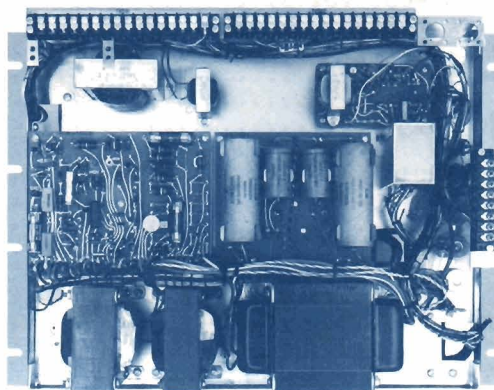


 **MOBILE RADIO**

MASTR PROGRESS LINE

TRANSMITTER-RECEIVER POWER SUPPLY MODEL 4EP38A12 & LINE AMPLIFIER MODELS 4EA24A12, 13



SPECIFICATIONS *

MODEL NUMBER:

4EP38A12

DIMENSIONS (W x H):

19" x 14"

INPUT:

117 VAC $\pm 20\%$, 50/60 Hz, 2.4 amps

OUTPUT:

	132—174 MHz 30 WATTS	25—50 MHz & 66—88 MHz 35/30 WATTS	406—470 MHz 35 WATTS	132—174 MHz 90 WATTS	25—50 MHz 100 WATTS	406—470 MHz 70 WATTS
Bias	-45 V @ 10 mA	-45 V @ 10 mA	-45 V @ 10 mA	-45 V @ 10 mA	-45 V @ 10 mA	-45 V @ 10 mA
Low B+	300 V @ 55 mA	300 V @ 52 mA	300 V @ 52 mA	300 V @ 105 mA	300 V @ 69 mA	300 V @ 105 mA
High B+	450 V @ 160 mA	450 V @ 150 mA	300 V @ 200 mA	680 V @ 220 mA	665 V @ 280 mA	665 V @ 270 mA
Regulated	-20 V @ 80 mA	-20 V @ 60 mA	-20 V @ 80 mA	-20 V @ 80 mA	-20 V @ 60 mA	-20 V @ 100 mA
Regulated	10 V @ 100 mA	10 V @ 100 mA	10 V @ 100 mA	10 V @ 100 mA	10 V @ 100 mA	10 V @ 100 mA
Regulated	12.6 V @ 3.8 amps	12.6 V @ 3.8 amps	12.6 V @ 3.8 amps	12.6 V @ 3.8 amps	12.6 V @ 3.8 amps	12.6 V @ 3.8 amps

FUSES:

F501 - 5 amps, 125 Volts
 F1 - 5 amps, 250 Volts
 F2 - 3/4 amp, 250 Volts
 F3 - 1/2 amp, 250 Volts

DUTY CYCLE:

Continuous

AMBIENT TEMPERATURE
RANGE

-30°C (-22°F) to +60°C (+140°F)

METERING:

All voltages measured at terminal strips on wiring side of power supply board with a 20,000 ohm-per-volt multimeter.

MODEL NUMBER:

4EA24A12, 13

(DM, PM & VM Combinations only)

INPUT POWER:

30 milliamperes @ +10 VDC

OUTPUT IMPEDANCE:

600 ohms

AUDIO FREQUENCY
CHARACTERISTICS:

4EA24A12

Within +1 dB to -3 dB of a 6-dB/octave de-emphasis from 300 to 3000 Hz.

4EA24A13

Within +1 dB to -8 dB of a 6-dB/octave de-emphasis from 300 to 3000 Hz.

DISTORTION:

Less than 5%

*These specifications are intended primarily for the use of the serviceman. Refer to the appropriate Specification Sheet for the complete specifications.

GENERAL  **ELECTRIC**

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WARNING

No one should be permitted to handle any portion of the equipment that is supplied with high voltage; or to connect any external apparatus to the units while the units are supplied with power. KEEP AWAY FROM LIVE CIRCUITS.

DESCRIPTION

POWER SUPPLY

The General Electric Transistorized Power Supply Model 4EP38A12 is a combined transmitter and receiver power supply for MASTR Progress Line Base Stations. The power supply provides:

- High B-plus for transmitter PA
- Low B-plus for the transmitter multiplier stages
- -45 Volts bias for the transmitter power amplifier
- Regulated -20 Volts for the transmitter exciter board
- Regulated +12.6 Volts for heaters, receiver audio, relays and lamps
- Regulated +10 Volts for the transmitter Channel Guard and receiver RF stages

The transmitter and receiver units are mounted on the front side of the Power Supply Panel. Power supply output voltages are connected to the transmitter and re-

ceiver through power cable plugs P103 and P443 respectively.

A fan is mounted on the front panel to provide air-cooling for the transmitter and regulator transistors. The fan turns on when the ambient temperature rises enough to operate thermostat switch S502, or when the station is keyed.

LINE AMPLIFIER

Line Amplifier Models 4EA24A12 and 4EA24A13 are used in MASTR DM, PM and VM Local/Remote and Remote Control station applications for matching the receiver output to a 600-ohm telephone pair. The line amplifier is also used in stations equipped with the receiver voting option.

The line amplifier provides the required de-emphasis, and amplifies the audio to drive the telephone pair. The line amplifier also contains a squelch circuit to eliminate noise feed-through to the telephone pair while the receiver is squelched. The amplifier assembly is mounted on the back of the power supply.

POWER TRANSFORMER HIGH & LOW VOLTAGE TAP CHART

Secondaries for the high voltage and low voltage supplies are tapped and are connected according to the high voltage B-plus required by the transmitter as shown in the chart below. The High B-plus readings shown the chart are typical readings.		
TRANSMITTER RATING	POWER TRANSFORMER SECONDARY TAPS (JUMPERS ON A501)	READING AT HIGH B+ OUTPUT AT A501-H57
132-174 MHz, 30 Watts	H60 to H65 H39 to H41	450 Volts
25-88 MHz, 30 Watts	H60 to H65 H39 to H41	450 Volts
406-470 MHz, 30-35 Watts	H60 to H65 H38 to H39	450 Volts
132-174 MHz, 80 Watts	H59 to H60 H38 to H39	680 Volts
25-50 MHz, 100 Watts	H59 to H60 H38 to H39	665 Volts
406-470 MHz, 60 Watts	H59 to H60 H38 to H39	665 Volts
132-174 MHz and 450-MHz limited 120-Watt input	H59 to H60 H38 to H39 H63 to H65	480 Volts
450 MHz limited 60-Watt input	H39 to H41 Remove F3	300 Volts
250/330-Watt Exciter	H39 to H41 Remove F3	300 Volts

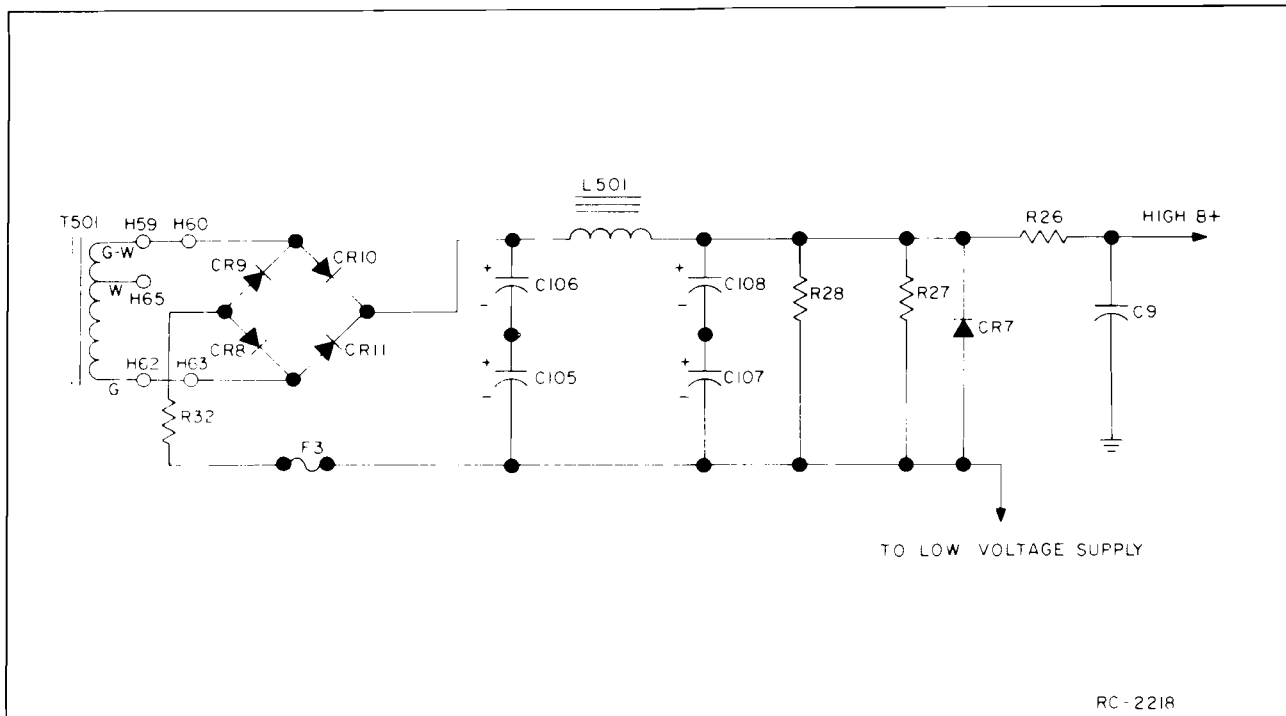


Figure 1 - High Voltage Supply Circuit

CIRCUIT ANALYSIS

POWER SUPPLY

When the power supply ON-OFF switch S501 is turned on, 117 Volts AC at 50/60 Hz from TB502-14 and -15 is applied across the primary (black leads) of power transformer T501. Fuse F501 is in series with switch S501 and TB502-14 to protect the power supply from overload.

The power transformer secondary consists of four windings (two with adjustable taps) to provide AC for the high and low B-plus supplies, bias and regulated supplies.

HIGH VOLTAGE SUPPLY (450-650 VOLTS)

The AC voltage developed across the high voltage secondary winding (green-white, white and green wires) is rectified by the bridge circuit silicon rectifiers CR8, CR9, CR10 and CR11. The rectified voltage is filtered by the pi-filter choke L501, capacitors C105, C106, C107 and C108. The 300 Volts B-plus from the low B-plus supply is "stacked" with the 150- or 350-Volt high B-plus supply (depending on tap setting) to provide the 450- or 650-Volt output of the high voltage supply.

300 Volts from the low voltage supply is connected into the high voltage bridge rectifier circuit at H30 through resistor R32.

Silicon rectifier CR7 is a protective device for the electrolytic filter capacitors. If fuse F3 should blow, reverse voltage across C105 through C107 will be shorted by CR7, thereby preventing damage to the capacitors. Resistors R27 and R28 are bleeder resistors.

LOW VOLTAGE SUPPLY (300 VOLTS)

T501 red, red-white, and orange secondary windings provide the 300-Volt AC voltage which is rectified by the bridge circuit of silicon rectifiers CR3, CR4, CR5 and

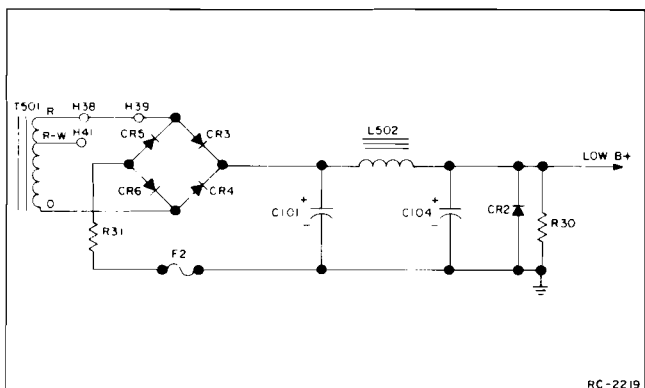


Figure 2 - Low Voltage Supply

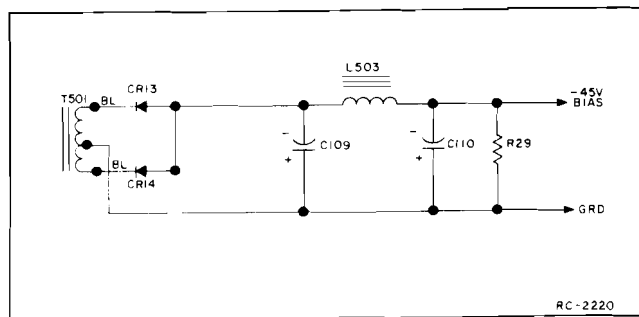


Figure 3 - -45 Volt Bias Supply

CR6. The rectified output voltage is filtered by pi-filter consisting of choke L502, capacitors C101 and C104. The output is in series with the high B-plus circuit to provide "stacking" for the high voltage circuit. Silicon rectifier CR2 is used as a protective device for the electrolytic filter capacitors. If fuse F2 should blow, reverse voltage across C101 and C104 will be shorted by CR2, thereby preventing damage to the capacitors (see Figure 2).

