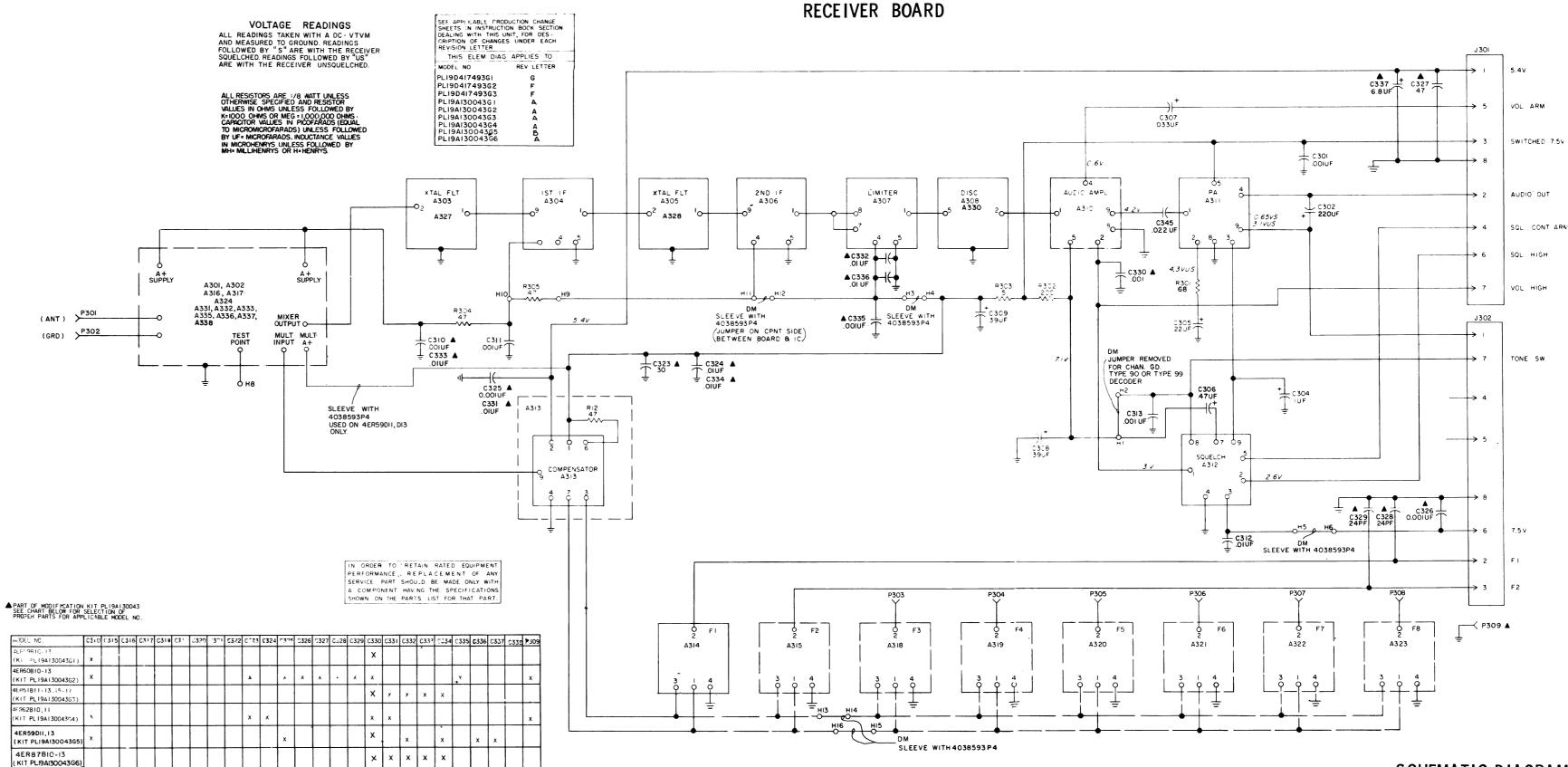
12

Issue 2



(19C321714, Rev. 1)

LBI30188



SCHEMATIC DIAGRAM

13

66-88 MHz RECEIVER MODELS 4ER87B10, 11, 12 & 13

Issue 2

(19R622036, Rev. 15)

