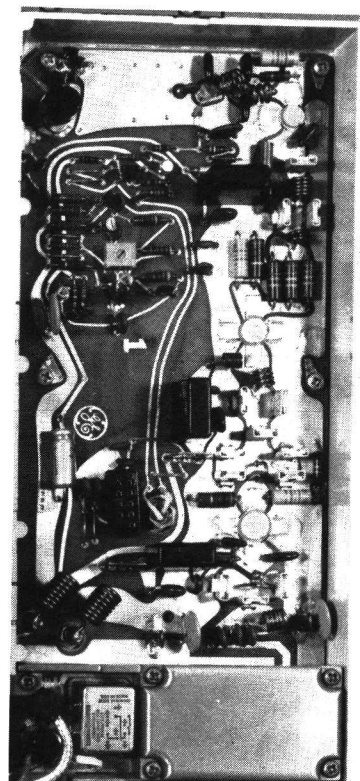
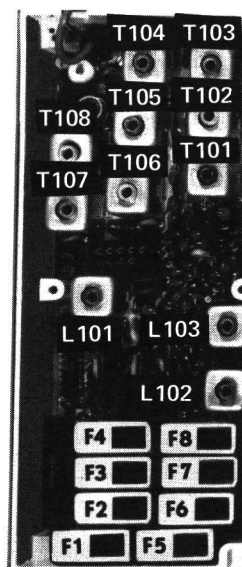




GE Mobile Communications



MASTR® II

25-50 MHz, 50-WATT STATION TRANSMITTER

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WARNING

Although the highest DC voltage in the MASTR II station transmitter is the 12 VDC supply voltage, high currents may be drawn under short circuit conditions. These currents can possibly heat metal objects such as tools, rings, watchbands, etc., enough to cause burns. Be careful when working near energized circuits! High-level RF energy in the transmitter Power Amplifier assembly can cause RF burns upon contact. KEEP AWAY FROM THESE CIRCUITS WHEN THE TRANSMITTER IS ENERGIZED!

DESCRIPTION

Transmitter Types KT-56-A, C, KT-57-A, C & KT-58-A, C are crystal-controlled, phase modulated transmitters designed for one-through four-frequency operation in the 25 to 50 megahertz band. The solid state transmitter utilizes both integrated circuits (ICs) and discrete components, and consists of the following assemblies:

- Exciter Board; with audio, modulator, amplifier and multiplier stages.
- Power Amplifier Assembly; with amplifier, driver, PA, power control, filter and antenna switch.

CIRCUIT ANALYSIS

EXCITER

The exciter uses nine transistors and two integrated circuits to drive the PA assembly. The exciter can be equipped with up to four Integrated Circuit Oscillator Modules (ICOMs). The ICOM crystal frequency ranges from approximately 8.33 to 16.67 megahertz, and the crystal frequency is multiplied three times (divided by four and multiplied by 12 for a multiplication of three).

Audio, supply voltages and control functions are connected through P902.

Centralized metering jack J103 is provided for use with GE Test Set Model 4EX3A11 or Test Kit 4EX3K12. The test set meters the modulator, multiplier and amplifier stages.

ICOMS

Three different types of ICOMs are available for use in the exciter. Each of the ICOMs contains a crystal-controlled Colpitts oscillator, and two of the ICOMs contain compensator ICs. The different ICOMs are:

- 5C-ICOM - contains an oscillator and a 5 part-per-million ($\pm 0.0005\%$) compensator IC. Provides compensation for EC-ICOMs.
- EC-ICOM - contains an oscillator only. Requires external compensation from a 5C-ICOM.
- 2C-ICOM - contains an oscillator and a 2 PPM ($\pm 0.0002\%$) compensator IC. Will not provide compensation for an EC-ICOM.