-45 VOLT BIAS SUPPLY (Figure 3)

The AC developed across the two blue wires (with center tap to ground) of the T501 secondary is rectified by full-wave silicon rectifiers CR13 and CR14. The rectified voltage is filtered by pi-filter choke L503 and capacitors C109 and C110 to supply a negative 45-Volt bias.

REGULATED -20 VOLT SUPPLY (Figure 4)

The -45 Volts unregulated is also taken off at the minus side of filter capacitor C110 and connects to the normally open contact 5R on relay K501. When the transmitter is keyed, K501 energizes and contacts 5R and 6R close, applying -45 Volts to the regulator circuit. Voltage dropping resistor R18 provides the negative bias to turn on Q504. Zener diode VR3 provides reference for the regulator.

When the output voltage attempts to go more negative, the voltage at the base of Q4 also goes more negative. This causes a change in the base-emitter bias on Q4, making it conduct more heavily. When Q4 conducts, there is less base bias on Q504, and less base current flow. With less base current flow, the voltage drop across Q504 is larger, and the output voltage tends to remain constant.

When the output voltage starts to go less negative, the forward bias on Q4 decreases. Q4 conducts less and reduces the voltage drop across R18 so that the forward bias on Q504 is increased and the output voltage remains constant.

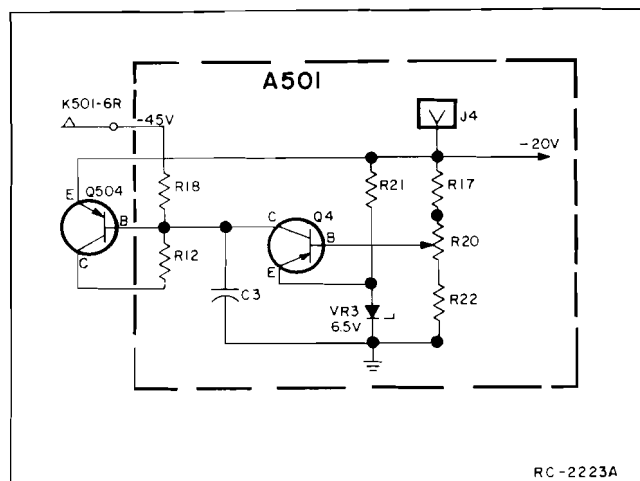


Figure 4 - Regulated -20 Volt Supply

Capacitor C3 prevents high frequency oscillation and also helps to filter the input voltage. R21 provides voltage to operate the zener diode VR3. R17, R20 and R22 form a voltage divider that can be varied by potentiometer R20 to adjust the base voltage of Q1 which sets the output at -20 Volts. This output is measured at jacks J1 and J4 on power supply board A501. The voltage is regulated to -20 Volts $\pm 5\%$.

REGULATED 12.6-VOLT SUPPLY (Figure 5)

The AC developed across the two brown secondary wires of T501 is rectified by full-wave rectifier circuit CR501 and CR502. The output is taken off at the center tap of T501 and is filtered by choke input filter L504 and capacitors C102 and C103. The output of the filter circuit is applied to the emitters of Q501 and Q502 through R1 and R2. When the output of Q501 and Q502 tries to rise, the base of Q1 is made more positive. This increases the current flow through R3 and R4, decreasing the positive voltage at the base of driver transistor Q503. Q503 will then conduct more heavily, causing a greater voltage drop across R501. The bases of Q501 and Q502 will become more positive, causing them to conduct less. This keeps the voltage at the output terminal at the regulated voltage level. R1 and R2 equalize current through Q501 and Q502.

When the output of Q501 and Q502 tries to drop, Q1 will conduct less. This decreases the forward bias on Q503 to reduce the voltage drop across R501. This will cause Q501 and Q502 to conduct more heavily and hold the output voltage constant. Zener diode VR1 provides a voltage reference for regulator Q1. VR501 protects Q503 from current surges. C502 prevents high frequency oscillation in Q503 under light load conditions.

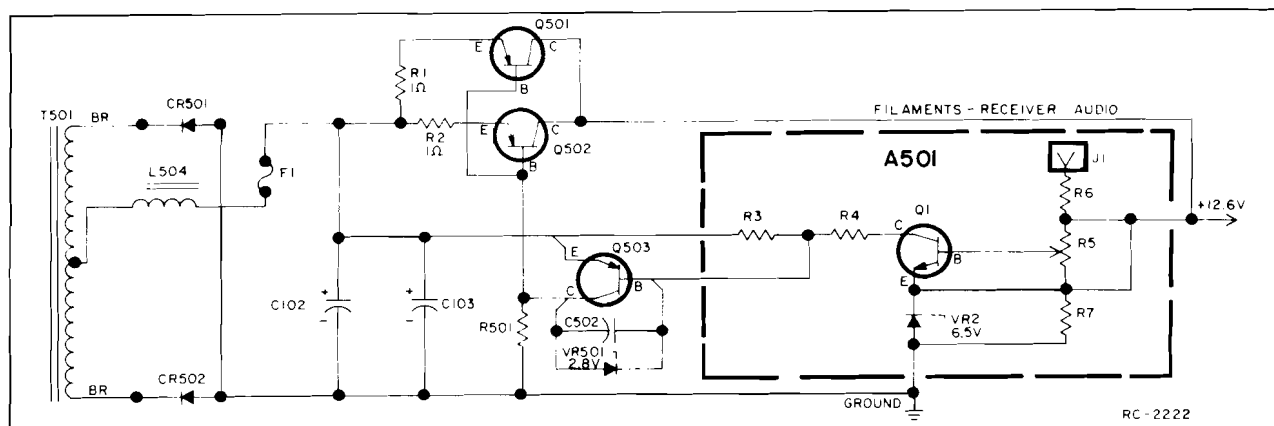


Figure 5 - Regulated 12.6-Volt Supply

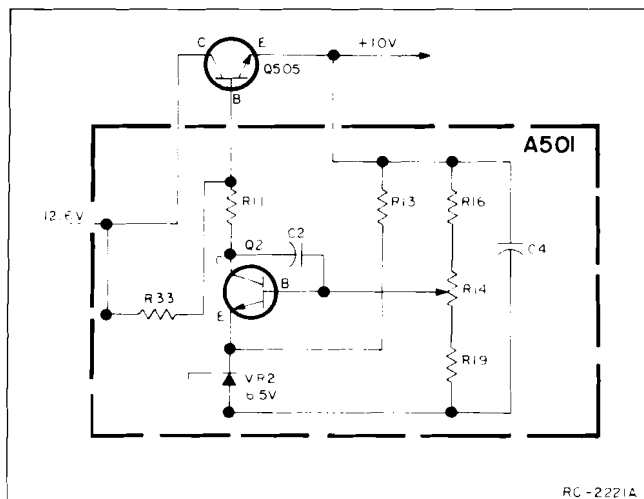


Figure 6 - Regulated +10-Volt Supply

The output is set by potentiometer R6 to produce a +12.6-Volt, $\pm 5\%$ reading. The output is measured at J1 and J2 on A501.

REGULATED +10-VOLT SUPPLY (Figure 6)

The input voltage to the 10-Volt regulator circuit is taken from the +12.6-Volt regulated supply.

When the supply voltage (or output) starts to increase, the voltage at the base of Q2 also increases. As the emitter voltage of Q2 is kept constant by zener diode VR2, the emitter-base voltage increases. This causes Q2 to conduct more which means less base current for Q505. The voltage drop across Q505 becomes larger and the output remains constant.

When the input voltage starts to drop, the output voltage also tends to drop and

Q2 will conduct less. This increases the forward bias on Q505 and reduces the voltage drop across the transistor to keep the output constant.

Potentiometer R14 is used to set the emitter-base voltage of Q2 for the desired 10-Volt, $\pm 5\%$ output. R11 and R16 limit maximum current through Q2. R13 provides bias current for zener diode VR2, and R33 provides bias for Q505. C2 and C4 prevent high frequency oscillation. The output voltage is metered at J3 and J1 (GRD) on A501.

RECEIVER MUTING

Transistor Q3 operates as a switch for the receiver muting +10 Volts. A continuous +10 Volts is applied to the collector of Q2. When the transmitter is unkeyed, +12.6 Volts is applied to the base of the transistor, causing it to conduct. When conducting, the +10 Volts at the emitter of Q2 is coupled through P443-2 to the base of receiver DC amplifier Q9, turning it on. When Q9 conducts, DC amplifier Q10 is turned off, and the receiver operates normally.

Keying the transmitter grounds the base of Q3, turning it off. This removes the +10 Volts to receiver DC amplifier, turning it off. Turning off Q9 causes Q10 to turn on, which turns off the receiver audio amplifiers and mutes the receiver.

VOLTAGE SUPPLY RELAY (K501)

When the transmitter is keyed, K501 becomes energized and the following connections are made:

- High voltage supply stacked on low voltage supply
- Low voltage to transmitter

- Input to 20-Volt regulator
- 12.6 Volts is applied to TB501-15 to light the red transmitter pilot lamp (also mutes additional receivers if used)
- Starts blower

ANTENNA RELAY OPTION (K502)

In the normally closed position, antenna relay K502 connects the receiver to the antenna system. The relay becomes energized when the transmitter is keyed, thereby opening the receiver antenna circuit and connecting the transmitter into the antenna system.

Jack J503 is connected to the common lead of K502 and is mounted into the left hole in the antenna mounting bracket. The incoming antenna transmission line plugs into J503. The receiver antenna cable from K502 plugs into J441 on the receiver and transmitter antenna lead from K502 plugs into J103 on the transmitter.

MICROPHONE INPUT CIRCUIT

In Local/Remote, Remote or Repeater applications, the circuit connected to mike jack J902 is used to isolate the high impedance mike and the control panel which acts as a 600-ohm impedance across TB502-12 and -13.

CR1 is a back-biased diode in series with the audio input from the control panel. A +9-Volt drop across R15 provides +1 Volt with respect to ground at the cathode of CR1 which reverse biases the diode. The mike is then loaded only by the transmitter and R25 (both are relatively high impedances).

The control unit attached to TB502-12 and -13 supplies +10 Volts (when keyed) which is divided by voltage divider R23 and R24, producing 4.5 Volts on the anode of the diode. This forward biases the diode, allowing the audio from the control unit to modulate the transmitter.

C7, R24 and R25 are also used to equalize the high frequency response when the control panel is the audio source.

LINE AMPLIFIER

Line Amplifier Model 4EA24A12 is used in stations without Channel Guard, and Model 4EA24A13 is used in stations with Channel Guard. All connections to the amplifier board are made at TB1501. The LINE LEVEL ADJUST potentiometer R1501 is mounted on the power supply chassis adjacent to the VOLUME and SQUELCH controls. Supply voltage is provided by the power supply 10-Volt

regulator. Receiver audio applied to the line amplifier is taken from volume high (R511-3) on the station power supply.

Model 4EA24A12

Audio from the receiver is applied to the base of 1st audio amplifier Q1. The output of Q1 is applied to de-emphasis network R5 and C2 which provides 6-dB per octave roll-off. Following the de-emphasis network is buffer-amplifier Q2. The output of the buffer stage is coupled through LINE LEVEL ADJUST potentiometer R1501 to the base of 2nd amplifier Q3. R1501 is normally set for 2.7 Volts (+11 dB) at the telephone pair. For standard remote control stations, instructions for setting R1501 are contained in the Adjustment Section (see Table of Contents). For stations equipped with the satellite receiver option, instructions for setting R1501 are contained in the Maintenance Manual for the satellite receiver.

The output of the 2nd amplifier is applied to line driver transistors Q4 and Q5. Q4 operates as a current amplifier and Q5 as a voltage amplifier. The line driver output is coupled through line-matching transformer T1 to line output terminals TB1501-6 and -7. In standard remote stations, the line output is connected to TB701-1 and -2 on remote control panel Model 4KC16A12. For stations with the satellite receiver option, the line output is connected to TB1-4 and TB1-5 on the Tone/Audio board.

The operation of line drivers Q4 and Q5 is controlled by squelch switch transistors Q6 and Q7. When the station receiver unsquelches, the COS feed voltage applied to the base of Q6 rises from zero to approximately 3 Volts DC. This voltage turns on Q6, which turns on Q7. Turning on Q7 completes the current path for the line drivers, turning them on and applying audio to the line output.

When the receiver squelches, the COS feed voltage drops to zero, turning off Q6 and Q7. The +5 Volts developed across divider network R20 and R21 is applied to the emitter of Q5. This reverse-biases Q5, turning the line drivers off.

The action of C6 and R18 provides a slight delay in turning Q7 on or off. This delay prevents a switching "pop" from being heard in the speaker.

Model 4EA24A13

Audio from the receiver is applied to the base of 1st audio amplifier Q1. The amplifier output is coupled through a de-emphasis network (R5 and C7) and applied to buffer amplifier Q2. Following the buffer is a 180 hertz notch filter for attenuating

the Channel Guard tone. The filter consists of C8 through C14, L1 and R25.

