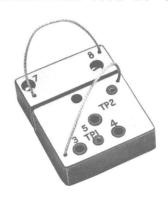
# MASTR Personal Series PROGRESS LINE

PE MODELS

132-174 MHz, 100 to 200 MILLIWATTS TRANSMITTER TYPE ET-95-A 132-174 MHz, 1 to 2 WATT TRANSMITTER TYPE ET-96-A



### **SPECIFICATIONS** \*

Type Number

Power Output ET-95-A ET-96-A

Modulation Deviation

Spurious Radiated Conducted

Audio Response

Audio Distortion

Crystal Multiplication Factor

RF Load Impedance

Modulation Sensitivity

ET-95-A & ET-96-A

100 to 200 milliwatts 1 to 2 Watts

0 to  $\pm 5$  kHz

Within +1 and -3 dB of a 6-dB/octave pre-emphasis from 300 to 3000 Hz except for an additional 6-dB/octave roll-off from 2500 to 3000 Hz per EIA.

Less than 8%

12

50 ohms

0.7 to 1.5 Millivolts

hese specifications are intended primarily for the use of the serviceman. Refer to the appropriate Specification Sheet for the complete specifications.

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### ---WARNING ----

No one should be permitted to handle any portion of the equipment that is supplied with high voltage; or to connect any external apparatus to the units while the units are supplied with power. KEEP AWAY FROM LIVE CIRCUITS.

### **DESCRIPTION**

Transmitter Types ET-95-A and ET-96-A are crystal controlled, phase modulated transmitters for one- or two-frequency operation in the 132-174 MHz band. The transmitters utilize both discrete components and Integrated Circuit Modules (IC's).

The transmitters consist of the audio, regulator oscillator, compensator and modulator IC's, and plug-in Exciter and PA modules.

Transmitter Type ET-95-A uses a matching network as the final stage to provide an RF output of 200 milliwatts. Transmitter Type ET-96-A uses a transistorized PA module to provide an RF power output of over two watts. All of the transmitter modules are mounted on System Board A701. Supply voltages for the transmitter are provided by the battery and Regulator. The different transmitter voltages are shown in the following chart:

Voltage	Used For:
Continuous 7.5 Volts	Regulator module
Keyed 7.5 Volts	Regulator 5.4-Volt keying, Exciter and PA modules.
Keyed 5.4 Volts	Compensator, Oscilla- tor, Audio and Modu- lator modules, and optional Compressor module.

References to symbol numbers mentioned in the following text are found on the Schematic Diagrams, Outline Diagrams

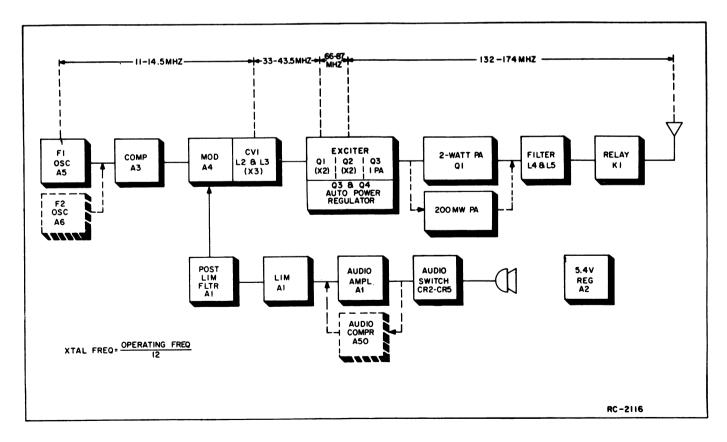


Figure 1 - Transmitter Block Diagram

and Parts Lists (see Table of Contents). The typical, simplified circuit diagrams used in the text are representative of the circuits in the IC modules. A block diagram of the transmitter is shown in Figure 1.

### **CIRCUIT ANALYSIS**

#### REGULATOR A2

The Regulator module operates from the 7.5-volt from the battery, and provides a continuous, regulated 5.4 volts and a switched 5.4 volts for operating the transmitter, receiver and tone options. A typical regulator circuit is shown in Figure 2.

Turning on the radio applies the battery voltage to Pin 2 of the Regulator, causing Q2 and then Q1 to conduct. When conducting, the continuous 5.4 volts at the collector of Q1 is taken from Pin 4 and applied to the receiver Compensator and Oscillator module.

Regulation is provided by Q2 and Q3, which operate as a differential amplifier. If the output of Q1 starts to increase, Q3 conducts harder, cuasing Q2 to conduct less. This causes Q1 to conduct less, keeping its output at 5.4 Volts. If the output of Q1 starts to decrease, Q3 conducts less, causing Q2 to conduct harder. This causes Q1 to conduct harder, keeping the output constant.

Q4 and Q5 operate as a DC switch. Keying the transmitter applies the battery voltage to Pin 7 and to the base of Q5, turning it on. This turns on PNP transistor Q4, so that the regulated 5.4 volts at Pin 6 is applied to the transmitter Compensator, Modulator, and audio module, and to the optional Compressor module and multi-frequency switch S1.

#### OSCILLATOR MODULE A5

Oscillator Model 4EG27A10 consists of a crystal-controlled Colpitts oscillator and a Channel Guard tone modulator. The entire oscillator is contained in a metal can with the transmitter operating frequency printed on the top. The crystal frequency ranges from 11 to 14.5 MHz, and the crystal frequency is multiplied 12 times.

The oscillator frequency is temperature compensated to provide instant frequency compensation with a frequency stability of  $\pm .0002\%$  from 0°C to +55°C and  $\pm .0005\%$  from -30°C to +60°C. The temperature compensation network is contained in Compensator module A3.

A typical oscillator circuit is shown in Figure 3.

In single-frequency transmitters, a jumper from Hole 20 to Hole 21 on the System Board connects the keyed 5.4 volt supply voltage to the oscillator module. Keying the transmitter applies the supply voltage to the oscillator, turning it on. The oscillator output is applied to Compensator A3.

In two-frequency transmitters, F2 Oscillator Module A6 is mounted on the board. The single-frequency supply jumper is removed, and the proper frequency is selected by connecting the keyed 5.4 volts to the selected oscillator module through frequency selector switch S1 on the control unit.

For Channel Guard applications, tone from the Channel Guard encoder is applied to the oscillator module. The tone is applied through Pin 3 to the voltage-variable capacitor on the oscillator module, which frequency modulates the oscillator output.

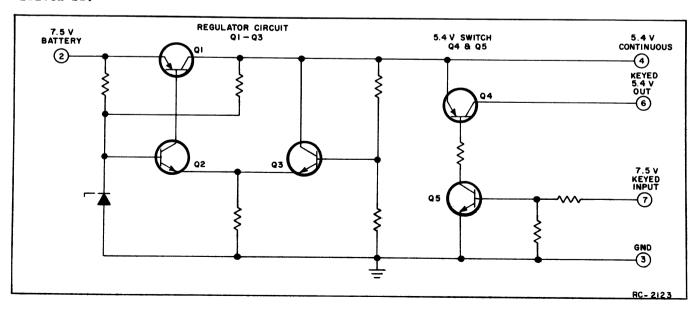


Figure 2 - Typical Regulator Circuit

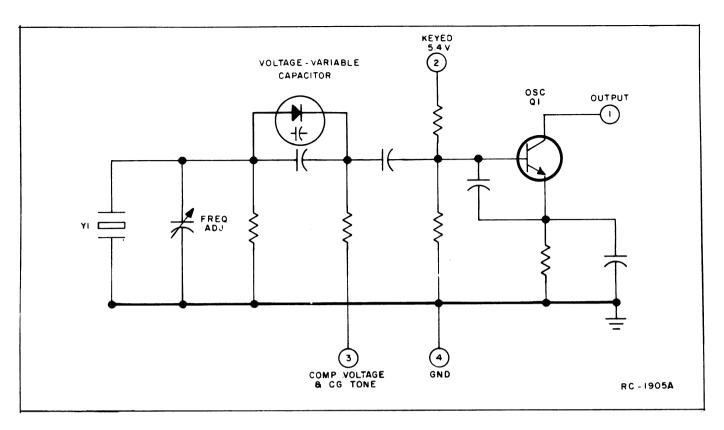


Figure 3 - Typical Oscillator Circuit

- NOTE -

All oscillator modules are individually compensated at the factory and cannot be repaired in the field. Any attempt to remove the oscillator cover will void the warranty.