LBI30188

14

PARTS LIST

LBI-30189A

66-88 MHz RECEIVER
MODEL 4ER87B10, B11 STANDARD
MODEL 4ER87B12, B13 CHANNEL GUAR

SYMBOL	GE PART NO.	DESCRIPTION
A336 and		FRONT END A336 19D423395G1 66-76 MHz
A337		A337 19D423395G2 76-88 MHz
C1 C2	19A116114P2041 19A116192P2	Ceramic: 22 pf ±5%, 100 VDCW; temp coef -80 PPM. Ceramic: 470 pf ±20%, 50 VDCW; sim to Erie
62		8111-A050-W5R-471M.
C3 C4	19A116114P2041 19A116192P13	Ceramic: 22 pf ±5%, 100 VDCW; temp coef -80 PPM. Ceramic: 1000 pf ±10%, 50 VDCW; sim to Erie
04	194116192913	8121-A050-W5R-102K.
C5	19A116114P6062	Ceramic: 82 pf ±5%, 100 VDCW; temp coef -470 PPM.
C6 C7	19A116114P2041 19A116114P2056	Ceramic: 22 pf ±5%, 100 VDCW; temp coef -80 PPM. Ceramic: 56 pf ±5%, 100 VDCW; temp coef -80 PPM.
C8	19A116114P6062	Ceramic: 82 pf ±5%, 100 VDCW; temp coef -470 PPM.
C9	19A116114P2030	Ceramic: 9 pf ±5%, 100 VDCW; temp coef -80 PPM.
C10	19A116114P8	Ceramic: 2.2 pf ±5%, 100 VDCW; temp coef 0 PPM.
C11	19A116114P2033	Ceramic: 12 pf ±5%, 100 VDCW; temp coef -80 PPM.
C12	19A116114P8	Ceramic: 2.2 pf ±5%, 100 VDCW; temp coef 0 PPM.
C13	19A116192P1	Ceramic: 0.01 μ f \pm 20%, 50 VDCW; sim to Erie 8121-M050-W5R-103M.
C14	19A116114P2028	Ceramic: 8.2 pf ±5%, 100 VDCW; temp coef -80 PPM.
C15	19A116114P2016	Ceramic: 4.7 pf ±5%, 100 VDCW; temp coef -80 PPM.
C16	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie 8121 SPECIAL.
C17 and	19A116114P2036	Ceramic: 15 pf ±5%, 100 VDCW; temp coef -80 PPM.
C18	10411611470052	Commiss, 47 of ±50, 100 VDCW; town coof -90 DDW
C19 C20	19A116114P2053 19A116114P2036	Ceramic: 47 pf ±5%, 100 VDCW; temp coef -80 PPM. Ceramic: 15 pf ±5%, 100 VDCW; temp coef -80 PPM.
C21 and	19A116114P2020	Ceramic: 6 pf ±5%, 100 VDCW; temp coef -80 PPM.
C22 C23	19A116114P2016	Ceramic: 4.7 pf ±5%, 100 VDCW; temp coef -80 PPM.
-		·
CR1	19All5250Pl	Silicon, fast recovery, 225 mA, 50 PIV.
0.1.2		
Ll	19C320379G13	Coil. Includes:
- -	19B209436P1	Tuning slug.
L2	19B226958G2	Coil.
and L3		
L4	19C320379G11	Coil. Includes:
	1B209436P1	Tuning slug.
L5	19C32O379G15 19B2O9436P1	Coil. Includes: Tuning slug.
L6	19B226958G1	Coil.
and L7		
Q1 and Q2	19A116960P1	N Type, field effect; sim to Type 2N4416.

SYMBOL	GE PART NO.	DESCRIPTION	
R1	3R151P162J	Composition: 1.6K ohms ±5%, 1/8 w.	1
R2 and R3	3R151P102J	Composition: 1K ohms ±5%, 1/8 w.	
R4	3R151P203J	Composition: 20K ohms ±5%, 1/8 w.	
		RECEIVER BOARD 19D417493Gl 66-76 MHz 19D417493G3 76-88 MHz	
A303	19C304824G1	Band Pass Filter.	
A304	19C311879G3	IF Amplifier.	
A305	19C304824G1	Band Pass Filter.	
A306	19C311879G4	IF Amplifier.	
A307	19C311876G4	Limiter.	
A308	19C304504G3	Discriminator.	
A309*	19C311878G2	Audio Amplifier. Deleted in G3 by REV F.	
A310	19C311995G4	Audio Amplifier, Tone Filter.	
A311*	19C311877G4	Audio PA. In G2 of REV E & earlier, G3 of REV D & earlier:	
	19C311877G2	Audio PA.	
A312	19C311880G4	Squelch.	١
A313	19C320061G1	Oscillator Compensator.	
		NOTE: When reordering, give GE Part Number and specify exact frequency needed.	
A314, A315, A318 thru A323	4EG28A30	Oscillator Module. 66-76 MHz. $Fx = \frac{F_0 + \frac{F_1}{5}F}{5}$	
A314, A315, A318 thru A323	4EG28A31	Oscillator Module. 75-88 MHz. $Fx = \frac{F0 + 23}{5}$	
A327 and A328	19C304824G3	Band Pass Filter.	
A330	19C304504G6	Discriminator.	
C301	5495323P12	Ceramic: .001 µf +100% -20%, 75 VDCW.	
C302	19A116178P7 19A116089P1	Tantalum: 220 µf ±20%, 6 VDCW. Ceramic: 0.1 µf ±20%, 50 VDCW, temp range -55 to +85°C. Deleted by REV E in Gl. Deleted by REV D	
C304	5491674P28	in G3. Tantalum: 1.0 µf ±20%, 25 VDCW; sim to Sprague	
C305	5491674P35	Type 162D. Tantalum: 22 µf ±20%, 4 VDCW; sim to Sprague	
C306	5491674P27	Type 162D. Tantalum: .47 µf ±20%, 35 VDCW; sim to Sprague	
C307	5491674P31	Type 162D. Tantal COD.	
	5491674P30	Type 162D. Tantalum: 39 µf ±20%, 10 VDCW; sim to Sprague	1