The filter output is coupled through LINE LEVEL ADJUST potentiometer R1501 to the base 2nd audio amplifier Q3. R1501 is normally set for 2.7 Volts (+11 dB) at the telephone pair. For standard remote control stations, instructions for setting R1501 are contained in the Adjustment Section (see Table of Contents). For stations equipped with the satellite receiver option, instructions for setting R1501 are contained in the Maintenance Manual for the satellite receiver.

The output of the 2nd amplifier is applied to line driver transistors Q4 and Q5. Q4 operates as a current amplifier and Q5 as a voltage amplifier. The line driver output is coupled through line matching transformer T1 to line output terminals TB1501-6 and -7. In standard remote stations, the line output is connected to TB701-1 and -2 on remote control panel Model 4KC16A12. For stations with the satellite receiver option, the line output is connected to TB1-4 and TB1-5 on the Tone/Audio board.

The operation of line drivers Q4 and Q5 is controlled by squelch switch transistors Q6 and Q7.

When the station receiver unsquelches, the COS feed voltage applied to the base of Q6 rises from zero to approximately 3 Volts DC. This voltage turns on Q6, which turns on Q7. Turning on Q7 completes the current path for the line drivers, turning them on and applying audio to the line output.

When the receiver squelches, the COS feed voltage drops to zero, turning off Q6 and Q7. The +5 Volts developed across divider network R20 and R21 is applied to the emitter of Q5. This reverse-biases Q5, turning the line drivers off.

The action of C6 and R18 provides a slight delay in turning Q7 on or off. This delay prevents a switching "pop" from being heard in the speaker.

INITIAL ADJUSTMENT

POWER SUPPLY

The adjustment for the power supply includes turning on power switch S501, and adjusting VOLUME control R511 and SQUELCH control R512 as directed in the applicable procedure.

Local and Local/Remote Stations

Set the VOLUME control (R511) to approximately mid-range, and set the SQUELCH control (R512) fully clockwise. This will

enable the volume and squelch controls on the front panel to be adjusted for operation near mid-range.

Remote Control Stations

Set the VOLUME control (R511) to approximately mid-range, and set the SQUELCH control (R512) for quieting.

Repeater Stations

Set the VOLUME control (R511) for the desired listening level, and set the SQUELCH control (R512) for quieting.

LINE AMPLIFIER

In standard remote control stations, adjustment for the Line Amplifier consists of setting LINE LEVEL ADJUST R1501. To set R1501:

1. Make sure that TB701-1 and -2 on the remote control panel are terminated with the 600-ohm telephone pair, or a 620-ohm, 1/2-Watt resistor.
2. Connect a signal generator to the receiver antenna jack and apply a 1000 microvolt, 1000 Hz signal with two-thirds rated system deviation.
3. Adjust R1501 for an AC-VTVM reading of 2.7 Volts RMS (+11 dB) at TB701-1 and -2 on the remote control panel.

For stations equipped with the satellite receiver option, refer to the Maintenance Manual for the Satellite Receiver or Voting Selector.

MAINTENANCE

TEST CABLE

A 26-inch coax transmitter test cable is clipped to the outside of the fan ventilating cover on the front side of the power supply so that the transmitter can be swung out for servicing.

To troubleshoot the transmitter, disconnect the cable plug P103 from transmitter jack J103 (refer to Outline Diagram); then remove the extension cable from the ventilating cover and plug one end of the extension cable into J103 on the transmitter and the other end into P103 on the cable that was connected to the transmitter.

MUFFIN FAN LUBRICATION

Muffin fans equipped with a metal bearing cap are permanently lubricated and will require no oiling. For fans equipped with a rubber cap, Oiler Kit No. 19263 (list price \$4.25 subject to change) is recommended for oiling the muffin fan. This kit permits oiling without removing the fan from its mounting. A syringe and a supply of oil are provided in the kit. Lubricate the fan at intervals as shown in the following chart.

For Ambient Temperature of: (Approx.)	Lubricate Fan:
80°F or Less	Every 18 to 24 months
100°F	Every 12 to 15 months
120°F or greater	Every 6 to 8 months

A good grade of light instrument oil, such as Aeroshell Fluid No. 12 or Esso Univis® P-38 should be used.

RELAY SERVICING

The relays in these units require

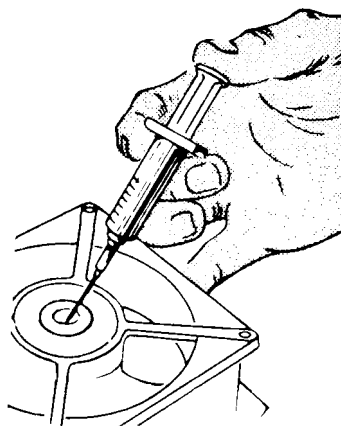
little care. However, they should be inspected periodically to assure maximum operating efficiency. If the contacts become pitted, they should be cleaned with a burnishing tool to smooth out any metallic deposits. When relay contacts carry little or no current, the contacts do not clean themselves and an insulating coating is apt to form. This coating may be removed by cleaning the contacts with a burnishing tool. Do not oil the relay bearings. When relays are in dusty locations, lubricated bearings will collect dust and grit, and will wear more rapidly than non-lubricated bearings.

Some of the relays used are of the multiple-contact type and, in the unenergized position, should have contact spacings of approximately .010 to .020 inch. More important, the contact spacings on any multiple-contact relay should be equal so that the contact pressures will be equal when the relay is energized. The back pressure of the antenna relay should be at least 15 grams. Low back pressure will shorten the life of contacts, due to excessive arcing, and may also cause noise in the receiver due to chatter of the antenna relay contact under vibration.

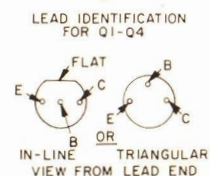
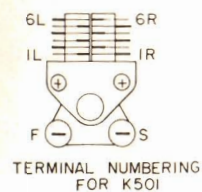
OILING INSTRUCTIONS

To inject oil in the bearing:

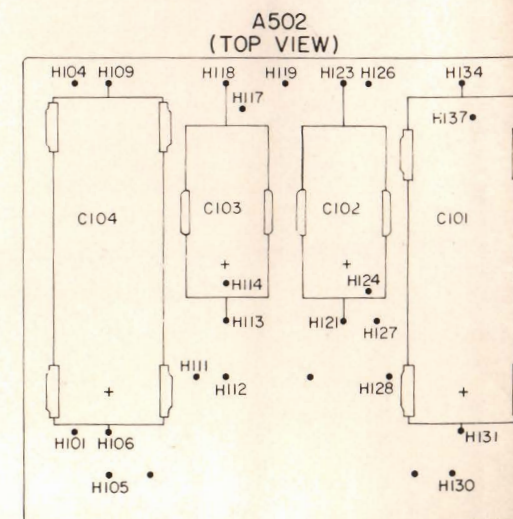
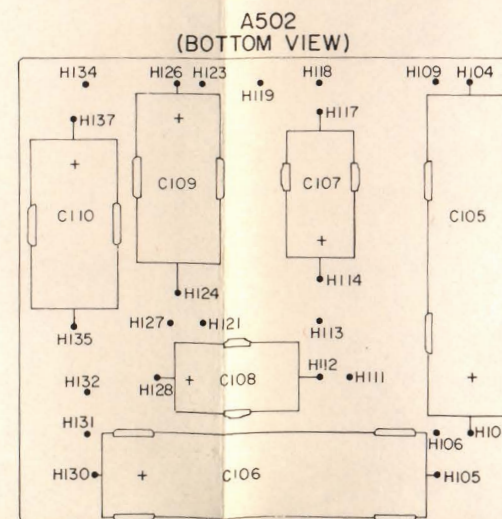
1. Position the needle at an angle of 45° as shown, and pierce the rubber cap.
2. Depress plunger firmly until oil has gone down one calibration line.
3. Withdraw the needle and wipe off excess oil. Oil may be left in the syringe for future use.

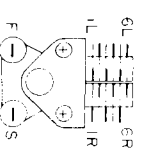
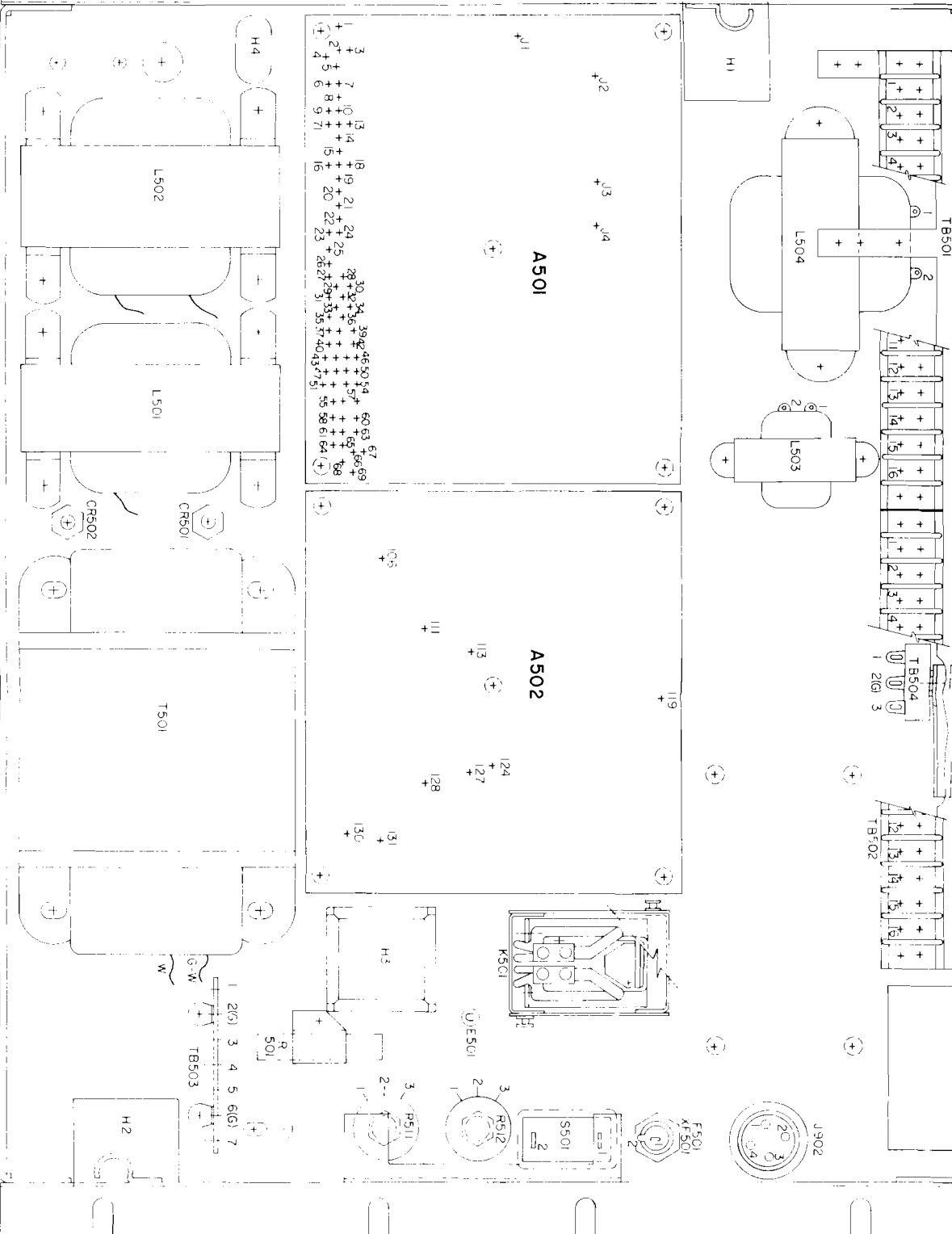


SYMPTOM	CHECK THE FOLLOWING:
No receiver noise when SQUELCH control opened	<ol style="list-style-type: none"> 1. Fuse F501 and F1 2. Open Q3 collector 3. 12.6-Volt regulator 4. 10-Volt regulator
No high B+ but low B+ correct	<ol style="list-style-type: none"> 1. Fuse F3 2. Resistor R32 3. Relay contacts K501-5L and -6L, or -3L and -4L. 4. Resistor R26 5. Diode CR12
No low B+	<ol style="list-style-type: none"> 1. Fuse F2 2. Resistor R31 3. Relay contacts K501-1L and -2L
No -45 Volts bias	Diodes CR13 or CR14
Can modulate transmitter from local microphone, but cannot from remote microphone	Diode CR1 and associated circuitry
Receiver work mute	<ol style="list-style-type: none"> 1. Shorted Q3 2. Excessive resistance in PTT line
10-VOLT REGULATOR	
No 10-Volt regulated output and DS1 burning brightly	Low resistance short on regulator output
No 10-Volt regulated output and DS1 not burning brightly	<ol style="list-style-type: none"> 1. Open Q505 2. Open Q2 3. Resistor R33
Output voltage too high, and cannot be adjusted by R14	<ol style="list-style-type: none"> 1. Open VR2 2. Open Q2 3. Shorted Q505 4. R14 defective
Very low output voltage	Shorted VR2
-20 VOLT REGULATOR	
No -20 Volt output	<ol style="list-style-type: none"> 1. Open Q504 2. Relay contacts K501-5R and -6R
Very low output voltage	Shorted Q4 or VR3
Output voltage too high	Open VR3 or Q4
Output equals input	Shorted Q504

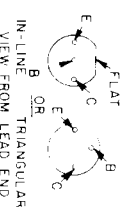


NOTE: LEAD ARRANGEMENT, AND NOT CASE SHAPE, IS DETERMINING FACTOR FOR LEAD IDENTIFICATION.