The ICOMs are enclosed in an RF shielded can with the type ICOM (5C-ICOM, EC-ICOM or 2C-ICOM) printed on the top of the can. Access to the oscillator trimmer is obtained

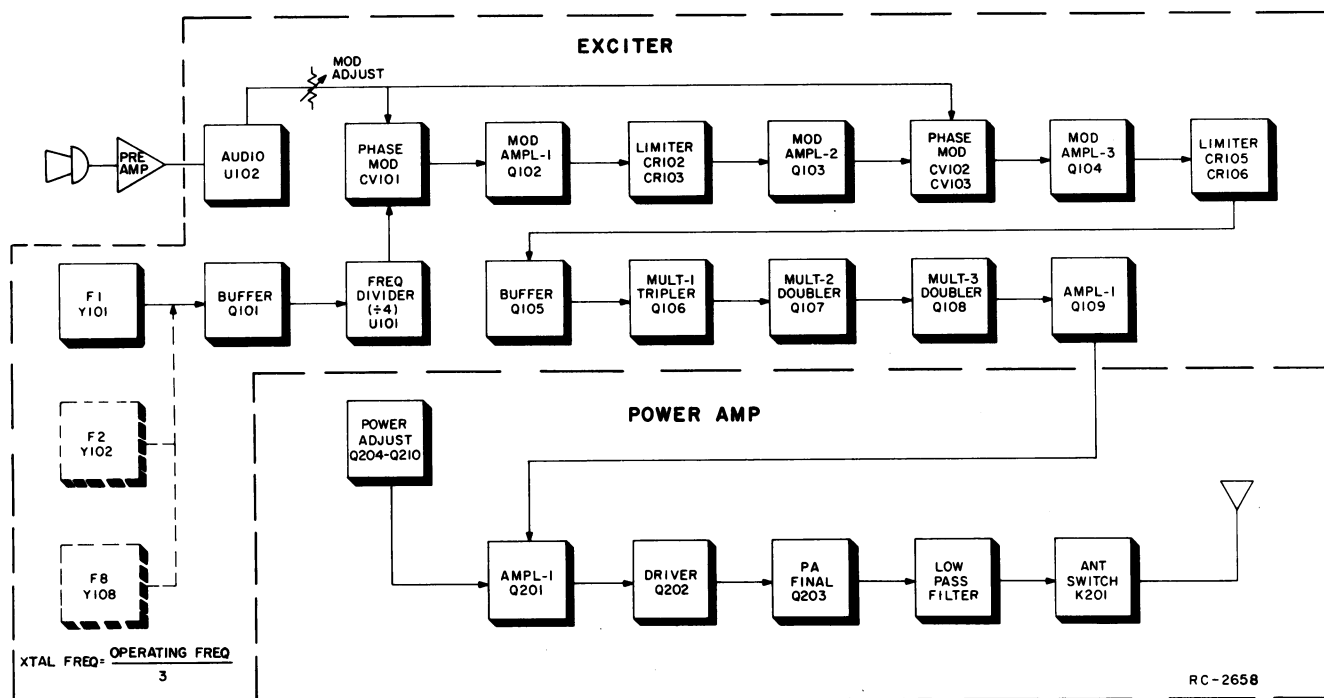


Figure 1 - Transmitter Block Diagram

by prying up the plastic tab on the top of the can. The tabs can also be used to pull the ICOMs out of the radio.

Frequency selection is accomplished by switching the ICOM keying lead (terminal 6) to A-. The oscillator is turned on by applying a keyed +10 Volts to the external oscillator load resistor.

CAUTION

All ICOMs are individually compensated at the factory and cannot be repaired in the field. Any attempt to repair or change an ICOM frequency will void the warranty.

In transmitter types KT-56-A, C using EC-ICOMs, at least one 5C-ICOM must be used. The 5C-ICOM is normally used in the receiver F1 position, but can be used in any transmit or receive position. One 5C-ICOM can provide compensation for up to 8 EC-ICOMs in the transmit and receiver. Should the 5C-ICOM compensator fail in the open mode, the EC-ICOMs will still maintain 2 PPM frequency stability from 0°C to 55°C (+32°F to 131°F) due to the regulated compensation voltage (5 Volts) from the 10-Volt regulator IC. In transmitter types KT-57-A, C and KT-58-A, C at least one 5C-ICOM is required for the transmitter and at least one 5C-ICOM is required for the receiver. If desired, up to 8 5C-ICOMs may be used in the station.