### COMPENSATOR A3

Compensator module A3 contains a buffer-amplifier, and the temperature compensating network for the oscillator. A typical Compensator circuit is shown in Figure 4.

RF from the oscillator at Pin 7 is coupled through a DC-blocking capacitor to the base of buffer-amplifier Q1. This stage isolates the oscillator from the modulator. The output of Q1 connects from Pin 9 to the modulator.

In the compensation network, the keyed 5.4 volts at Pin 2 is applied to a thermistor-compensated voltage divider. The output at Pin 3 (2.35 volts measured with a VTVM) is applied to Pin 3 and to the voltage-variable capacitor in the oscillator module. At temperatures below -10°C, the compensated voltage increases to maintain the proper voltage on the oscillator voltage-variable capacitor.

### - SERVICE NOTE -

An abnormally low VTVM reading (or no reading) at Pin 3 may indicate a short or leakage path in the oscillator. This can be checked by unsoldering Pin 3, raising it off the printed board and taking another reading. If this reading is normal the problem is in the oscillator module. If the reading remains low (or zero) the problem is in the Compensator.

### AUDIO AMPLIFIER AND LIMITER A1

Audio from the microphone is coupled through the audio switching circuit to Pin 1 and then to the base of audio amplifier Q1 (see Figure 5). In Type 90 encoder applications, the encode tone is applied to the amplifier at Pin 2.

The amplifier output is applied directly to the transistorized limiter stage (Q2). Following the limiter is a combined postlimiter filter and de-emphasis network. The filter output at Pin 8 is coupled through Mod Adjust potentiometer R8 to the Modulator module A4.

When the Audio Compressor option is used, audio from the microphone is coupled

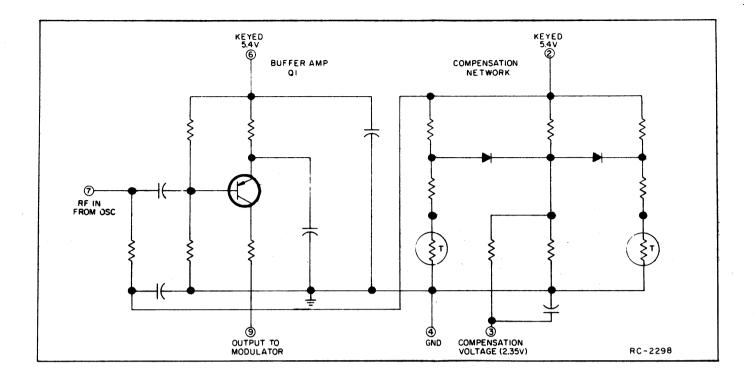


Figure 4 - Typical Compensator Circuit

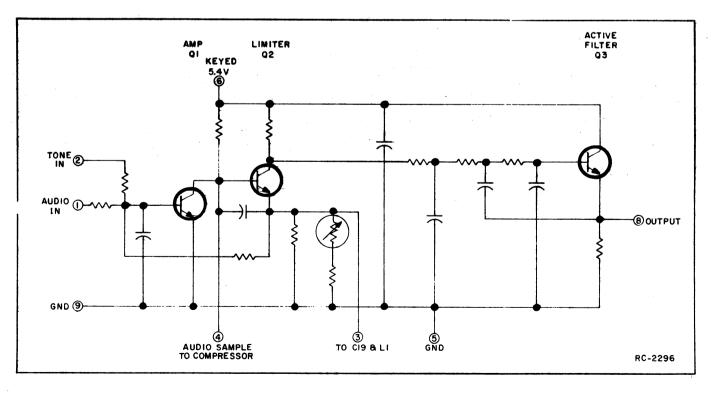


Figure 5 - Typical Audio Amplifier & Limiter Circuit

through the compressor and then applied to the audio amplifier stage. An audio sample from the collector of amplifier Ql is connected from Pin 4 to the compressor circuit, keeping the audio output to the modulator constant.

#### AUDIO COMPRESSOR A50

The optional Audio Compressor Module provides a relatively constant audio output to the Audio Amplifier-Limiter module over a 30-dB change in input level. The compressor module also provides 13-dB additional gain for increased microphone sensitivity. A typical diagram of the Compressor is shown in Figure 6.

Audio from the microphone is coupled through R52 on the System Board to Pin 1 of the Compressor. The audio is applied to pre-amplifier Q1 which provides the 13 dB gain. The pre-amplifier output at Pin 4 is connected to Pin 1 of Amplifier-Limiter module A1.

At the same time, an audio sample voltage from Audio module Al is applied to Pin 9 and to audio amplifier Q3 in the Compressor module. The output of Q3 is rectified by the two diodes, and the resultant voltage applied to the base of DC amplifier Q4. The DC output of Q4 controls the operation of the compressor-control transistor Q2.

An increase in the audio sample voltage increases the DC voltage applied to Q2. This reduces the AC impedance of Q2, which decreases the audio output voltage at Pin 4. A decrease in the audio sample voltage decreases the DC voltage applied to Q2. This increases the AC impedance of Q2, and increases the audio output voltage at Pin 4.

#### PHASE MODULATOR

The phase modulator circuit consists of Modulator A4, voltage-variable capacitor CV1 and tuneable coil L2. CV1 and L2 are mounted on System Board A701. A typical modulator circuit is shown in Figure 7.

With CV1 in series with L2, the network appears as a series-resonant circuit when RF from the oscillator is applied to Pin 1. Applying audio from Audio Limiter Al to Pin 4 of Modulator A4 varies the bias of CV1, resulting in a phase modulated output.

Buffer Ql isolates the modulator from the loading effects of the following multiplier stage, and also provides some amplification. Following the buffer stage is tripler Q2. The output of Q2 is coupled through L3 (on the System Board) to the exciter module. L3 is tuned to three times the crystal frequency.

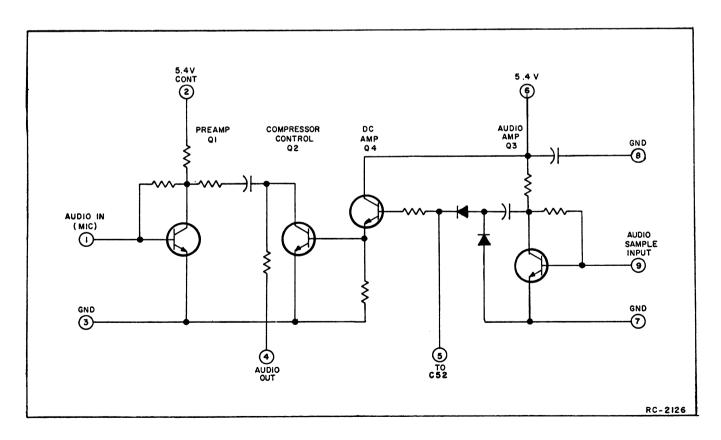


Figure 6 - Typical Audio Compressor Circuit

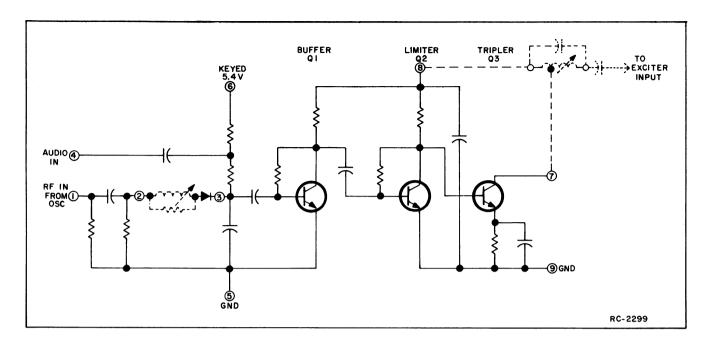


Figure 7 - Typical Phase Modulator Circuit

#### **EXCITER**

Exciter Board Models 4EG29A10 (132-174 MHz) and 4EG29A11 (150.8 - 174 MHz) consist of two class C doubler stages; a class C amplifier stage, and an Automatic Power Level Control (APLC) circuit.