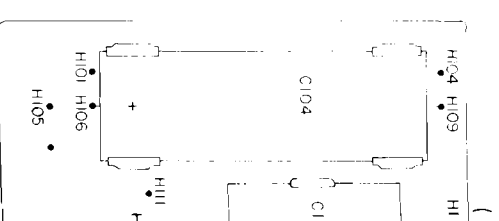
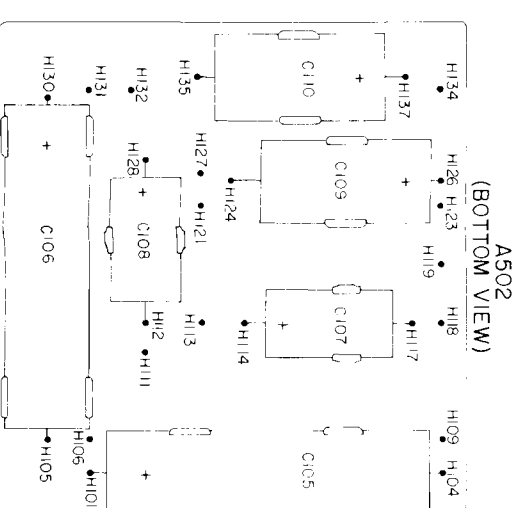
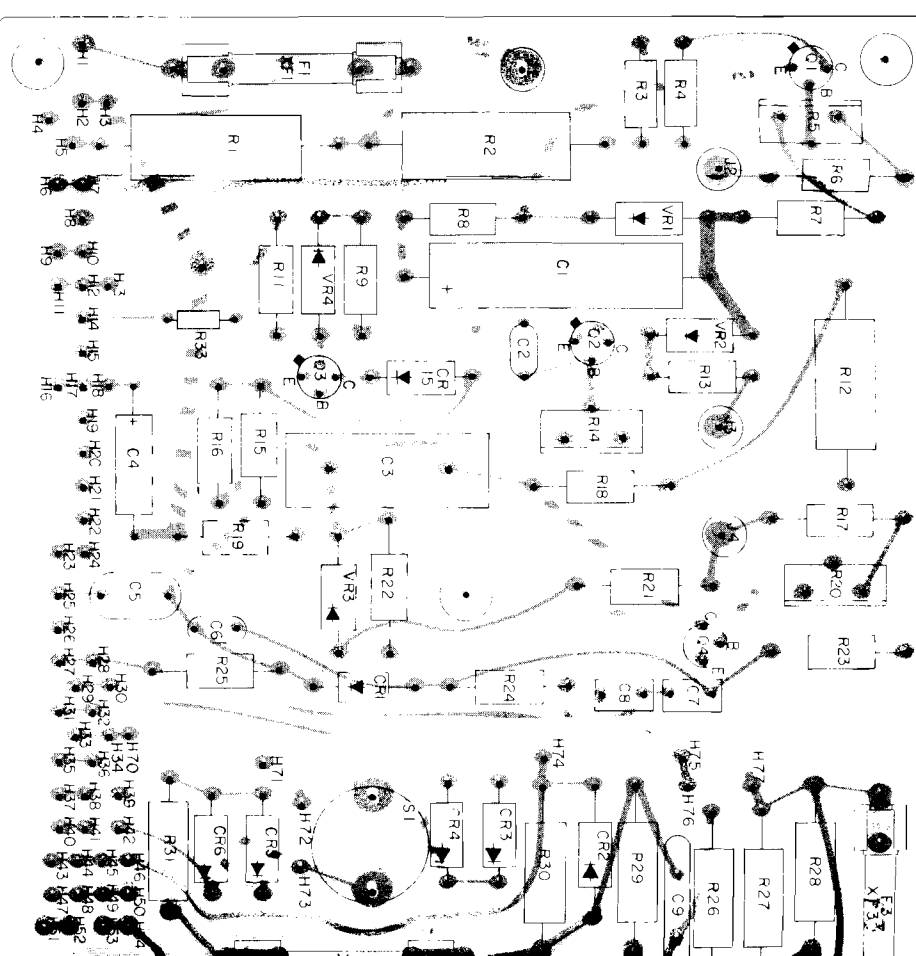
TERMINAL NUMBERING FOR K501



LEAD IDENTIFICATION
FOR Q1-Q4

IN-LINE TRIANGULAR
VIEW FROM LEAD END

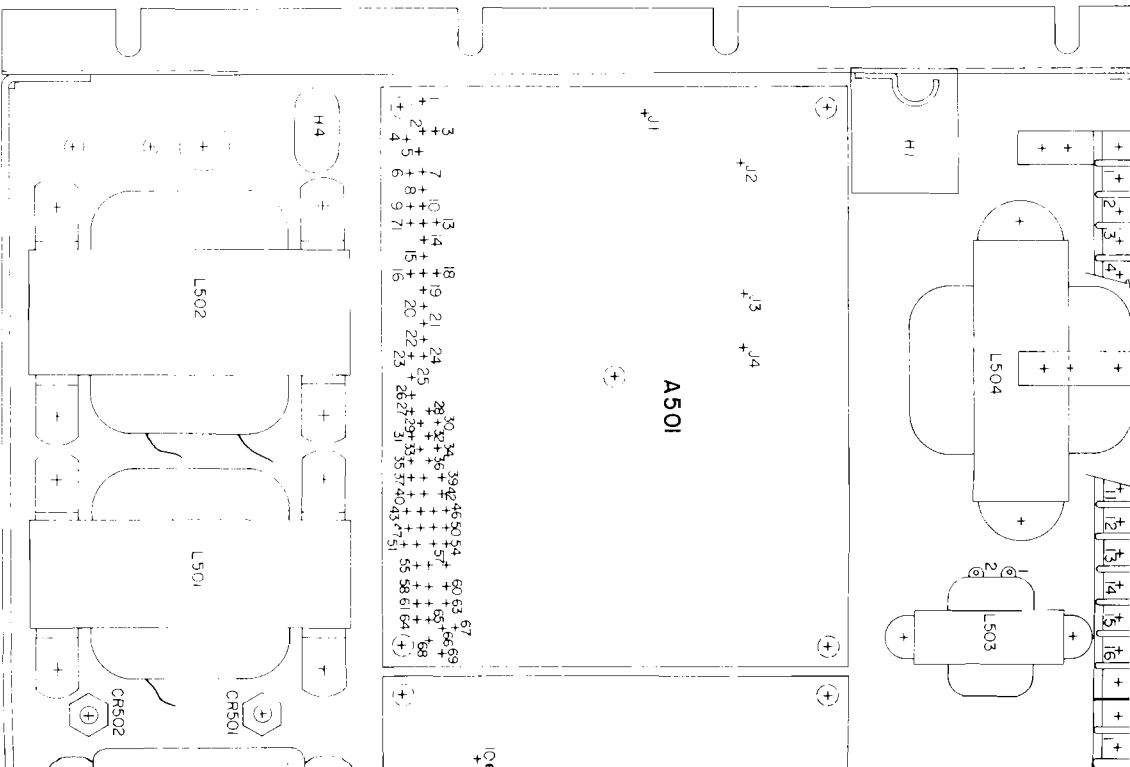
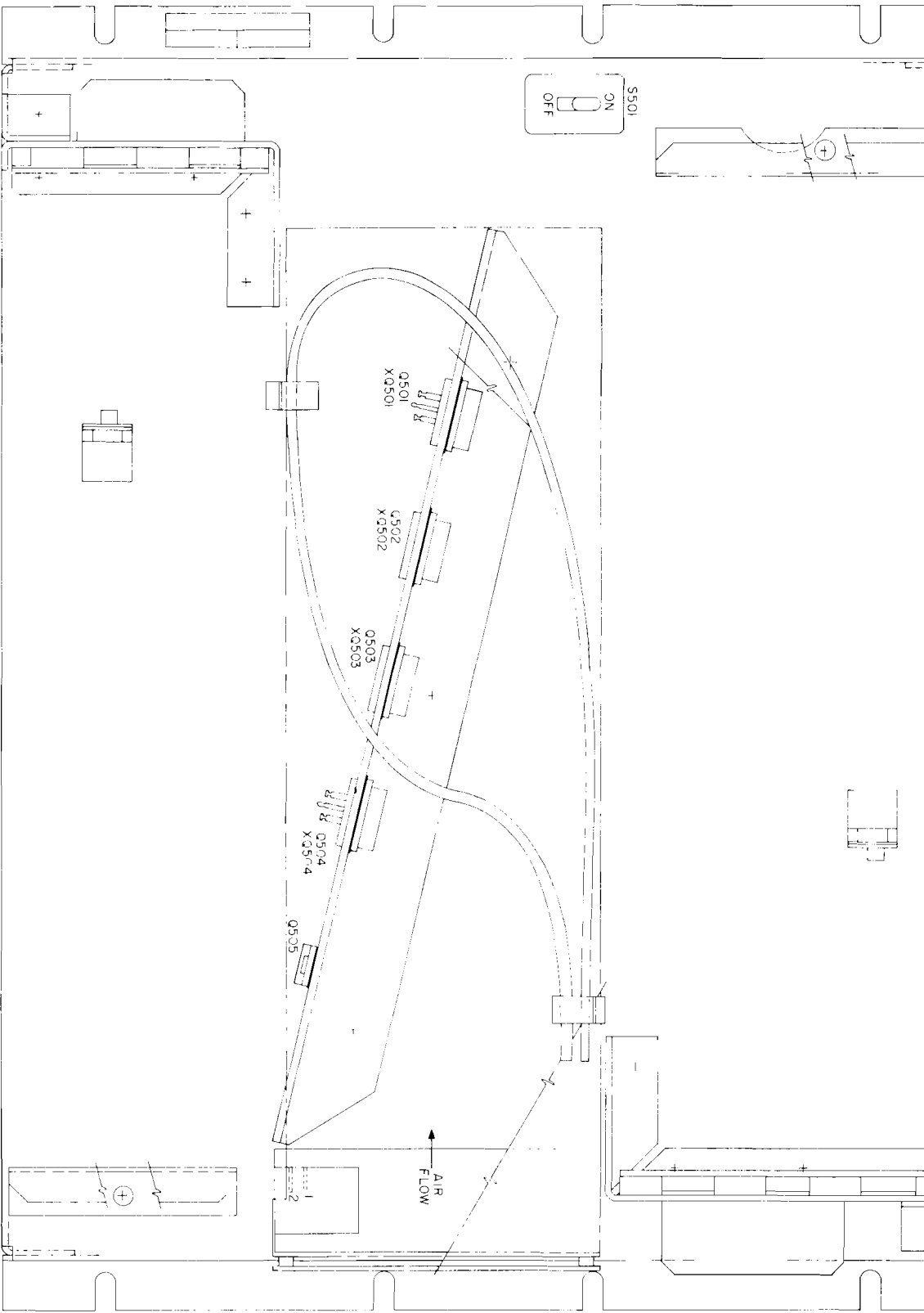
NOTE LEAD ARRANGEMENT, AND NOT CASE SHAPE, IS DETERMINING FACTOR FOR LEAD IDENTIFICATION.