		SYMBOL	GE PART NO.	DESCRIPTION
		C311	5495323P12	Ceramic: .001 µf +100% -20%, 75 VDCW.
		C312	19A116192P1	Ceramic: 0.01 μ f $\pm 20\%$, 50 VDCW; sim to Erie 8121 SPECIAL.
	П	C313	5495323P12	Ceramic: .001 µf +100% -20%, 75 VDCW.
		C314*	5495323P12	Ceramic: .001 μ f +100% -20%, 75 VDCW. Deleted by REV C.
		C345*	19A116192P6	Ceramic: 0.022 µf ±20%, 50 VDCW; sim to Erie 8131-M050-W5R-223M. Added by REV F.
		J301 and J302	19A116122P1	JACKS AND RECEPTACLES Terminal, feed-thru: sim to Warren Co. 1-B-2994
		P301 and P308	19A115834P4	Contact, electrical: sim to AMP 2-332070-9.
		R301*	3R151P680J	
		R301+	3813170003	In REV B & earlier:
			3R151P101J	Composition: 100 ohms ±5%, 1/8 w.
		R302	3R151P201J	Composition: 200 ohms ±5%, 1/8 w.
		R303	3R151P150J	Composition: 15 ohms ±5%, 1/8 w.
		R304 and R305	3R151P470J	Composition: 47 ohms ±5%, 1/8 w.
:				CAPACITOR KIT 19A130043G6
		C330*	5495323P12	Ceramic: .001 µf +100% -20%, 75 VDCW. Added by REV A.
-		C331 thru C334	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie 8121 SPECIAL.
				MISCELLANEOUS
			19B216316P1	Insulator. (Used with J301, J302).
			4035306P11	Washer, fiber. (Used with Q1, Q2 on Front End)
			19A127737P1	Spring, ground tab. (Soldered to Front End can
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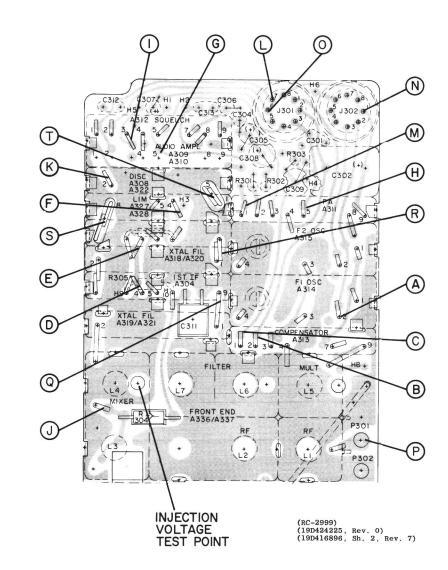
*COMPONENTS ADDED, DELETED OR CHANGED BY PRODUCTION CHANGES

PRODUCTION CHANGES

Changes in the equipment to improve performance or to simplify circuits are identified by a "Revision Letter," which is stamped after the model number of the unit. The revision stamped on the unit includes all previous revisions. Refer to the Parts List for descriptions of parts affected by these revisions.

To improve IF filtering. Added C330.