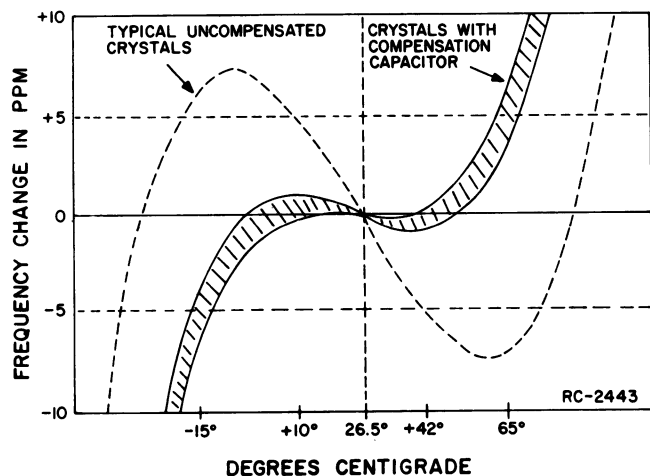


Figure 2 - Typical Crystal Characteristics

The 2C-ICOMs are self-compensated at 2 PPM and will not provide compensation for EC-ICOMs.

Oscillator Circuit

The quartz crystals used in ICOMs exhibit the traditional "S" curve characteristics of output frequency versus operating temperature.

At both the coldest and hottest temperatures, the frequency increases with increasing temperature. In the middle temperature range (approximately 0°C to +55°C), frequency decreases with increasing temperature.

Since the rate of change is nearly linear over the mid-temperature range, the output frequency change can be compensated by choosing a parallel compensation capacitor with a temperature coefficient approximately equal and opposite that of the crystal.

Figure 2 shows the typical performance of an uncompensated crystal as well as the typical performance of a crystal which has been matched with a properly chosen compensation capacitor.

At temperatures above and below the mid-range, additional compensation must be introduced. An externally generated compensation voltage is applied to a varactor (voltage-variable capacitor) which is in parallel with the crystal.

In transmitter types KT-56-A, C a constant bias of 5 Volts (provided from Regulator IC U901 in parallel with the compensator) establishes the varactor capacity at a constant value over the entire mid-temperature range. With no additional compensation, all of the oscillators will provide 2 PPM frequency stability from 0°C to 55°C (+32°F to 131°F).

Compensator Circuits

Both the 5C-ICOMs and 2C-ICOMs are temperature compensated at both ends of the temperature range to provide instant frequency compensation. An equivalent ICOM circuit is shown in Figure 3.

The cold end compensation circuit does not operate at temperatures above 0°C. When the temperature drops below 0°C, the circuit is activated. As the temperature decreases, the equivalent resistance decreases and the compensation voltage increases.

The increase in compensation voltage decreases the capacity of the varactor in the oscillator, increasing the output frequency of the ICOM.

The hot end compensation circuit does not operate at temperatures below +55°C. When the temperature rises above +55°C, the circuit is activated. As the temperature