### Doubler & Amplifier Stages

The modulator output is coupled through T1 to the base of 1st doubler Q1. The 1st doubler stage as well as the modulator stage is metered at TP1. The 1st doubler output is coupled through T2 to the base of 2nd doubler Q2. T2 is tuned to six times the crystal frequency.

Following the 2nd doubler is an impedance-matching network consisting of C14, C16, C17, C19 and L2. The network matches the high impedance 2nd doubler output to the low impedance amplifier input. L2 is tuned to 12 times the crystal frequency.

2nd doubler Q2, amplifier Q3 and the PA transistor are tuned by measuring the total current drain of the radio. An ammeter with a one ampere full scale meter is used in series with the radio 7.5-Volt supply. GE Test Regulator Model 4EX18A10 and Test Set Model 4EX3A11 may be used in place of the ammeter.

A constant-K, DC collector feed network consisting of L1, L7, C4 and C12 provides improved 2nd doubler stability. Similar collector-feed networks are used in the amplifier and PA stages.

The output of amplifier Q3 is applied to the PA module.

#### APLC Circuit

The APLC circuit (Q4 and Q5) provides a more constant transmitter power output by controlling the output of the 1st and 2nd doubler. The circuit also extends the battery life by regulating the current to amplifier Q3.

When Q3 starts to conduct harder and draw more collector current, the voltage drop across R7 increases, causing Q4 to conduct harder. This increases the voltage at the base of Q5. Increasing the voltage at the base of Q5 causes it to cinduct less, which increases the voltage drop across Q5 and reduces the collector voltage of Q1 and Q2. This reduces the drive to amplifier Q3 and reduces the collector current.

In low power transmitters, Power Adjust Potentiometer R8 is used to set the power output. In 2-Watt transmitters, R8 can be used to limit the maximum power output.

#### PA MODULES

Two plug-in PA modules are available for use in the transmitter, depending on the power output required.

In one- to two-watt transmitters, PA module 4EF39A10 (132 to 150.8 MHz) or 4EF39A11 (150.8 to 174 MHz) is used. The output of the exciter is coupled through a tuned circuit to the base of class C amplifier Ql. The amplifier output is applied through a series-tuned circuit to the low-pass filter.

In 100 to 200 milliwatt transmitters, the PA consists of a "T"-Type matching network. The output of the network is applied to the low-pass filter. An RF adaptor cable is available for connecting the transmitter RF output to a wattmeter. Connecting the RF adaptor cable to J702 opens a set of contacts on the antenna strip line assembly. This disconnects the collapsible antenna and connects the transmitter output to J702-3. Connection to chassis ground is made at J702-4.

#### LOW-PASS FILTER

The low-pass filter is mounted on Systems Board A701. The filter consists of L4, C5, C8, C10 and C18. The filter output is fed to System switching relay K1 and then coupled through a 50-ohm antenna matching network (L701 and C15) to the antenna.

		•
		•

### **MODULATION LEVEL ADJUSTMENT**

The MOD ADJUST (R8) was adjusted to the proper setting before shipment and should not normally require readjustment. This setting permits approximately 75% modulation for the average voice level. The audio peaks which would cause overmodulation are clipped by the modulation limiter. The limiter, in conjunction with the de-emphasis network, instantaneously limits the slope of the audio wave to the modulator, thereby preventing overmodulation while preserving intelligibility.

### TEST EQUIPMENT

- 1. Audio oscillator Model 4EX6A10
- 2. A deviation meter
- 3. An output meter or a VTVM
- 4. Test Adaptor Model 4EX12A10

#### PROCEDURE

- 1. Connect the equipment as shown in the Test Procedure on the back of this page.
- 2. Apply a 140 millivolt signal at 1000 Hz to the Test Adaptor. If the Test Adaptor is not used, apply a 14 millivolt signal to Pin 4 (Mike Hi) and Pin 1 of Accessory
- 3. With the signal applied, adjust Tuning Control 1 for zero modulation symmetry on the lowest channel frequency.
- 4. For transmitters without Channel Guard, set MOD ADJUST R8 for a 4.5-kilohertz swing with the deviation polarity which gives the highest reading as indicated on the frequency modulation monitor.
- 5. For transmitters with Channel Guard, check the Channel Guard Modulation as shown in Step 2 of the transmitter Test Procedure. With Channel Guard tone applied, set the deviation as described in Step 4 above.
- 6. For multifrequency transmitters, set the deviation as described in Step 4 on the channel producing the largest amount of deviation.

### REDUCED POWER OPERATION

Option 4475 - 1 Watt Output per EIA Also for Shipboard Marine Applications (FCC Part 83)

In some services, FCC regulations do not permit the use of the two-watt rated output. In addition, operating at a reduced power output will extend the battery life in those applications where the two-watt output is not required. After completing Step 9 of the two-watt Alignment Procedure, reduce the output power as follows:

- 1. Turn Tuning Control 9 (PWR ADJ) clockwise until the power output is reduced by onehalf of the difference between the power output noted in Step 9 and the desired power output. For example: If the output noted in Step 9 was 2 watts and the desired output is 1 watt, adjust Tuning Control 9 for an output of 1.5 watts.
- 2. Adjust Tuning Control 6 to reduce the output to the desired level. Tuning Control 6 has two positions producing the same power output. Select the position having the lowest current drain.
- 3. If the limit of adjustment on Tuning Control 6 is reached and the power output or current drain is still too high, detune Tuning Control 8 in the direction resulting in less current drain and power output. Then alternately adjust Tuning Controls 6 and 8 for the desired output and drain.

### 100-200 MILLIWATT TRANSMITTER ALIGNMENT (ET-95-A)

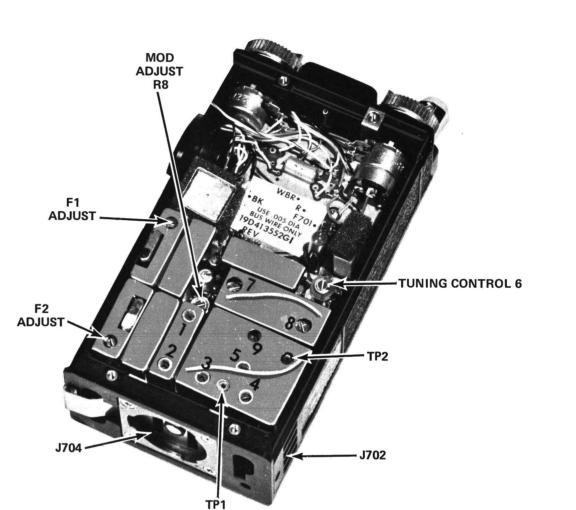
- GE Test Set Model 4EX3Al0 (or 4EX8Kll) or equivalent 20,000 ohm-per-volt meter.
- GE Test Regulator Model 4EX18A10, or an ammeter capable of measuring 0.25 amperes.
- A 50-ohm, terminating wattmeter connected to external antenna jack J702 thru RF adaptor cable 19C317633G2

#### PRELIMINARY CHECKS AND ADJUSTMENTS

- 1. In multi-frequency transmitters, set the channel selector switch to the lowest channel frequency.
- 2. Set the slugs in Tuning Controls 1 thru 5 even with the top of the can. When properly aligned, the slugs will
- 3. Set Tuning Control 9 (PWR ADJ) fully counterclockwise, and Tuning Control 6 to mid range.
- 4. If using Test Set 4EX3A10 and Test Regulator 4EX18A10, connect the Test Set to the metering jack on the Test Regulator, and set the Regulator for 7.5 volts. Switch the Test Set range to the Test 1 position. Place the test selector switch on position "I" to check the supply voltage (read on the 1-volt scale as 10-volts full scale. Switch to position "G" for current drain readings (read on the 1-volt scale as 1 ampere full scale).
- 5. Test Point meter reading made with the (+) meter lead to TPl and the (-) lead to system ground.
- 6. All adjustments made with the transmitter keyed.