RE	V. A Thru E -	Receiver Board 19D417493G2
RE	V. A Thru D -	Receiver Board 19D417493G3
		Incorporated into initial shipment.
RE	V. F	Receiver Board 19D417493G2
Re	v. E	Receiver Board 19D417493G3
		To improve audio quality Changed A311.
RE	V. F	Receiver Board 19D417493G3
		To eliminate Non-Channel Guard receiver boards. Deleted callouts of ■A309 and circle (♠) in front of A310. Deleted callout of C303 .1 μf. And just the circle (♠) for C345. Deleted NOTES: ■Use for Non-Channel Guard receivers. ♠Use for Channel Guard receivers.
RE	V. A -	Capacitor Kit 19A130043G6



QUICK CHECKS

SYMPTOM	PROCEDURE
No audio	1. Check audio waveform at the top of the Volume Control (see Step 2).
	 If audio is present, check voltage read- ings of Audio and Squelch modules (see Schematic Diagram).
	 If audio is not present, check gain and current readings of Front End and IF modules (see Steps 1 & 3).
Poor Sensitivity	1. Measure the injection voltage for a minimum level of 400 millivolts. If the reading is low, check the output of the Oscillator and Compensator modules with an RF voltmeter.
	 Measure the gain of the Mixer stage (see Step 3). If low, measure the gain of the RF amplifier and IF modules.
Improper Squelch Operation	1. Check the noise waveform at the input to the Squelch module and at Squelch Control high (see Step 2).
	Measure the DC voltages for the Squelch module (squelched and unsquelched).

STEP 3-RF GAIN CHECKS	→ •	(a)	R	S	Ţ	H
EQUIPMENT REQUIRED: 1. RF probe and Test Amplifier Model 4EX16A10 connected to GE Test Set Model 4EX3A11, or an RF voltmeter. 2. A signal generator (M-800 or equivalent) connected to P301 (High) and P302 (Low). PROCEDURE FOR MIXER & IST IF: 1. Switch the Test Set to the Test 1 position and the Test Amplifier to the X50 position. 2. Connect the RF probe across the input of the stage to be measured as shown on the diagram. Increase the signal generator output to obtain a reference reading on Test Set 4EX3A11. Note the Test Set reading and the dB reading on the generator (dB1). 3. Connect the RF probe to the output of the stage to be measured as shown on the diagram. Decrease the generator output until the Test Set reference reading in Step 2 is obtained. Note the dB	P301 TUNING COILS FET RF AMP INJECTION VOLTAGE 400 MV INJECTION FILTER L7 L6	TUNING COIL L4 XTAL FILTER A318/A320 A304 TV AMA T	2 ND IF 42 dB TAL FILTER 1 9 2 ND IF A306 4 7.3 V 1.1 MA	1	5 DISC AJOB/AJ22 2 AJOB/AJ309/AJ30 3 VOLUME VOLUME	O.6 MA SQUELCHED 125 MA O.5 WATTS OUT AUDIO PA A311 SQUELCH A3 12 SQUELCH SQUELCH
reading on the generator (dB2). 4. Subtract the dB1 reading from the dB2 reading and check the results with the typical gains shown on the diagram. 35 dB (dB2) Example: 35 dB (dB1) 20 dB gain PROCEDURE FOR 2ND IF: 1. With no signal in, connect the RF probe to the output of the 2nd IF module. Increase the signal generator output until the Test Set reading increases by approximately 0.2 volt. Note Test Set and signal generator reading (dB2). 2. Connect the probe to the input of the 2nd IF module. Increase the	OSC A314/A315 COMP A313 (2) (2) (1)	STEP 2 -		SCOPE SETTINGS	(K) (L) (M) (7 V 0.5 MA
signal generator until the Test Set reference reading is obtained, and note the dB reading (dB1). 3. Now subtract dB2 from dB1 to obtain the gain of the 2nd IF amplifier module. LIMITER CHECK The Limiter module limits the noise so that the gain of the circuit cannot be measured. The following procedure provides a check to determine if the module is limiting. 1. Switch the Test Amplifier to the X1 position and the Test Set to the Test 1 position. Then connect the RF probe to the output of the Limiter module and check for a reading of approximately 0.4 volt. 2. Increase the signal generator output. There should be no appreciable increase in the limiter output meter reading.	STEP 1- MODULE CURRENT CHECKS (STEPS A THRU J) These current readings provide a method of checking the operation of each Integrated Circuit module using a milliammeter (Triplett 630 or equivalent). 1. Unsolder the + lead as shown in the Diagram of the module to be checked. 2. Connect the milliammeter in series with the + lead, and check for the indicated current drain and supply voltage. No current drain indicates that the module should be replaced. CAUTION When checking the current of Audio PA module A311, do not	AUDIO & SQUELCH WAV (STEPS K THRU 0) EQUIPMENT REQUIRED:	oints shown and or equivalent). standard signal requency modulated	PEAK-TO-PEAK SIGNAL		6 V

When checking the current of Audio PA module A311, do not short Pin 4 to ground or to + (Pin 5). To do so will destroy the Audio PA module.

TROUBLESHOOTING PROCEDURE

66—88 MHz RECEIVER MODELS 4ER87B10, 11, 12 & 13

Issue 2