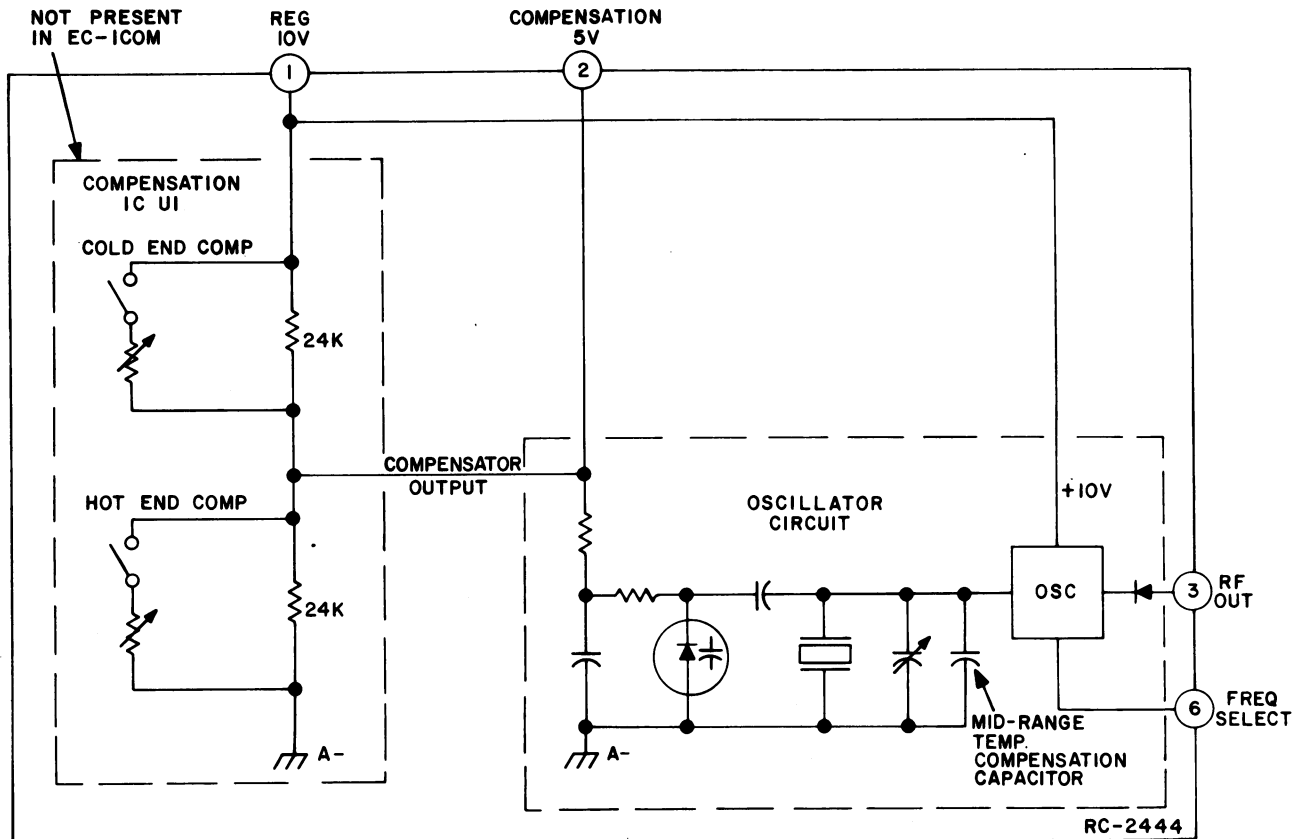


Figure 3 - Equivalent ICOM Circuit

increases, the equivalent resistance decreases and the compensation voltage decreases. The decrease in compensation voltage increases the capacity of the varactor, decreasing the output frequency of the ICOM.

SERVICE NOTE: Proper ICOM operation is dependent on the closely-controlled input voltages from the 10-Volt regulator. Should all of the ICOMs shift off frequency, check the 10-Volt regulator module or check output of 5C-ICOM.

AUDIO PREAMPLIFIER

The transmitter Audio Preamplifier is not part of the transmitter and is covered in the station Maintenance Manual.

AUDIO IC

The transmitter audio circuitry is contained in audio IC-U102. A simplified drawing of the audio IC is shown in Figure 4.

Audio from the Station Preamplifier at pin 12 is coupled through capacitor C1 to the base of Q1 in the operational amplifier-limiter circuit.

The operational amplifier-limiter circuit consists of Q1, Q2 and Q3. Q3 provides limiting at high signal levels. The gain of the operational amplifier circuit is fixed by negative feedback through R19, R20 and the resistance in the network (pin 9).

The output of Q3 is coupled through a de-emphasis network (R10 and C3) to an active post-limiter filter consisting of C4, C5, C6, R11, R12, R13, R15, R17 and Q4.

Following the post-limiter filter is class A amplifier Q5. The output of Q5 is coupled through MOD ADJUST potentiometer R127 to the phase modulators.

SERVICE NOTE: If the DC voltages to the Audio IC are correct and no audio output can be obtained, replace U102.