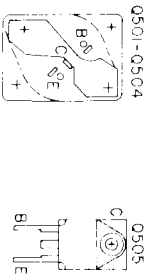
SOLDER SIDE

COMPONENT SIDE

(19R621777, Rev. 7)
(19D416172, Sh. 1, Rev. 7)
(19D416172, Sh. 2, Rev. 5)



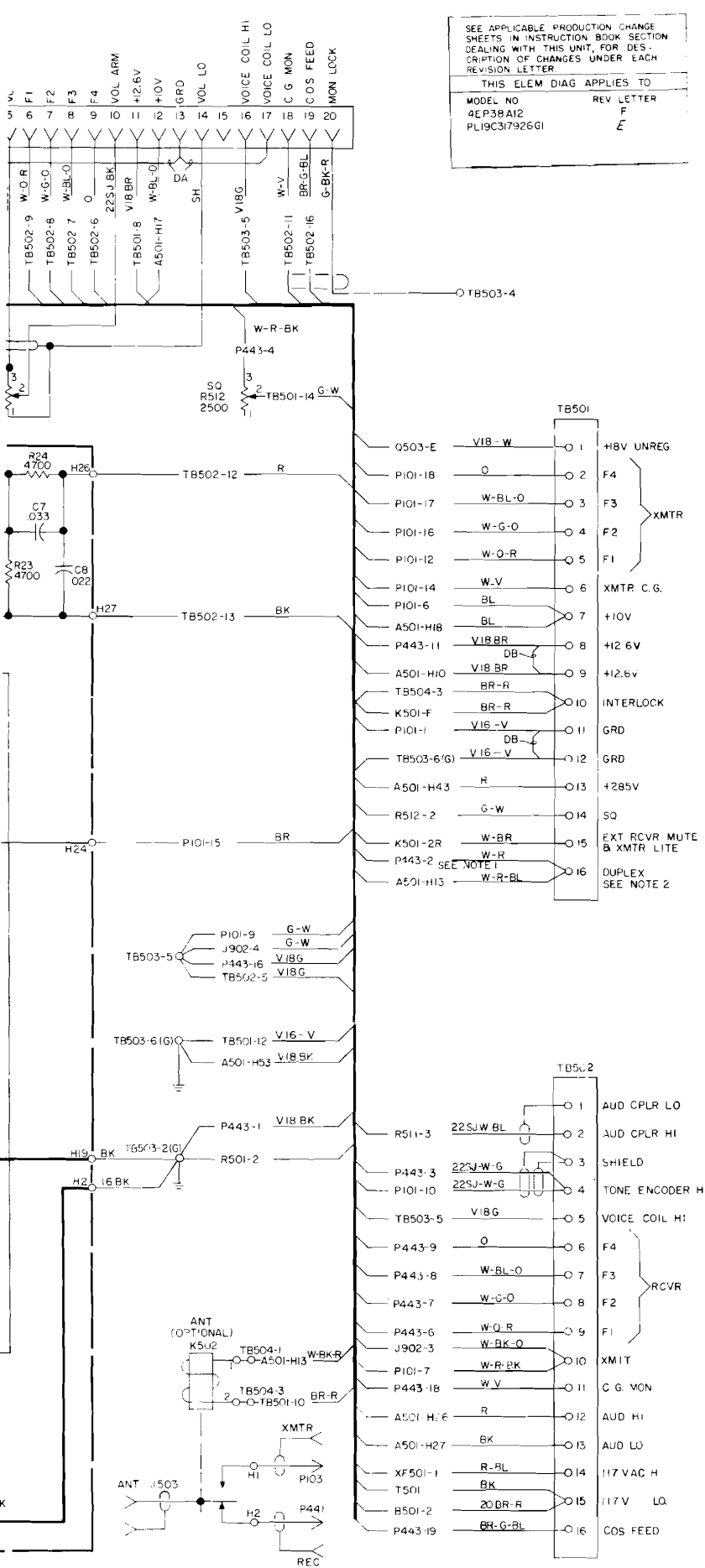
TERMINAL VIEW



— RUNS ON SOLDER SIDE

— RUNS ON BOTH SIDES

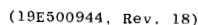
— RUNS ON COMPONENT SIDE



SCHEMATIC DIAGRAM

**TRANSMITTER-RECEIVER POWER SUPPLY
MODEL 4EP38A12**

REPLACEMENT OF ANY
EQUIPMENT SHALL BE
MADE ONLY WITHIN
THE SPECIFICATIONS
LIST FOR THAT PART



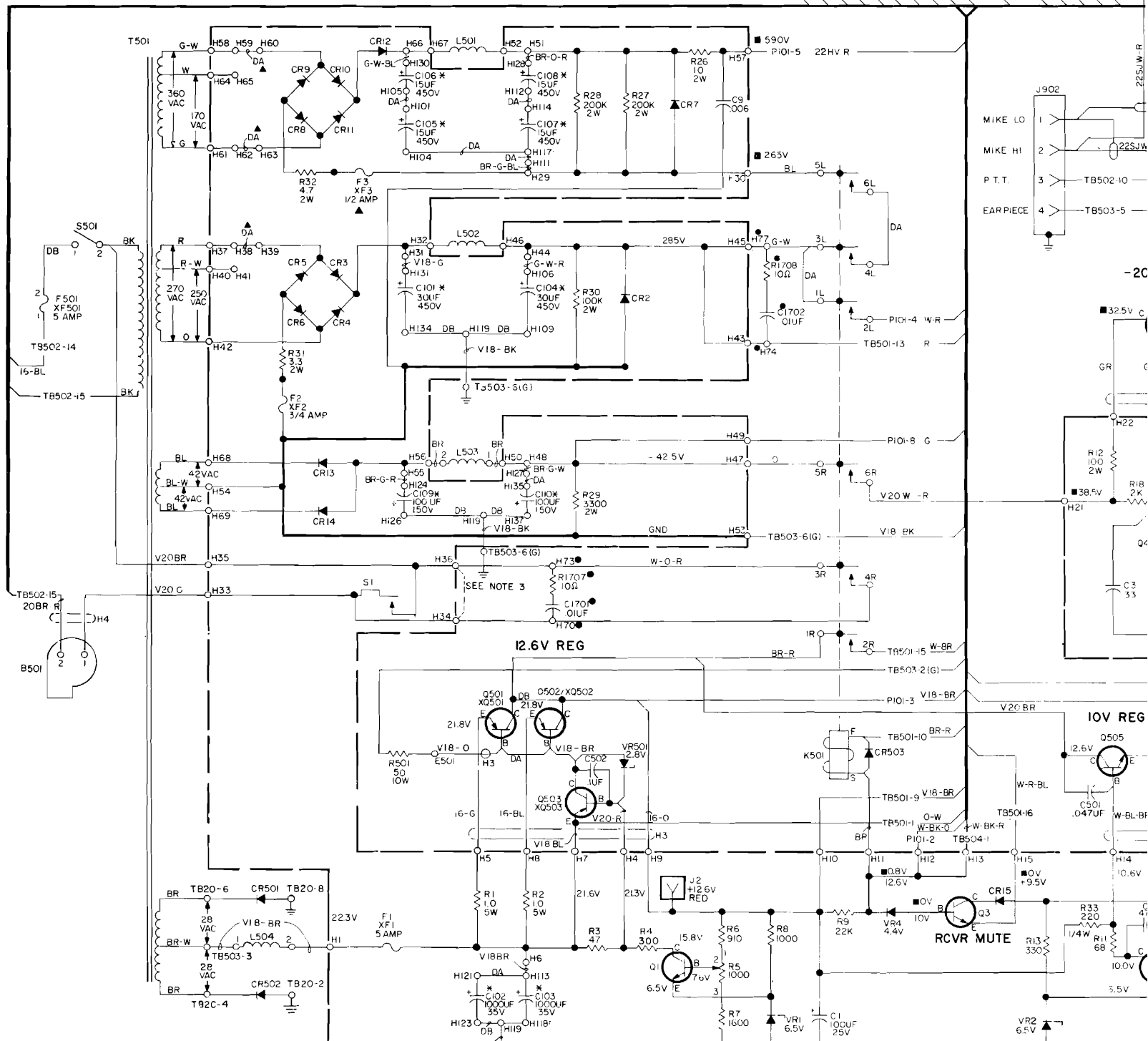
1. TERMINATE W-R WIRE AT TBS01-16 WITH 19B20926P013 TERMINAL. ALLOW SUFFICIENT LENGTH FOR WIRE TO REACH TBS01-7.
2. FOR DUPLEX OPERATION MOVE N24 W-R WIRE FROM TBS01-16 TO TBS01-7.
3. FOR CONTINUOUS BLOWER OPERATION ADD JUMPER FROM A501-H34 TO H36.
4. ALL WIRES ARE N22 UNLESS OTHERWISE SPECIFIED.
5. ALL VOLTAGES ARE TAKEN UNDER LOAD WITH A 20,000 OHMS PER VOLT METER AND ARE IN VDC UNLESS OTHERWISE SPECIFIED. READINGS MARKED WITH ■ ARE TAKEN WITH THE UNIT KEYED. ALL OTHER VOLTAGES ARE IN THE RECEIVE MODE.

ALL RESISTORS ARE 1/2 WATT UNLESS OTHERWISE SPECIFIED AND RESISTOR VALUES IN OHMS UNLESS FOLLOWED BY K=1000 OHMS OR MEG=1,000,000 OHMS. CAPACITOR VALUES IN PICOFARADS (EQUAL TO MICROMICROFARADS) UNLESS FOLLOWED BY UF= MICROFARAOS. INDUCTANCE VALUES IN MICROHENRYS UNLESS FOLLOWED BY MH= MILLIHENRYS OR H=HENRYS

POWER SUPPLY SHOWN CONNECTED FOR 90 WATT
HIGH BAND TRANSMITTER.

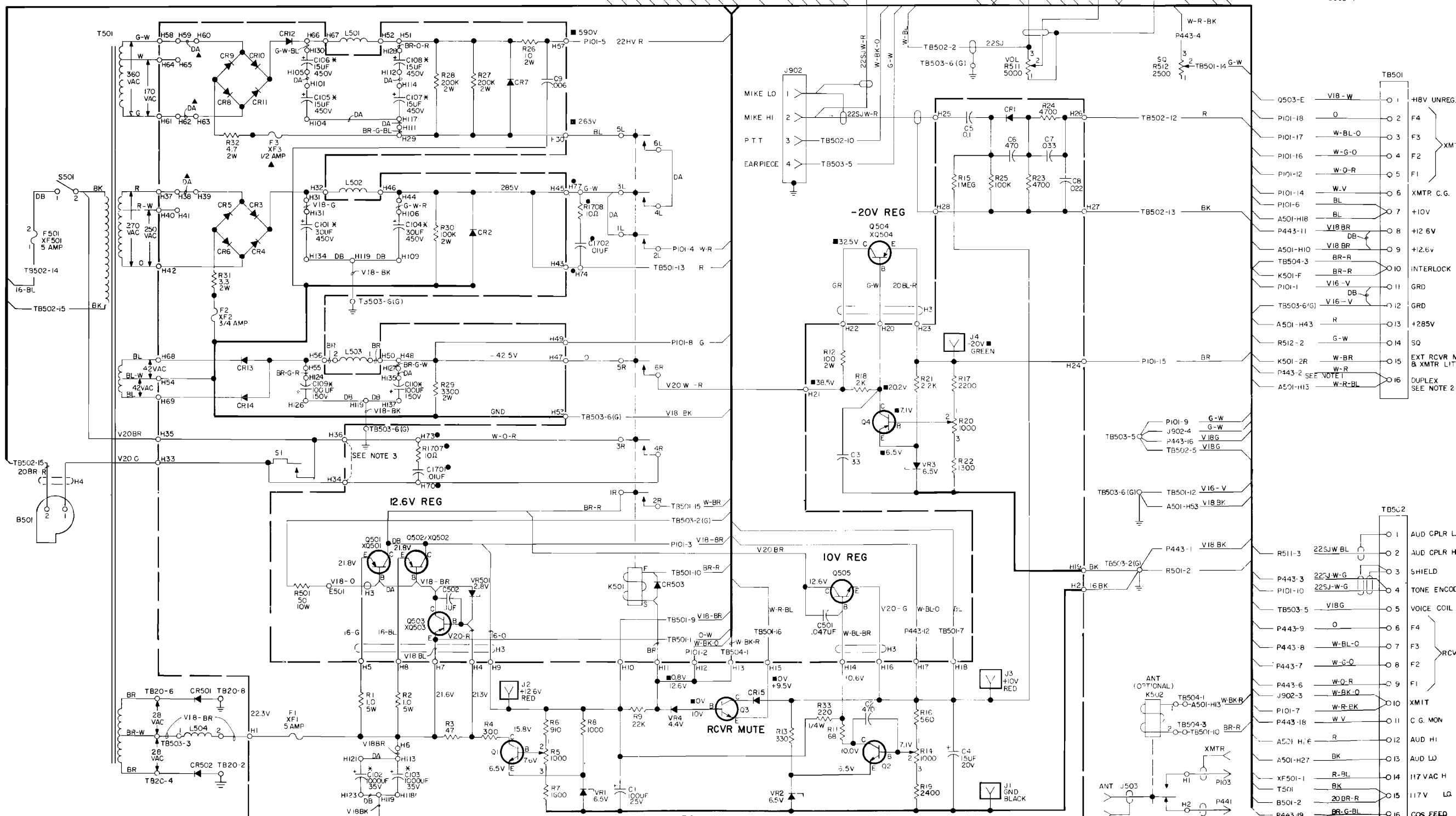
● PART OF MOD KIT PL19A129670G1

IN ORDER TO RETAIN RATED EQUIPMENT PERFORMANCE, REPLACEMENT OF ANY SERVICE PART SHOULD BE MADE ONLY WITH A COMPONENT HAVING THE SPECIFICATIONS SHOWN ON THE PARTS LIST FOR THAT PART.



POWER SUPPLY SHOWN CONNECTED FOR 90 WATT
HIGH BAND TRANSMITTER.