#### ADJUSTMENT PROCEDURE

Step	Tuning Control	Typical Meter Reading	Procedure	
1	1	Maximum mA	Adjust Tuning Control 1 for maximum transmitter current.	
2	2	Minimum mA	Adjust Tuning Control 2 for minimum transmitter current.	
3	3		Adjust Tuning Control 3 for maximum meter reading at TP1.	
4	1, 2 & 3	0.8 volts	Adjust Tuning Controls 1, 2 and 3 for maximum meter reading at TP1. Repeat the adjustments until no further increase in meter reading is obtained.	
5	4 and 5	Maximum mA	Adjust Tuning Controls 4 and 5 for maximum transmitter current.	
6	7 and 8	Maximum Power Output	Adjust Tuning Controls 7 and 8 for maximum power output.	
7	4, 5, 7, & 8	Maximum Power Output	Adjust Tuning Controls 4, 5, 7 and 8 for maximum power output. Repeat the adjustments until no further increase in power output is obtained.	
8	9 (PWR ADJ)		Set Tuning Control 9 (PWR ADJ) for the desired power output (100 to 200 milliwatts)	
			FREQUENCY ADJUSTMENT	
9			With no modulation, adjust the F1 (and F2) crystal trimmer for proper oscillator frequency. Next, refer to the Modulation Adjustment.	
			It is recommended that all frequency adjustments be made when the equipment is at a temperature of approximately 75°F. In no case should frequency adjustments be made when the equipment is outside the temperature range of 60°F to 90°F.	



### 1 to 2 WATT TRANSMITTER ALIGNMENT (ET-96-A)

### EQUIPMENT REQUIRED:

- GE Test Set Model 4EX3A10 (or 4EX8K11) or equivalent 20,000 ohm-per-volt meter.
- GE Test Regulator Model 4EX18A10, or an ammeter capable of measuring one ampere.
- A 50-ohm, terminating wattmeter connected to external antenna jack J702 thru RF adaptor cable 19C317633G2 (Option 4466).
- A frequency counter.

### PRELIMINARY CHECKS AND ADJUSTMENTS

- 1. In multi-frequency transmitters, set the channel selector switch to the lowest channel frequency,
- 2. Set the slugs in Tuning Controls 1 thru 5 even with the top of the can. When properly aligned, the slugs will be between the
- 3. Set Tuning Control 9 (PWR ADJ) fully counterclockwise, and Tuning Control 6 to mid range.
- 4. If using Test Set 4EX3AlO and Test Regulator 4EX18AlO, connect the Test Set to the metering jack on the Test Regulator. Then connect the Regulator output to J704 on the radio, and set the Regulator for 6 volts. Switch the Test Set range to the Test 1 position. Place the test selector switch on position "I" to check the supply voltage (read on the 1-volt scale as 10-volts full scale). Switch to position "G" for current drain readings (read on the 1-volt scale as 1 ampere full scale).

Tunical Meter

- 5. Test Point meter reading made with the (+) meter lead to TPl and the (-) lead to system ground.
- 6. All adjustments made with the transmitter keyed.

### ADJUSTMENT PROCEDURE

Step	Tuning Control	Typical Meter Reading	Procedure
1	1	Maximum mA	Adjust Tuning Control 1 for maximum transmitter current
2	2	Minimum mA	Adjust Tuning Control 2 for minimum transmitter current
3	3		Adjust Tuning Control 3 for maximum meter reading at TPl
4	1, 2 & 3	0.8 volts	Adjust Tuning Controls 1, 2 and 3 for maximum met reading at TPl. Repeat the adjustments until no further increase in meter reading is obtained.
5	4, 5, & 7	Maximum mA	Adjust Tuning Controls 4, 5 and 7 for maximum transmitter current.
6	8, 6, 7, 5, & 4	Maximum Power Output	Adjust Tuning Controls 8, 6, 7, 5 and 4 in that order for maximum power output.
7	1, 2 & 3		Repeat Step 4
8			Increase the supply voltage to 7.5 volts
9	7	Maximum Power Output	Adjust Tuning Control 7 for maximum power output and note the power output. If the power output is correct, Tuning Controls 6 and 8 can be alternated tuned for the best ratio of current drain to RF power output. If the power output is too low, refer to Step 10.  If the power output is too high, refer to REDUCED POWER OPERATION elsewhere on this page.
10	6, 8 & 7	Maximum Power Output	If the power output is too low, readjust Tuning Controls 6, 8 and 7 in that order for maximum power output. Repeat until the desired power output is obtained.
			FREQUENCY ADJUSTMENT
11			With no modulation, adjust the F1 (and F2) crystal trimmer for proper oscillator frequency. Next, refer to the Modulation Adjustment.  NOTE  It is recommended that all frequency adjustments be made when the equipment is at a temperature of approximately 75°F.  In no case should frequency adjustments
			be made when the equipment is outside the temperature range of 60°F to 90°F.

### ALIGNMENT PROCEDURE

132—174 MHZ TRANSMITTER TYPES ET-95-A & ET-96-A

Issue 1

LBI-4646

### TEST PROCEDURES

These Test Procedures are designed to assist you in servicing a transmitter that is operating—but not properly. Problems encountered could be low power output, tone and voice deviation, defective audio sensitivity and modulator adjust control set too high. By following the sequence of test steps starting with Step 1, the defect can

be quickly localized. Once a defect is pin-pointed, refer to the "Service Check" and the additional corrective measures included in the Transmitter Troubleshooting Procedure. Before starting with the Transmitter Test Procedures, be sure the transmitter is tuned and aligned to the proper operating frequency.

# TEST EQUIPMENT REQUIRED for test hookup shown:

- 1. Wattmeter similar to: Bird # 43
- 4. Deviation Meter (with a .75 kHz scale) sim-ilar to:
  - Measurements # 140 Lampkin # 205A

2. VTVM similar to:
Triplett # 850

Heath # 1M-21

- 5. GE Test Adaptor Model 4EX12A10.
- 3. Audio Generator similar to:

  GE Model 4EX6A10 or

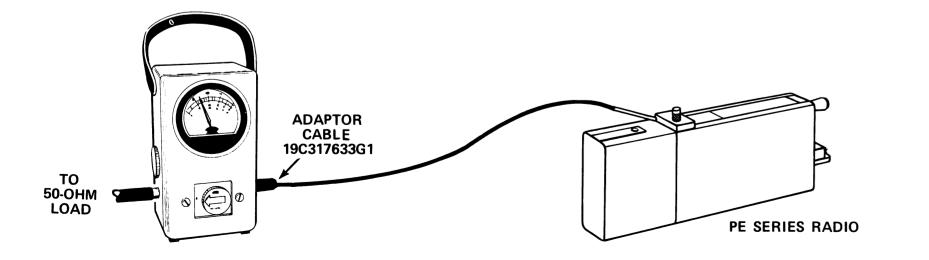
  Heath # IG-72

### STEP 1

### POWER MEASUREMENT

### **TEST PROCEDURE**

A. Connect transmitter output to wattmeter as shown below. GE adaptor cable 19C317633G1 is recommended for accurate power output readings.