For radios equipped with Channel Guard, tone from the encoder is applied to the phase modulators through CHANNEL GUARD MOD ADJUST potentiometer R128, and resistors R110, R121 and R124. Instructions for setting R128 and station gain Control R14 on the Audio Preamplifier are contained in the modulation adjustment section of the Transmitter Alignment Procedure.

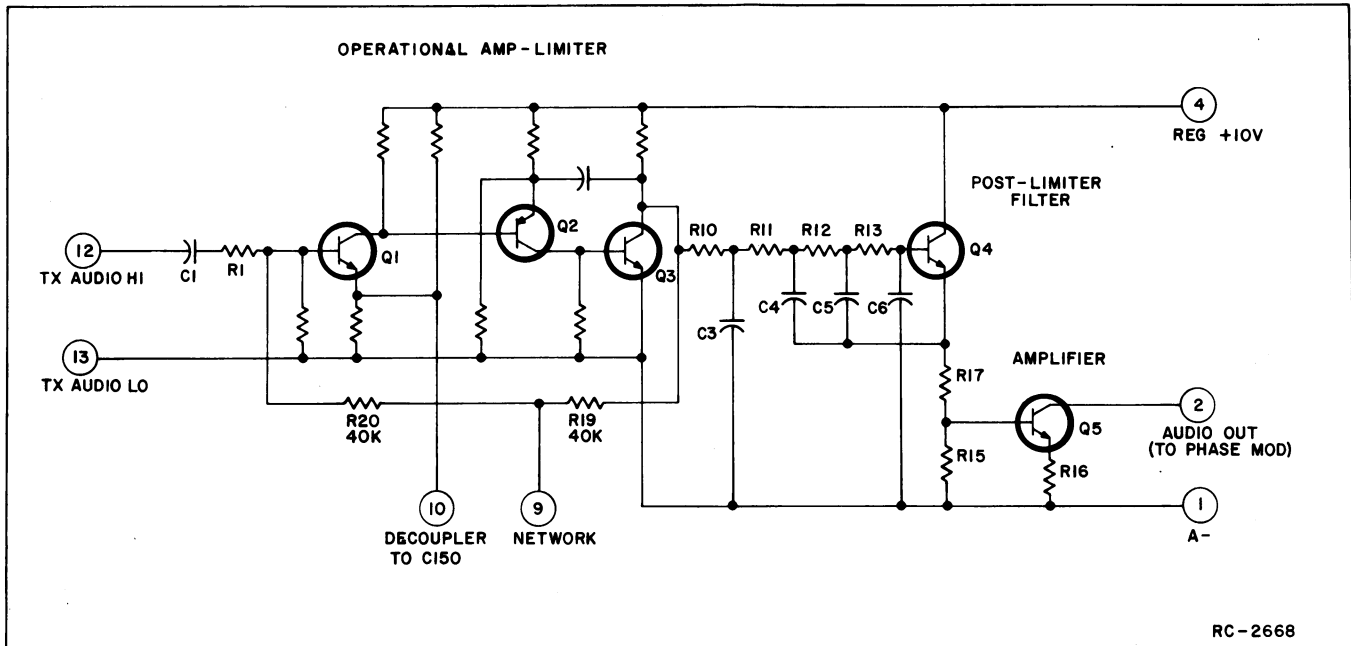


Figure 4 - Simplified Audio IC

FREQUENCY DIVIDER IC

The output at pin 3 of the selected ICOM is coupled through buffer amplifier Q101 to frequency divider U101, which divides the oscillator frequency by 4. The divider consists of two J-K flip-flops connected as a binary counter.

When the transmitter is not keyed (no ICOMs on), Q101 is saturated (turned on) with its collector voltage near zero. Keying the transmitter starts one of the ICOMs, and its output cuts Q101 on and off once each cycle. As Q101 turns off during each cycle, the drop in collector voltage causes the left flip-flop to change state. Assume the flip-flop was in the "0" state (the output at "Q" near A-). The first cycle of the oscillator output causes it to switch to the "1" stage (output at "Q" approximately 5 Volts). The second cycle will cause the flip-flop to switch back to the "0" state. Therefore, it requires two oscillator cycles to switch the left flip-flop through one complete cycle from "0" to "1" and back to "0".