* CAPACITORS MARKED WITH * ARE LOCATED ON BOARD A502.
● PART OF MOD KIT PL19A129670G1



TRANSMITTER/RECEIVER POWER SUPPLY
MODEL 4EP38A12

SYMBOL	GE PART NO.	DESCRIPTION
A501		COMPONENT BOARD 19C317926G1
		----- CAPACITORS -----
C1	19A115680P5	Electrolytic: 100 μ f +150% -10%, 25 VDCW; sim to Mallory Type TT.
C2	7774750P1	Ceramic disc: .00047 μ f +100% -0%, 500 VDCW.
C3	19A115028P117	Polyester: 0.33 μ f \pm 20%, 100 VDCW.
C4	5496267P14	Tantalum: 15 μ f \pm 20%, 20 VDCW; sim to Sprague Type 150D.
C5	19A116080P107	Polyester: 0.1 μ f \pm 10%, 50 VDCW.
C6	5494481P7	Ceramic disc: 470 pf \pm 20%, 1000 VDCW; sim to RMC Type JF Discap.
C7	19A116080P104	Polyester: 0.033 μ f \pm 10%, 50 VDCW.
C8	19A116080P103	Polyester: 0.022 μ f \pm 10%, 50 VDCW.
C9	19C301693P20	Ceramic disc: .006 μ f \pm 10%, 1000 VDCW; sim to RMC Type JF Discap.
		----- DIODES AND RECTIFIERS -----
CR1	19A115050P1	Germanium.
CR2* thru CR12*	4037822P7	Silicon.
		In REV B and earlier:
	4037822P2	Silicon.
CR13 and CR14	4037822P1	Silicon.
CR15*	4037822P1	Silicon. Added by REV E.
		----- INDICATING DEVICES -----
DS1*	4034664P1	Lamp, incandescent: 28 v; sim to GE2148. Deleted by REV A.
		----- FUSES -----
F1	1R16P8	Quick blowing: 5 amps at 250 v; sim to Littelfuse 312005 or Bussmann MTH-5.
F2	1R16P2	Quick blowing: 3/4 amp at 250 v; sim to Littelfuse 312.750 or Bussmann AGC-3/4.
F3	1R16P1	Quick blowing: 1/2 amp at 250 v; sim to Littelfuse 312.500 or Bussmann AGC-1/2.
		----- JACKS AND RECEPTACLES -----
J1	4037265P1	Jack, tip: black plastic body; sim to Component Mfg Service A-1128.
J2 and J3	4037265P2	Jack, tip: red plastic body; sim to Component Mfg Service A-1128.
J4	4037265P3	Jack, tip: green plastic body; sim to Component Mfg Service A-1128.
		----- TRANSISTORS -----
Q1* thru Q3*	19A116755P1	Silicon, NPN; sim to Type 2N3947.
		In REV A and earlier:
	19A115123P1	Silicon, NPN; sim to Type 2N2712.

R1 and R2

R3
R4*

R5

R6

R7

R8

R9

R11

R12

R13

R14

R15

R16

R17

R18

R19

R20

R21

R22

R23 and R24

R25

R26

R27* and R28*

R29

R30

R31

R32

R33*

S1

VR1 thru VR3

VR4

XF1* thru XF3*

5493035P2

3R77P470J

3R77P301J

3R77P431J

19B209358P103

3R77P911J

3R77P162J

3R77P102J

3R77P223K

3R77P680J

3R79P101K

3R77P331J

19B209358P103

3R77P105J

3R77P561J

3R77P222J

3R77P202J

3R77P242J

19B209358P103

3R77P222J

3R77P132J

3R77P472J

3R77P104J

19A116310P23

19A116310P42

3R79P204J

3R79P332J

3R79P104J

19B209022P27

19B209022P31

3R152P221J

19A115687P2

4036887P6

4036887P4

19A116688P1

----- RESISTORS -----

Wirewound: 1 ohm \pm 5%, 5 w; sim to Hamilton Hall Type HR.

Composition: 47 ohms \pm 5%, 1/2 w.

Composition: 300 ohms \pm 5%, 1/2 w.

In REV C and earlier:

Composition: 430 ohms \pm 5%, 1/2 w.

Variable, carbon film: approx 25 to 1000 ohms \pm 10%, 0.2 w; sim to CTS Type X-201.

Composition: 910 ohms \pm 5%, 1/2 w.

Composition: 1600 ohms \pm 5%, 1/2 w.

Composition: 1000 ohms \pm 5%, 1/2 w.

Composition: 22,000 ohms \pm 10%, 1/2 w.

Composition: 68 ohms \pm 5%, 1/2 w.

Composition: 100 ohms \pm 10%, 2 w.

Composition: 330 ohms \pm 5%, 1/2 w.

Variable, carbon film: approx 25 to 1000 ohms \pm 10%, 0.2 w; sim to CTS Type X-201.

Composition: 1.0 megohm \pm 5%, 1/2 w.

Composition: 560 ohms \pm 5%, 1/2 w.

Composition: 2200 ohms \pm 5%, 1/2 w.

Composition: 2000 ohms \pm 5%, 1/2 w.

Composition: 2400 ohms \pm 5%, 1/2 w.

Variable, carbon film: approx 25 to 1000 ohms \pm 10%, 0.2 w; sim to CTS Type X-201.

Composition: 2200 ohms \pm 5%, 1/2 w.

Composition: 1300 ohms \pm 5%, 1/2 w.

Composition: 4700 ohms \pm 5%, 1/2 w.

Composition: 0.10 megohm \pm 5%, 1/2 w.

Composition: 10 ohms \pm 5%, 2 w.

Composition: 0.20 megohm \pm 5%, 1/2 w.

In REV C and earlier:

Composition: 0.20 megohm \pm 5%, 1/2 w.

Composition: 3300 ohms \pm 5%, 1/2 w.

Composition: 0.10 megohm \pm 5%, 1/2 w.

Wirewound: 3.3 ohms \pm 5%, 2 w; sim to IRC Type BWH.

Wirewound: 4.7 ohms \pm 5%, 2 w; sim to IRC Type BWH.

Composition: 220 ohms \pm 5%, 1/4 w. Added by REV A.

----- SWITCHES -----

Thermostat, switch: contacts close at approx 110°F \pm 6°F and open at approx 90°F \pm 5°F.

----- VOLTAGE REGULATORS -----

Silicon, Zener.

Silicon, Zener.

----- SOCKETS -----

Clip, fuse: sim to Littelfuse Inc 102068. (2 used with each socket).

In REV B and earlier:

A502

C101

C102 and C103

C104

C105 and C106

C107 and C108

C109 and C110

B501*

COMPONENT BOARD
19C317935G1

----- CAPACITORS -----

Electrolytic: 30 μ f +50 -10%, 450 VDCW.

Electrolytic: 1000 μ f +250 -15%, 35 VDCW.

Electrolytic: 30 μ f +50 -10%, 450 VDCW.

Electrolytic: 15 μ f +50 -10%, 450 VDCW.

Electrolytic: 15 μ f +50% -10%, 450 VDCW; sim to Mallory Type TC.

Electrolytic: 100 μ f +100% -10%, 150 VDCW; sim to Mallory Type TC.

----- MOTORS -----

Fan, single phase: 115 VAC, 60 Hz; sim to Rotron "Centaur" Muffin Venturi Fan Model CT3A2.

In REV G and earlier:

Fan assembly, single phase: 115 VAC, 60 Hz, 14 w, ccw rotation; sim to Rotron "Gold Seal Venturi Muffin Fan".

----- CAPACITORS -----

Polyester: 0.047 μ f \pm 20%, 50 VDCW.

Polyester: 0.1 μ f \pm 20%, 50 VDCW.

----- DIODES AND RECTIFIERS -----

Silicon.

In REV D and earlier:

Silicon.

Silicon.

----- FUSES -----

Cartridge, medium blowing: 5 amps at 125 v; sim to Bussmann MDX-5.

----- JACKS AND RECEPTACLES -----

Includes:

Connector: 4 female contacts; sim to Amphenol Type 91-PN4F-1000.

Hex nut: 13/16-27N.

Lockwasher, internal tooth, 13/16.

----- RELAYS -----

Armature, open: 12 VDC nominal, 3 w max operating, 6 form A contacts rated at 3 amps at 115 VAC or 28 VDC; sim to Magnecraft 22X636.

----- INDUCTORS -----

Reactor: 5 h ind min at 0.3 amp DC, 80 ohms DC res max, 1000 VDC operating.

Reactor: 3 h ind min at 0.4 amp DC, 30 ohms DC res max, 600 v peak, 300 VDC operating.

Reactor: 800 mh ind min at 0.1 amp DC, 30 ohms DC res max, 100 v peak, 45 VDC operating.

Reactor: 12 mh min, 0.3 ohm DC res max, 18 VDC operating.

----- PLUGS -----

Includes:

Connector, phenolic: 20 contacts rated at 5 amps.

P443

19C303506P1

19A121589G1

Q501 and Q502

19A115268P1

Q503

19A115376P1

Q504

19A115267P1

Q505

19A116203P3

R501

5493035P51

R511

2R76P12

R512

2R76P10

S501

7144140P1

T501

19C307137P1

TB1

7775500P25

TB501 and TB502

19C301086P10

TB503

7775500P23

TB504

7775500P7

VR501

4036887P2

W501*

19C320059G3

19C320059G1

W502

19C320059G2

XF501

19B209005P1

XQ501 thru XQ504

5491888P1

K502

19B204628G1

19C307103P1

19B209044P16

PRODUCTION CHANGES

Changes in the equipment to improve performance or to simplify circuits are identified by a "Revision Letter", which is stamped after the model number of the unit. The revision stamped on the unit includes all previous revisions. Refer to the Parts List for descriptions of parts affected by these revisions.

REV. A - 4EP38A12

To correct connections to Q504.
Moved the G-R wire from Q504-E to Q504-C.
Moved the BL-R wire from Q504-C to Q504-E.

REV. B - To improve regulation.
Changed R4.

REV. C - To incorporate new fuseholders on A501.
Changed XF1, XF2 and XF3.

REV. D - To improve harness.
Changed W501.

REV. E - To incorporate new diodes.
Changed CR501 and CR502.

REV. F - To prevent microphonics with fan
start-up.
Changed Fan. Added grommet and
flatwasher.

REV. A - 10 V Regulator 19C317926G1

To improve reliability of 10 Volt
Regulator.
Deleted DS1 and added R33.

REV. B - 10 V Regulator 19317926G1

To improve reliability of 10 Volt
Regulator. Replace Q1-Q4.

REV. C - 10 V Regulator 19C317926G1

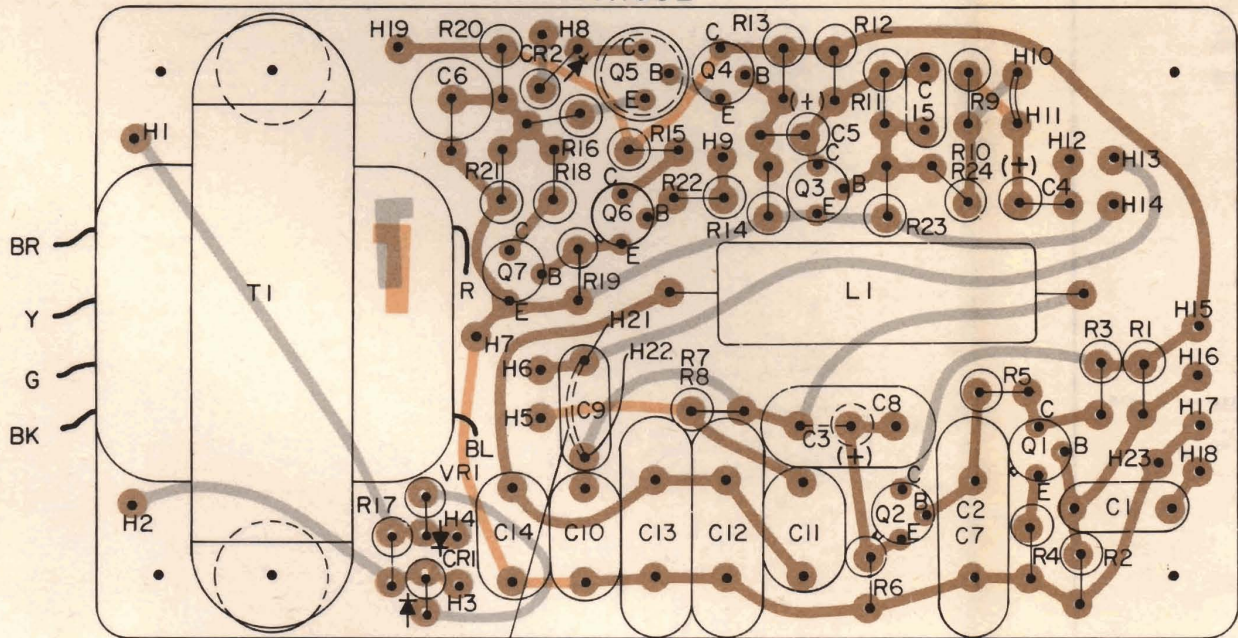
To incorporate diodes with higher
breakdown voltage.
Changed CR2 through CR12.

REV. D - To incorporate improved bleeder resistors
in high voltage circuit.
Changed R27 and R28.

REV. E - To prevent the transmitter from being
keyed by a short on the +10 Volt line.
Added CR15.