B. Key transmitter and check wattmeter for desired power output..

### SERVICE CHECK

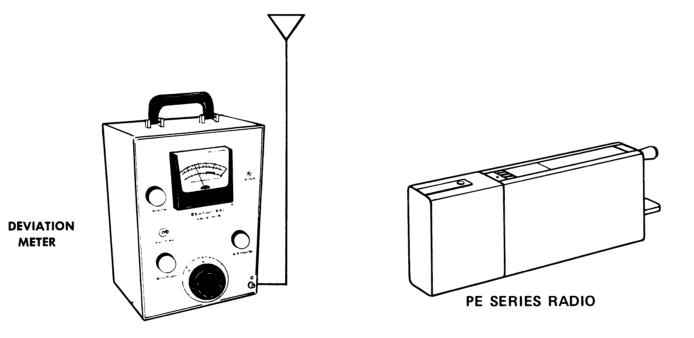
Refer to Service Hints on Transmitter Troubleshooting Procedure.

### STEP 2

### TONE DEVIATION WITH CHANNEL GUARD

### TEST PROCEDURE

A. Set up Deviation Meter and monitor output of transmitter as shown below:



- B. Set MOD ADJUST R8 fully counterclockwise.
- C. Key transmitter and check for approximately 0.75-kHz deviation. If reading is low or high, refer to the Channel Guard Troubleshooting Procedure (see Table of Contents)

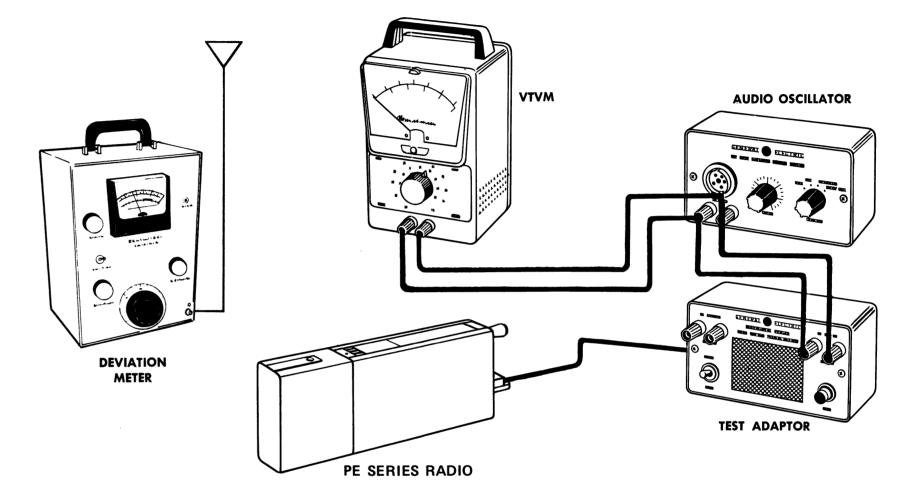
NOTES--The Tone Deviation Test Procedures should be repeated every time the Tone Frequency is changed.

### STEP 3

### **VOICE DEVIATION AND SYMMETRY**

### TEST PROCEDURE

A. Connect test equipment to transmitter as shown below:

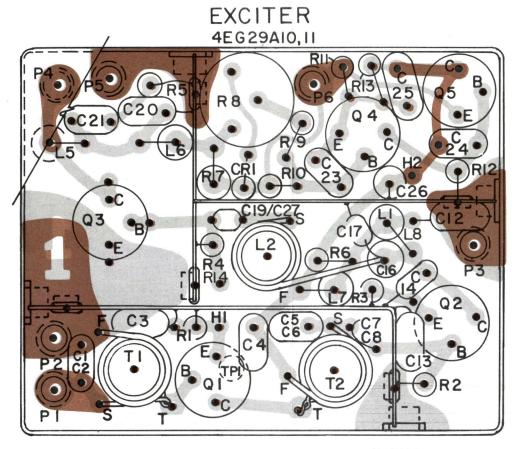


- B. Set the generator output to 140 millivolts RMS and frequency to 1 kHz. If the Test Adaptor is not used, set the generator output for 14 millivolts.
- C. Key the transmitter and adjust Deviation Meter to carrier frequency.
- Deviation reading should be  $\pm 4.5$  kHz. If the deviation is not 4.5 kHz, set the deviation as directed on the Transmitter Alignment Procedure (see Table of Contents).

NOTES -- These transmitters are adjusted for 4.5 kHz deviation at the factory. The factory adjustment will prevent the transmitter from deviating more than 5.0 kHz under the worst conditions of frequency, voltage and temperature.

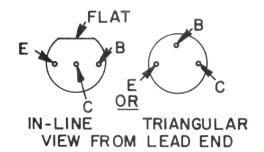
If the deviation reading plus (+) or minus (-) differs by more than 0.5 kHz:

- E. Refer to the Modulation Adjustment on the Transmitter Alignment Procedure.
- F. Check Audio Sensitivity by reducing generator output until deviation falls to 3 kHz. Voltage should be LESS than 14 millivolts.



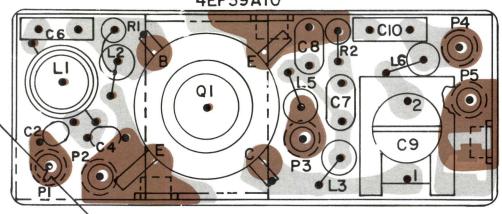
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# LEAD IDENTIFICATION FOR EXCITER BD



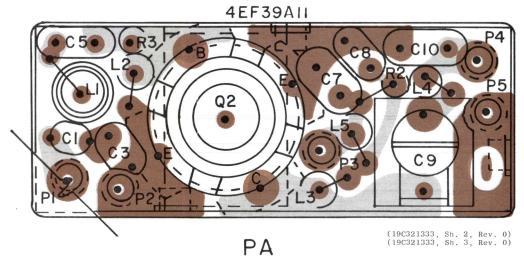
NOTE: LEAD ARRANGEMENT, AND NOT CASE SHAPE, IS DETERMINING FACTOR FOR LEAD IDENTIFICATION.

### PA 4EF39A10

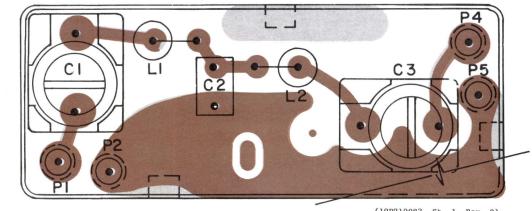


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## PA



19B219084G1



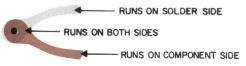
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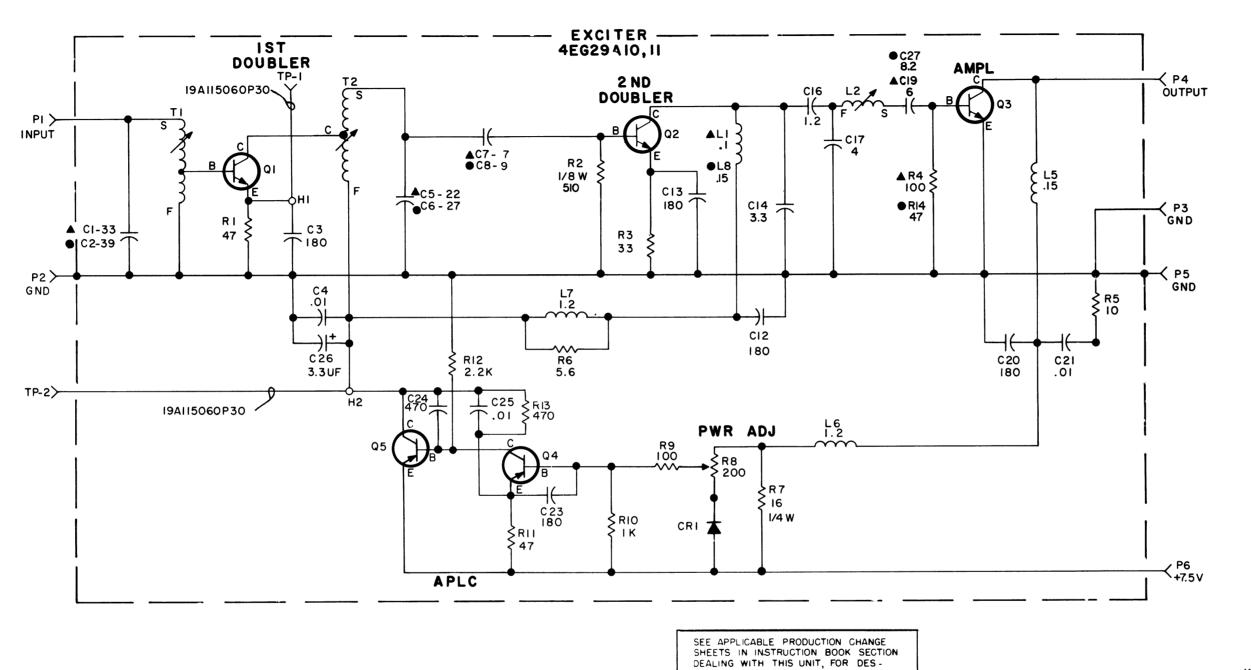
### **OUTLINE DIAGRAM**

(19B219346, Rev. 2)

132—174 MHz TRANSMITTER EXCITER AND PA ASSEMBLY

Issue 2



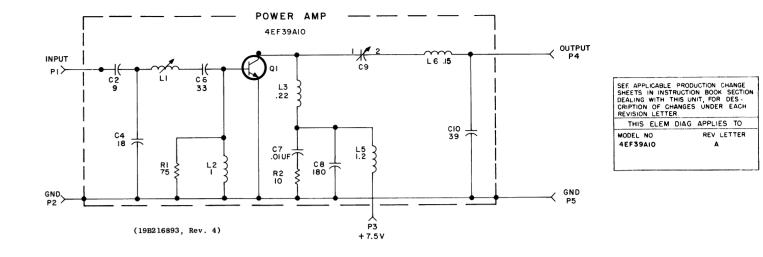


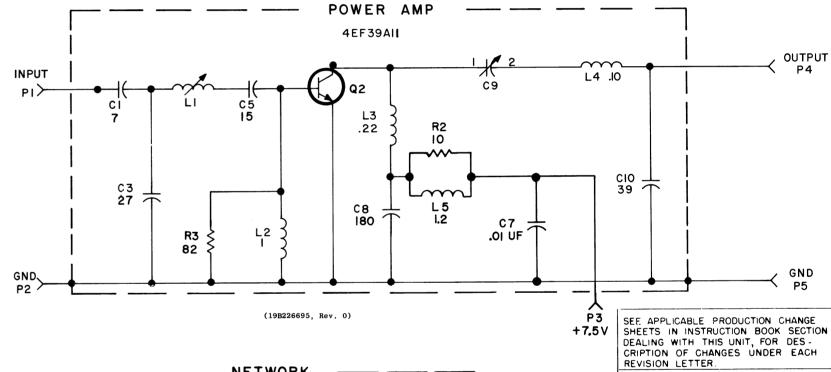
IN ORDER TO RETAIN RATED EQUIPMENT PERFORMANCE, REPLACEMENT OF ANY SERVICE PART SHOULD BE MADE ONLY WITH A COMPONENT HAVING THE SPECIFICATIONS SHOWN ON THE PARTS LIST FOR THAT PART

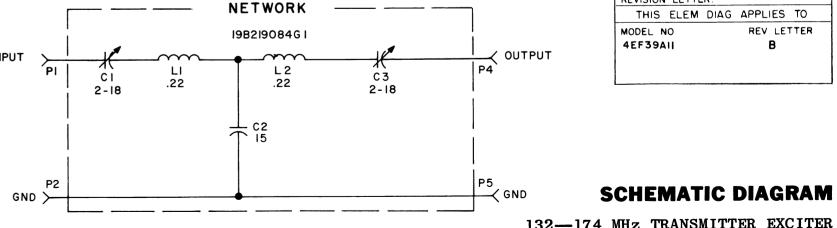
ALL RESISTORS ARE 1/10 WATT UNLESS OTHERWISE SPECIFIED AND RESISTOR VALUES IN OHMS UNLESS FOLLOWED BY K=1000 OHMS OR MEG=1,000,000 OHMS CAPACITOR VALUES IN PICOFARADS (EQUAL TO MICROMICROFARADS) UNLESS FOLLOWED BY UF = MICROFARADS, INDUCTANCE VALUES IN MICROHENRYS UNLESS FOLLOWED BY
MH= MILLIHENRYS OR H=HENRYS

CRIPTION OF CHANGES UNDER EACH REVISION LETTER. THIS ELEM DIAG APPLIES TO REV LETTER MODEL NO 4EG29AIO 4EG29AII

ALL RESISTORS ARE 1/8 WATT UNLESS OTHERWISE SPECIFIED AND RESISTOR VALUES IN OHMS UNLESS FOLLOWED BY K=1000 OHMS OR MEG = 1,000,000 OHMS CAPACITOR VALUES IN PICOFARADS (EQUAL TO MICROMICROFARADS) UNLESS FOLLOWED BY UF = MICROFARADS. INDUCTANCE VALUES IN MICROHENRYS UNLESS FOLLOWED BY MH= MILLIHENRYS OR H=HENRYS.







(19B216892, Rev. 1)

132-174 MHz TRANSMITTER EXCITER AND PA ASSEMBLY

(19C317404, Rev. 6)