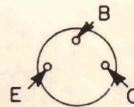
OUTLINE DIAGRAM

A1501
A1502



JUMPER IN
A1501 ONLY

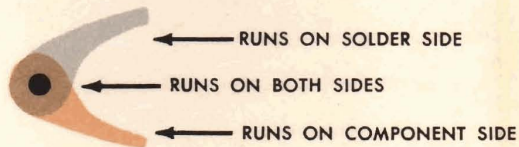
LEAD IDENTIFICATION
FOR Q1, Q2, Q5 & Q6



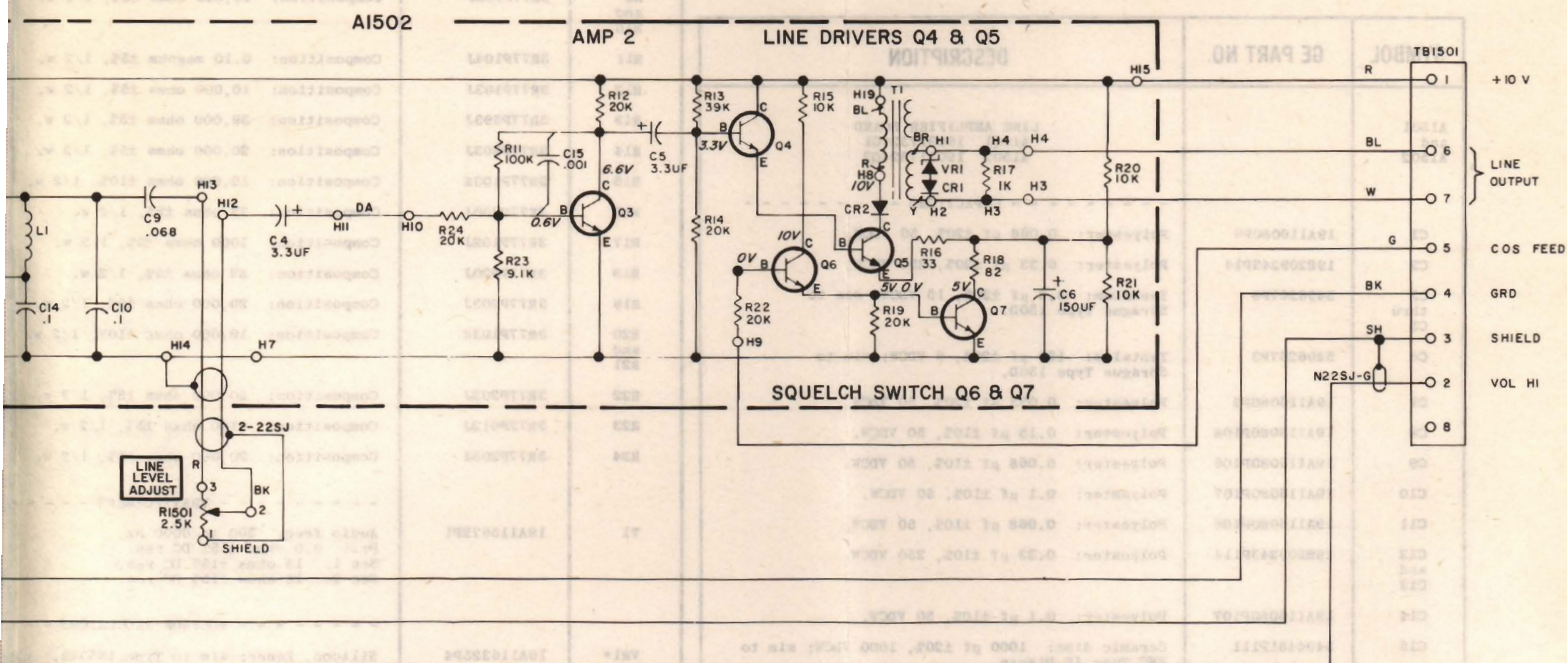
TRIANGULAR
VIEW FROM LEAD END

NOTE: LEAD ARRANGEMENT, AND NOT
CASE SHAPE, IS DETERMINING
FACTOR FOR LEAD IDENTIFICATION.

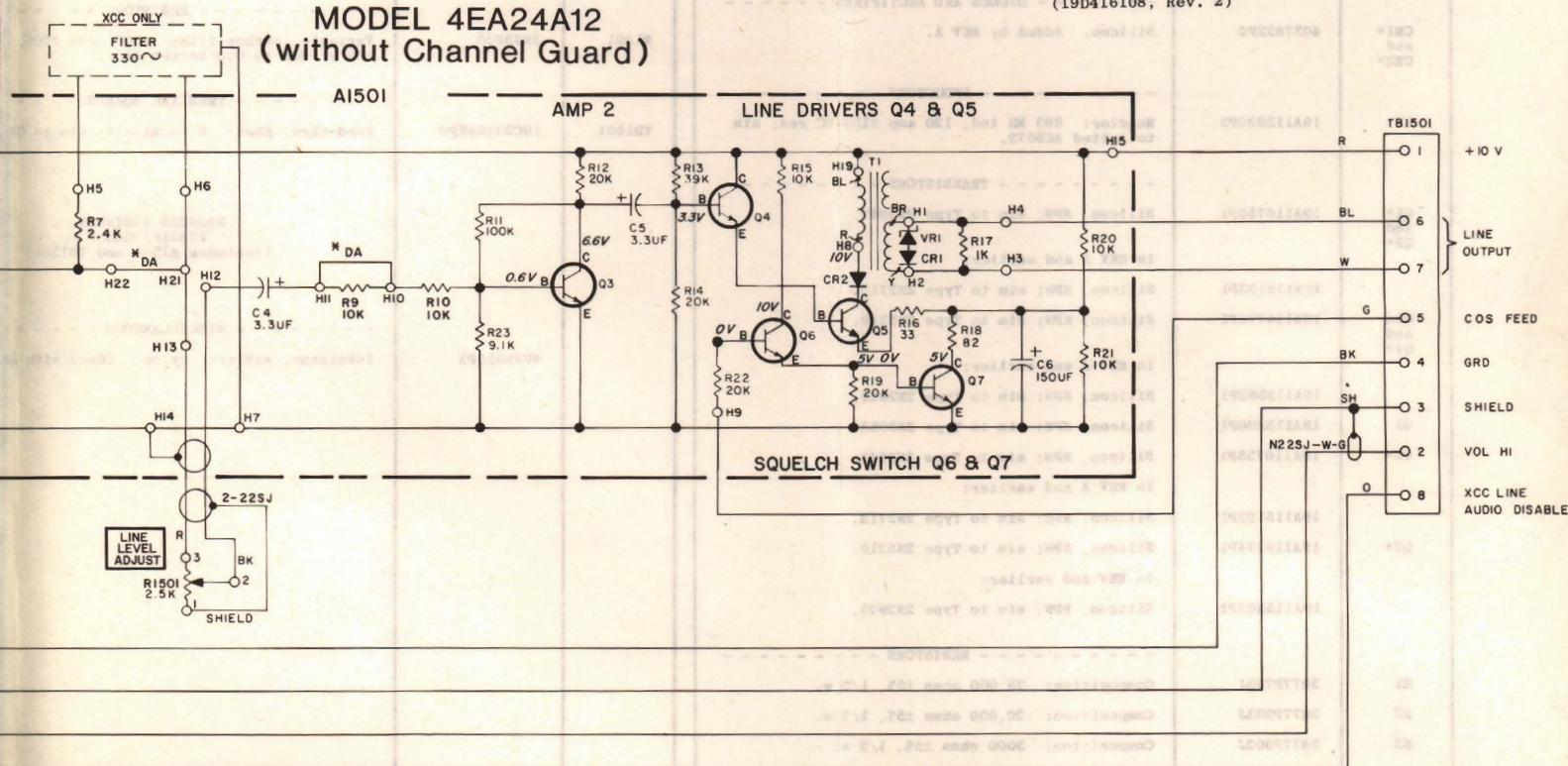
(19C320095, Rev. 2)
(19B216797, Sh. 1, Rev. 1)
(19B216797, Sh. 2, Rev. 1)



SCHEMATIC DIAGRAM

MODEL 4EA24A13
(with Channel Guard)

(19D416108, Rev. 2)

MODEL 4EA24A12
(without Channel Guard)

(19D416107, Rev. 2)

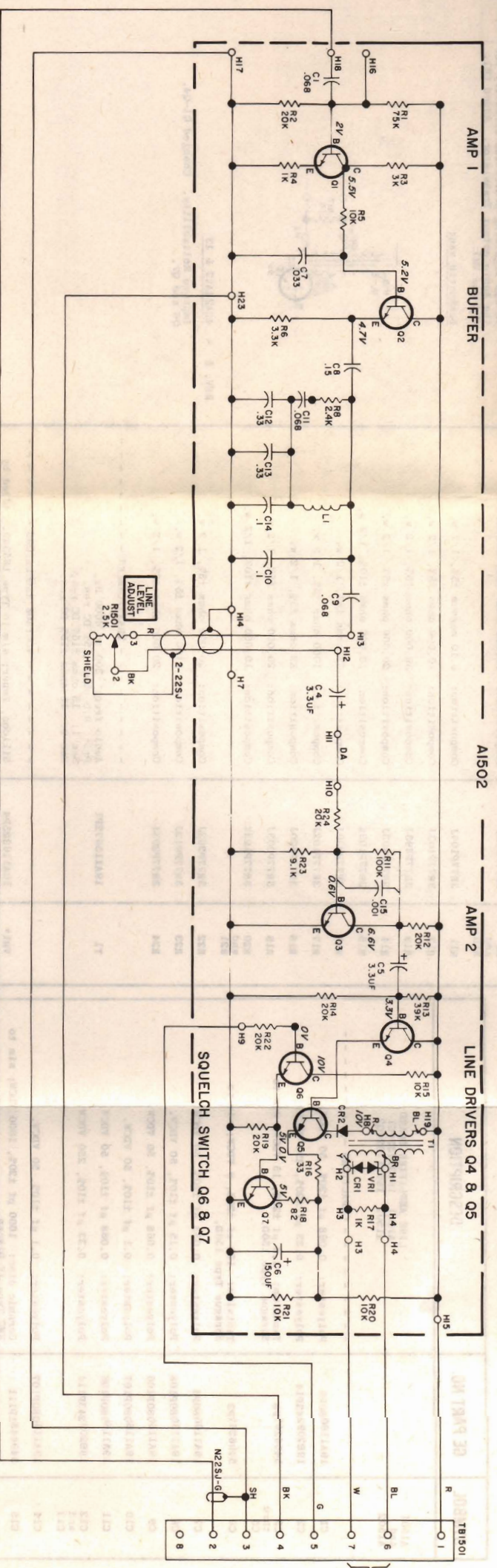
SCHEMATIC & OUTLINE DIAGRAM

LINE AMPLIFIER MODELS 4EA24A12, 13

NOTE:

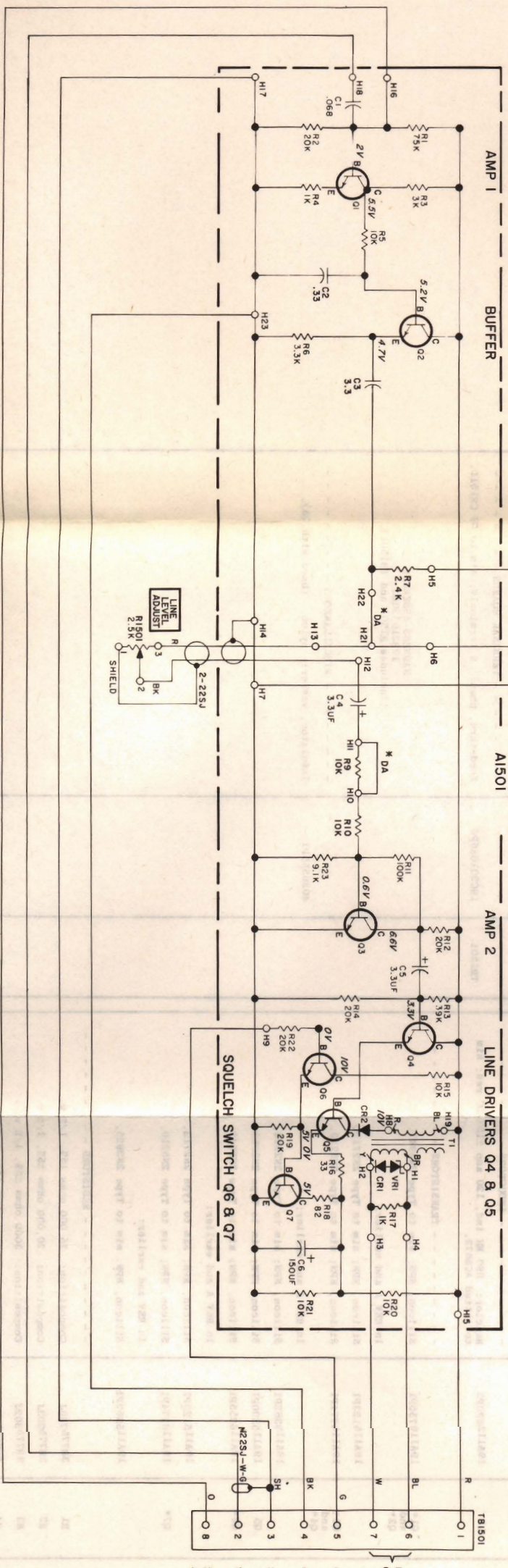
- * JUMPERS REMOVED IN XCC APPLICATION.
- ALL WIRE IS N22 UNLESS OTHERWISE SPECIFIED

IN ORDER TO RETAIN RATED EQUIPMENT PERFORMANCE, REPLACEMENT OF ANY SERVICE PART SHOULD BE MADE ONLY WITH A COMPONENT HAVING THE SPECIFICATIONS SHOWN ON THE PARTS LIST FOR THAT PART.