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THIS ELEM DIAG APPLIES TO

REV LETTER

### LBI-4646

### PARTS LIST

#### LBI-4233D

	100 154	LBI-4233D				C9	19A116462P3	Variable: less than 2 pf to more than 20 pf, 100 VDCW, -320 PPM/°C.
	132-174 MHZ T	RANSMITTER EXCITER AND PA MODULES	Pl thru	19A115834P4	Contact, electrical: sim to AMP 2-332070-9.	C10	19A116114P50	Ceramic: 39 pf ±5%, 100 VDCW; temp coef 0 PPM.
			P6		TT. 170 - 0 TO - 0			Nues.
			Ql	19A115328P1	TRANSISTORS	P1	19A115834P4	
SYMBOL	GE PART NO.	DESCRIPTION	Q2*	19A116201P3	Silicon, NPN.	thru P5	13411003424	contact, electrical. Sim to Amr 2-332070-9.
STMIDGE	GETAKTING.	DESCRIPTION	'-		Earlier than REV A:	"		
1		EXCITER MODULE		19A116201P1	Silicon, NPN.	Ll	19B216921G1	Coil. Includes:
		MODEL 4EG29A10 132-150.8 MHz MODEL 4EG29A11 150.8-174 MHz	Q3	19A116201P1	Silicon, NPN.		19B209436P1	Tuning slug.
		model thousand 100,0-114 mile	Q4	19A115768P1	Silicon, PNP; sim to Type 2N3702.	L2*	19B209420P113	Coil, RF: 1.00 μh ±10%, 0.74 ohms DC res max;
			and Q5					sim to Jeffers 4426-6.
C1	19A116114P2047	Ceramic: 33 pf ±5%, 100 VDCW; temp coef -80 PPM.			RESISTORS		19B209420P125	Earlier than REV A:  Coil, RF: 10.0 µh ±10%, 3.10 ohms DC res max:
C2	19A116114P2050	Ceramic: 39 pf ±5%, 100 VDCW; temp coef -80 PPM.	R1	3R151P470J	Composition: 47 ohms ±5%, 1/8 w.		1982094209123	sim to Jeffers 4446-4.
C3	19A116114P10073	Ceramic: 180 pf ±10%, 100 VDCW; temp coef -3300 PPM.	R2*	3R151P511J	Composition: 510 ohms ±5%, 1/8 w.	L3	19B209420P105	Coil, RF: 0.22 µh ±10%, 0.14 ohms DC res max; sim to Jeffers 4416-5.
C4	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie 8121-050-W5R.			In 4EG29AlO of REV B and earlier: In 4EG29All of REV A and earlier:	L4	19B209420P101	Coil, RF: 0.10 µh ±10%, 0.08 ohms DC res max;
C5	19A116114P2041	Ceramic: 22 pf ±5%, 100 VDCW; temp coef -80 PPM.		3R151P122J	Composition: 1200 ohms ±5%, 1/8 w.			sim to Jeffers 4416-1.
C6	19A116114P2045	Ceramic: 30 pf ±5%, 100 VDCW; temp coef -80 PPM.	R3	3R151P330J	Composition: 33 ohms ±5%, 1/8 w.	L5	19B209420P114	Coil, RF: 1.20 µh ±10%, 0.18 ohms DC res max; sim to Jeffers 4436-1.
C7*	19A116114P24	Ceramic: 7 pf ±5%, 100 VDCW; temp coef 0 PPM.	R4*	3R151P101J	Composition: 100 ohms ±5%, 1/8 w. Deleted in	L6	19B209420P103	Coil, RF: 0.15 µh ±10%, 0.10 ohms DC res max;
		Earlier than REV A:	R5	3R151P100J	4EG29A10 by REV B.  Composition: 10 ohms ±5%, 1/8 w.	1		sim to Jeffers 4416-3.
	19A116114P14	Ceramic: 4 pf ±5%, 100 VDCW; temp coef 0 PPM.	R6	3R151P5R6J	Composition: 10 ohms 15%, 1/8 w.  Composition: 5.6 ohms ±5%, 1/8 w.	1		TRANSISTORS
C8*	19A116114P30	Ceramic: 9 pf ±5%, 100 VDCW; temp coef 0 PPM.	R7	3R152P160J	Composition: 16 ohms ±5%, 1/4 w.	Q1 *	19A116259P1	Silicon, NPN. Deleted in Gl by REV B.
		Earlier than REV A:	R8	19A116412P1	Variable, cermet: 200 ohms ±10% 1/2 w:	Q2*	19A134132P1	Silicon, NPN. Added to Gl by REV B.
	19A116114P18	Ceramic: 5 pf ±5%, 100 VDCW; temp coef 0 PPM.			sim to Helipot Model 62 PF.			RESISTORS
C12 and	19A116114P10073	Ceramic: 180 pf ±10%, 100 VDCW; temp coef -3300 PPM.	R9	3R151P101J	Composition: 100 ohms ±5%, 1/8 w.	R1*	3R151P750J	Composition: 75 ohms ±5%, 1/8 w. Deleted in
C13			R10 R11	3R151P102J 3R151P470J	Composition: 1000 ohms ±5%, 1/8 w.	1	2015101007	G1 by REV B.
C14	19A116114P12	Ceramic: 3.3 pf ±5%, 100 VDCW; temp coef 0 PPM.	R12	3R151P470J	Composition: 47 ohms ±5%, 1/8 w.  Composition: 2200 ohms ±5%, 1/8 w.	R2 R3*	3R151P100J 3R151P820J	Composition: 10 ohms ±5%, 1/8 w.  Composition: 82 ohms ±5%, 1/8 w. Added to G1
C16	5491601P122	Phenolic: 1.2 pf ±5%, 500 VDCW.	R13	3R151P471J	Composition: 470 ohms ±5%, 1/8 w.	1 134	3R131F8203	by REV B.
C17 C19*	19A116114P14 19A116114P20	Ceramic: 4 pf ±5%, 100 VDCW; temp coef 0 PPM.	R14*	3R151P470J	Composition: 47 ohms ±5%, 1/8 w. Added to			200 MILLIWATT PA MODULE
C19+	194110114P20	Ceramic: 6 pf ±5%, 100 VDCW; temp coef 0 PPM. Deleted in 4EG29AlO by REV B.			4EG29A10 by REV B.			19B219084G1
C20	19A116114P10073	Ceramic: 180 pf ±10%, 100 VDCW; temp coef -3300 PPM.			TRANSFORMERS			
C21	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie 8121-050-W5R.	Tl	19B216910G2	Coil. Includes:	C1	19A116462P3	Variable: less than 2 pf to 20 pf, 100 VDCW, -320 PPM/°C.
C23	19A116114P10073	Ceramic: 180 pf ±10%, 100 VDCW; temp coef	T2	19B209436P1 19B216934G1	Tuning slug. Coil. Includes:	C2	19Al16114P36	Ceramic: 15 pf ±5%, 100 VDCW; temp coef 0 PPM.
C24	19A116192P2	-3300 PPM.	"	19B209436P1	Tuning slug.	СЗ	19A116462P3	Variable: less than 2 pf to 20 pf, 100 VDCW, -320 PPM/°C.
C24	19411019292	Ceramic: 470 pf ±20%, 50 VDCW; sim to Erie 8111-050-W5R.						
C25	19A116192P1	Ceramic: 0.01 µf ±20%, 50 VDCW; sim to Erie 8121-050-W5R.			2 WATT PA MODULE MODEL 4EF39A10 132-150.8 MHz MODEL 4EF39A11 150.8-174 MHz	ы	198209420P5	
C26	5491674P36	Tantalum: 3.3 µf ±20%, 10 VDCW; sim to Sprague Type 162D.				and L2		sim to Jeffers 4416-5.
C27*	19A116114P28	Ceramic: 8,2 pf ±5%, 100 VDCW; temp coef 0 PPM.	C1	19A116114P24	Corporate 7 of the 100 MDGW town and 0 DDW			
		Added to 4EG29A10 by REV B.	C2	19A116114P24 19A116114P30	Ceramic: 7 pf ±5%, 100 VDCW; temp coef 0 PPM.  Ceramic: 9 pf ±5%, 100 VDCW; temp coef 0 PPM.	Pl thru	19A115834P4	Contact, electrical: sim to AMP 2-332070-9.
		DIODES AND RECTIFIERS	C3*	19A116114P2044	Ceramic: 27 pf ±5%, 100 VDCW; temp coef -80 PPM.	P5		
CR1	19A115250P1	Silicon.			In REV A and earlier:			MISCELLANEOUS
				19A116114P36	Ceramic: 15 pf ±5%, 100 VDCW; temp coef 0 PPM.		19A127337P2	Nut, knurled: 8-32. (Used with Q1 and Q2 in
Ll*	19B209420P101	Coil, RF: 0.10 µh ±10%, 0.08 ohms DC res max;	C4	19Al16114P38	Ceramic: 18 pf ±5%, 100 VDCW; temp coef 0 PPM.			PA Module).
		sim to Jeffers 4416-1. Deleted in 4EG29A10 by REV B,	C5*	19A116114P36	Ceramic: 15 pf ±5%, 100 VDCW; temp coef 0 PPM.			
L2	19B216935G1	Coil. Includes:			In REV A and earlier:			
	19B209436P1	Tuning slug.		19A116114P8039	Ceramic: 20 pf ±5%, 100 VDCW; temp coef -80 PPM.			
L5	19B209420P103	Coil, RF: 0.15 µh ±10%, 0.10 ohms DC res max; sim to Jeffers 4416-3.	C6	19A116114P2047	Ceramic: 33 pf ±5%, 100 VDCW; temp coef -80 PPM.			
L6	19B209420P114	Coil, RF: 1.20 µh ±10%, 0.18 ohms DC res max;	C7	19A116192P1	Ceramic: 0.01 $\mu$ f $\pm 20\%$ , 50 VDCW; sim to Erie 8121-050-W5R.			
and L7		sim to Jeffers 4436-1.	C8	19A116114P10073	Ceramic: 180 pf ±10%, 100 VDCW; temp coef -3300 PPM.			
L8*	19B209420P103	Coil, RF: 0.15 µh ±10%, 0.10 ohms DC res max; sim to Jeffers 4416-3. Added to 4EG29Al0 by REV B.						
1						-		
						1		
l	1	j j	1	1	i l	1	I	I

GE PART NO.