MODEL 4EA24A12
(without Channel Guard)

(19DA16108, Rev. 2)



VOLTAGE READINGS

THESE READINGS ARE TYPICAL DC
READINGS MEASURED FROM TRANSISTOR
PINS TO GROUND WITH A 20,000 OHM-
PER-VOLT METER.

ALL RESISTORS ARE 1/2 WATT UNLESS
OTHERWISE SPECIFIED AND RESISTOR
VALUES IN OHMS UNLESS FOLLOWED BY
A UNIT DESIGNATION.

IN ORDER TO RETAIN RATED EQUIPMENT

(19DA16107, Rev. 2)

SCHEMATIC & OUTLINE DIAGRAM

PARTS LIST

LBI-4322B
 LINE AMPLIFIER
 MODEL 4EA24A12 (19A129097G1)
 MODEL 4EA24A13 (19A129097G2)

SYMBOL	GE PART NO.	DESCRIPTION
A1501 and A1502		LINE AMPLIFIER BOARD A1501 19C317324G1 A1502 19C317324G2
		----- CAPACITORS -----
C1	19A116080P6	Polyester: 0.068 μ f \pm 20%, 50 VDCW.
C2	19B209243P14	Polyester: 0.33 μ f \pm 20%, 250 VDCW.
C3 thru C5	5496267P9	Tantalum: 3.3 μ f \pm 20%, 15 VDCW; sim to Sprague Type 150D.
C6	5496267P3	Tantalum: 150 μ f \pm 20%, 6 VDCW; sim to Sprague Type 150D.
C7	19A116080P4	Polyester: 0.033 μ f \pm 20%, 50 VDCW.
C8	19A116080P108	Polyester: 0.15 μ f \pm 10%, 50 VDCW.
C9	19A116080P106	Polyester: 0.068 μ f \pm 10%, 50 VDCW.
C10	19A116080P107	Polyester: 0.1 μ f \pm 10%, 50 VDCW.
C11	19A116080P106	Polyester: 0.068 μ f \pm 10%, 50 VDCW.
C12 and C13	19B209243P114	Polyester: 0.33 μ f \pm 10%, 250 VDCW.
C14	19A116080P107	Polyester: 0.1 μ f \pm 10%, 50 VDCW.
C15	5494481P111	Ceramic disc: 1000 pf \pm 20%, 1000 VDCW; sim to RMC Type JF Discap.
		----- DIODES AND RECTIFIERS -----
CR1* and CR2*	4037822P2	Silicon. Added by REV A.
		----- INDUCTORS -----
L1	19A115690P3	Reactor: 880 MH ind, 120 amp \pm 15% DC res; sim to Arttd AC5672.
		----- TRANSISTORS -----
Q1* and Q2*	19A116755P1	Silicon, NPN; sim to Type 2N3947. In REV A and earlier:
	19A115123P1	Silicon, NPN; sim to Type 2N2712.
Q3* and Q4*	19A116774P1	Silicon, NPN; sim to Type 2N5210. In REV A and earlier:
	19A115362P1	Silicon, NPN; sim to Type 2N2925.
Q5	19A115300P1	Silicon, NPN; sim to Type 2N3053.
Q6*	19A116755P1	Silicon, NPN; sim to Type 2N3947. In REV A and earlier:
	19A115123P1	Silicon, NPN; sim to Type 2N2712.
Q7*	19A116774P1	Silicon, NPN; sim to Type 2N5210. In REV and earlier:
	19A115362P1	Silicon, NPN; sim to Type 2N2925.
		----- RESISTORS -----
R1	3R77P753J	Composition: 75,000 ohms \pm 5%, 1/2 w.
R2	3R77P203J	Composition: 20,000 ohms \pm 5%, 1/2 w.
R3	3R77P302J	Composition: 3000 ohms \pm 5%, 1/2 w.
R4	3R77P102J	Composition: 1000 ohms \pm 5%, 1/2 w.
R5	3R77P103K	Composition: 10,000 ohms \pm 10%, 1/2 w.

SYMBOL	G-E PART NO	DESCRIPTION
R6	3R77P332K	Composition: 3300 ohms \pm 10%, 1/2 w.
R7 and R8	3R77P242J	Composition: 2400 ohms \pm 5%, 1/2 w.
R9 and R10	3R77P103J	Composition: 10,000 ohms \pm 5%, 1/2 w.
R11	3R77P104J	Composition: 0.10 megohm \pm 5%, 1/2 w.
R12	3R77P103J	Composition: 10,000 ohms \pm 5%, 1/2 w.
R13	3R77P393J	Composition: 39,000 ohms \pm 5%, 1/2 w.
R14	3R77P203J	Composition: 20,000 ohms \pm 5%, 1/2 w.
R15	3R77P103K	Composition: 10,000 ohms \pm 10%, 1/2 w.
R16	3R77P330J	Composition: 33 ohms \pm 5%, 1/2 w.
R17	3R77P102J	Composition: 1000 ohms \pm 5%, 1/2 w.
R18	3R77P820J	Composition: 82 ohms \pm 5%, 1/2 w.
R19	3R77P203J	Composition: 20,000 ohms \pm 5%, 1/2 w.
R20 and R21	3R77P103K	Composition: 10,000 ohms \pm 10%, 1/2 w.
R22	3R77P203J	Composition: 20,000 ohms \pm 5%, 1/2 w.
R23	3R77P912J	Composition: 9100 ohms \pm 5%, 1/2 w.
R24	3R77P203J	Composition: 20,000 ohms \pm 5%, 1/2 w.
		----- TRANSFORMERS -----
T1	19A115672P1	Audio freq: 300 to 6000 Hz, Pri: 9.0 ohms \pm 15% DC res, Sec 1: 15 ohms \pm 15% DC res, Sec 2: 15 ohms \pm 15% DC res.
		----- VOLTAGE REGULATORS -----
VR1*	19A116325P4	Silicon, Zener; sim to Type 1N5349. REV A.
		----- RESISTORS -----
R1501	2R75P50	Variable, carbon film: 2500 ohms \pm 20% 1/2 w; sim to CTS Series 45.
		----- TERMINAL BOARDS -----
TB1501	19C301086P6	Feed-thru, phen: 8 terminals; sim to
		HARNESS ASSEMBLY 19B219320G3 (Includes R1501 and TB1501)
		----- MISCELLANEOUS -----
	4036555P1	Insulator, washer: nylon. (Used with

*COMPONENTS ADDED, DELETED OR CHANGED BY PRODUCTION CHANGES

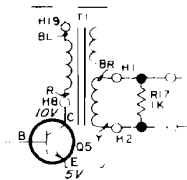
PRODUCTION CHANGES

Changes in the equipment to improve performance or to simplify circuits are identified by a "Revision Letter", which is stamped after the model number of the unit. The revision stamped on the unit includes all previous revisions. Refer to the Parts List for descriptions of parts affected by these revisions.

REV. A - 4EA24A12 & 13

To add lightning protection circuitry for the Audio output Transistor. Added CR1, CR2 and VR1.

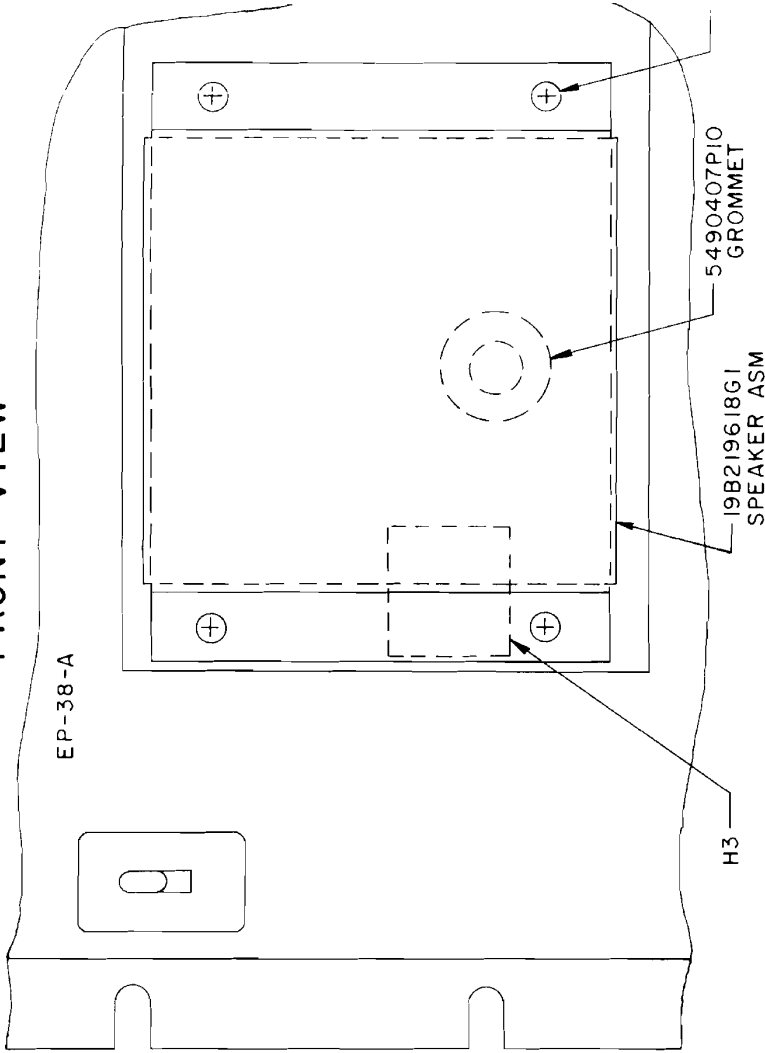
Schematic was:



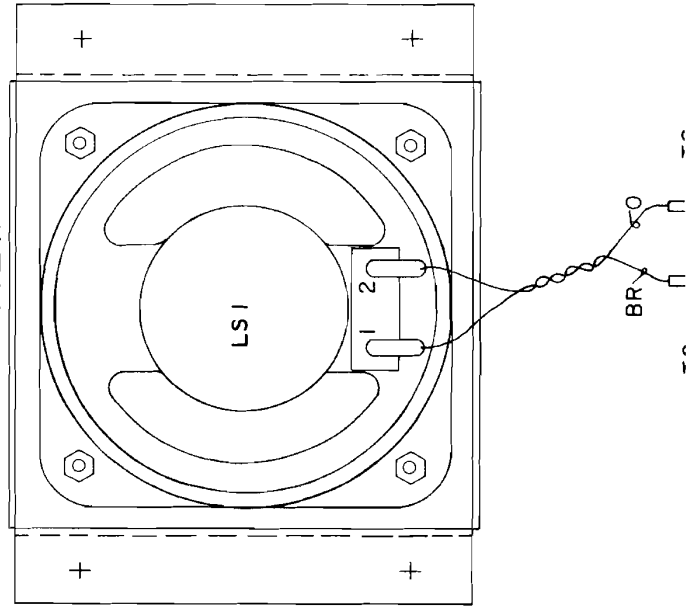
REV. B - 4EA24A12 & 13

Improve Reliability. Changed Q1-Q4, Q6 and Q7.

FRONT VIEW



REAR VIEW



SYMBOL	GE PART NO.	DESCRIPTION
LS1	19A115964P1	Weatherproof, Permanent Magnet: 3-1/2 inch, 18 ohm $\pm 10\%$ imp at 1000 Hz, 15-19 ohms DC; sim to Oaktron S-9847.
	19B219615P1	Cover.
	19B209260P103	Terminal, solderless: sim to AMP 60495-1.
	5490407P10	Grommet.
	19B201074P304	Tap screw: No. 6-32 x 1/4.

ORDERING SERVICE PARTS

Each component appearing on the schematic diagram is identified by a symbol number to simplify locating it in the parts list. Each component is listed by symbol number, followed by its description and GE Part Number.

Service parts may be obtained from Authorized GE Communication Equipment Service Stations or through any GE Radio Communication Equipment Sales Office. When ordering a part, be sure to give:

1. GE Part Number for component
2. Description of part
3. Model number of equipment
4. Revision letter stamped on unit

These instructions do not purport to cover all details or variations in equipment nor to provide for every possible contingency to be met in connection with installation, operation or maintenance.

Should further information be desired, or should particular problems arise which are not covered sufficiently for the purchaser's purposes, contact the nearest Radio Communication Equipment Sales Office of the General Electric Company.

MAINTENANCE MANUAL

LBI-4323

ORIGINAL SOURCE PARTS

DF-0060

MOBILE RADIO DEPARTMENT
GENERAL ELECTRIC COMPANY • LYNCHBURG, VIRGINIA 24502

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