SYMBOL

DESCRIPTION

SYMBOL | GE PART NO.

DESCRIPTION

14 COMPONENTS ADDED, DELETED OR CHANGED BY PRODUCTION CHANGES

### PRODUCTION CHANGES

Changes in the equipment to improve performance or to simplify circuits are identified by a "Revision Letter", which is stamped after the model number of the unit. The revision stamped on the unit includes all previous revisions. Refer to the Parts List for descriptions of parts affected by these revisions.

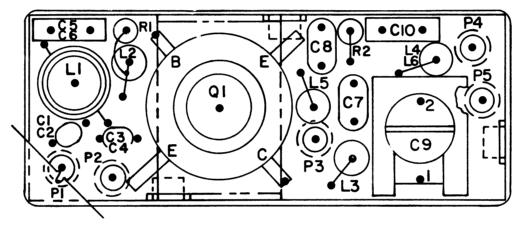
- REV. A Exciter Model 4EG29A10, 11
- To increase Power Output. Changed C7, C8 and Q2.
- REV. B Exciter Model 4EG29A10

To improve operation. Added C27, R14 and L8. Changed R6. Deleted C19, L1 and R4.

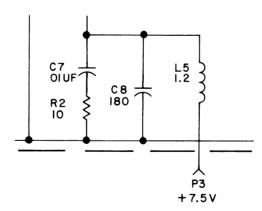
- REV. C To reduce current drain in exciter. Changed R2.
- REV. B Exciter Model 4EG29A11
  - To reduce current drain in exciter. Changed R2.
- REV. A Power Amplifier Models 4EF39Al0 & 11
- To improve power output. Changed L2.
- REV. B Power Amplifier Model 4EF39All

To incorporate new transistor.
Deleted Q1 and R1. Changed PWB, C3 and C5.
Added Q2 and R3.

Outline Diagram was:

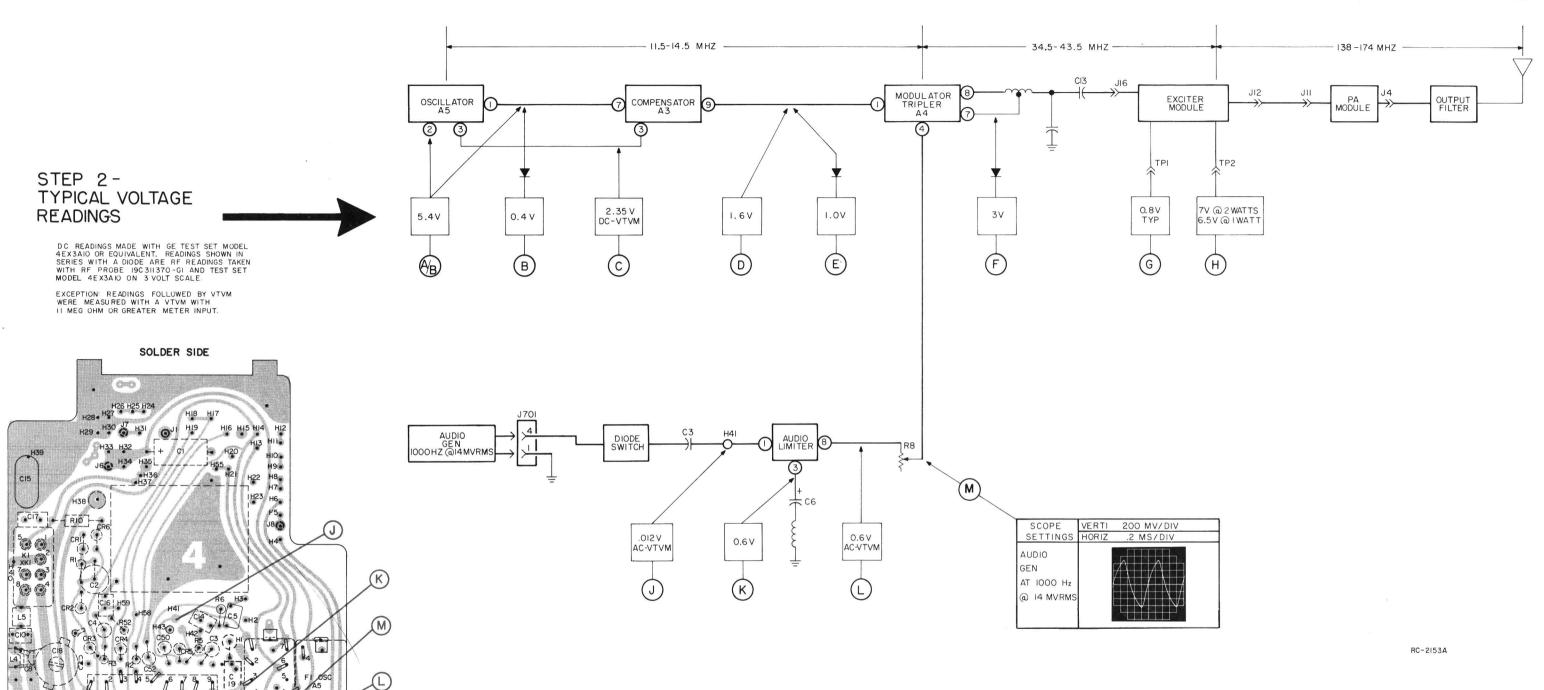


### Schematic Diagram was:



### STEP - 1 QUICK CHECKS

SYMPTOM	QUICKCHECK			
No power output	1. Replace the PA module unit with Network 19B219084Gl. Realign the transmitter and check for 200 milliwatt output.  2. If the proper output is obtained, re-			
	place the PA module.  3. If no power output is obtained after performing Step 1, check the reading at TP1. If no reading is obtained at TP1, check readings at (F), (E) and (B). If TP1 reading is correct, replace the Exciter module.			
Low power output	Low battery voltage (refer to Battery Checks in operation section of the manual).      Check the transmitter alignment.			
Distorted or no audio with normal RF output.	1. Check voltage readings at ① , ⑥ , ① and ᠓			
	2. Improper setting of Mod Adjust R8.			
	3. Shorted C3 or C6 on Audio Board.			
	4. Bad microphone.			
No reading at TP1	Check voltage readings at $\widehat{\mathbb{A}}$ , $\widehat{\mathbb{B}}$ , $\widehat{\mathbb{D}}$ , $\widehat{\mathbb{E}}$ and $\widehat{\mathbb{F}}$ .			



MODULATOR COMPENSATOR

(RC-2156) (19D416226, Rev. 2) (19D413549, Sh. 2, Rev. 4)

### TROUBLESHOOTING PROCEDURE

132--174 MHz TRANSMITTER TYPE ET-95-A & ET-96-A

Issue 2

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### **ORDERING SERVICE PARTS**

Each component appearing on the schematic diagram is identified by a symbol number, to simplify locating it in the parts list. Each component is listed by symbol number, followed by its description and GE Part Number.

Service Parts may be obtained from Authorized GE Communication Equipment Service Stations or through any GE RadioCommunication EquipmentSales Office. When ordering a part, be sure to give:

- GE Part Number for component
   Description of part
   Model number of equipment
   Revision letter stamped on unit

These instructions do not purport to cover all details or variations in equipment nor to provide for every possible contingency to be met in connection with installation, operation or maintenance.

Should further information be desired, or should particular problems arise which are not covered sufficiently for the purchaser's purposes, contact the nearest Radio Communication Equipment Sales Office of the General Electric Company.

### MAINTENANCE MANUAL

LBI-4646

DF-3157

MOBILE RADIO DEPARTMENT
GENERAL ELECTRIC COMPANY ◆ LYNCHBURG, VIRGINIA 24502