When the left flip-flop switches from "1" to "0", it causes the right flip-flop to change state. It requires two cycles of the left flip-flop to switch the right flip-flop from "0" to "1" and back to "0". Therefore, four cycles of the oscillator output are required for each cycle of output from pin 9 of U101.

If U101 was operating into a pure resistive load, its output would be a square wave. However, the modulator circuit pre-

sents a tuned load to the IC, so that harmonics are filtered out and the waveform at the junction of C102 and C103 (modulator input) is essentially a sine wave at one-fourth the oscillator frequency. The output of the frequency divider is coupled through DC blocking capacitor C102 to the first modulator stage.

PHASE MODULATORS, AMPLIFIER & MULTIPLIERS

The first phase modulator is varactor (voltage-variable capacitor) CV101 in series with tunable coil L101. This network appears as a series-resonant circuit the RF output of the oscillator. An audio signal applied to the modulator circuit through blocking capacitor C115 varies the bias of CV101, resulting in a phase modulated output. A voltage divider network (R108 and R109) provides the proper bias for varactors CV101, CV102 and CV103.

The output of the first modulator is coupled through blocking capacitor C106 to the base of Class A amplifier Q102. The first modulator stage is metered through a metering network consisting of R115, R150, C107 and CR101. Diodes CR102 and CR103 remove any amplitude modulation in the modulator output.

Following Q102 is another Class A Amplifier, Q103. The output of Q103 is applied to the second modulator stage. The second modulator consists of two cascaded modulator circuits consisting of CV102, L102, L103 and CV103. Following the second modulator is a Class A amplifier Q104. The output of the second modulator stage is metered through

R133, R145, C117 and CR104, and is applied to the base of buffer Q105. Diodes CR105 and CR106 remove any amplitude modulation in the second modulator output.

Buffer Q105 is saturated when no RF signal is present. Applying an RF signal to Q105 provides a sawtooth waveform at its collector to drive the class C tripler, Q106. The tripler stage is metered through R146. The output of Q106 is coupled through tuned circuits T101, T102 and T103 to the base of doubler Q107. T101, T102 and T103 are tuned to one-fourth of the operating frequency. The doubler stage is metered through R147.

The output of Q107 is coupled through tuned circuits T104 and T105 to the base of second doubler Q108. T104 and T105 are tuned to one-half the operating frequency. Q108 is metered through R148.

The output of Q108 is coupled through three tuned circuits (T106, T107 and T108) to the base of amplifier Q109. The circuits are tuned to the transmitter operating frequency.

Q109 is a class C amplifier with a collector feed network consisting of C139, C141, L104, L108 and R143. The stage is metered through R149. The amplifier collector circuit consists of C142, C143, C146 and L105, and matches the amplifier output to the input of the power amplifier assembly.

POWER AMPLIFIER

The PA assembly uses three RF power transistors and seven transistors in the Power Control circuitry to provide a power output of 50 Watts. The broadband PA has no adjustments other than Power Control potentiometer R216.

Supply voltage for the PA is connected through power leads from the system board to feedthrough capacitors C297 and C298 on the bottom of the PA assembly. C297, C298 and C299, L296 and L297 prevent RF from getting on the Power leads. Diode CR295 will cause the main fuse in the fuse assembly to blow if the polarity of the power leads is reversed.

Centralized metering jack J205 is provided for use with GE Test Set Model 4EX3A11 or Test Kit 4EX8K12. The Test Set meters the Ampl-1 drive (exciter output), Ampl-1 power control, Driver and PA current.

RF AMPLIFIERS

The exciter output is coupled through an RF cable to PA input jack J203. The RF is coupled through DC blocking capacitor C202 to the base of Class C amplifier Q201 through a matching network. The network

matches the 50-ohm input to the base of Q201, and consists of C205, C206, C235, L201, L202 and L203.

Part of the RF input is rectified by CR201 and used to activate the Power Control circuit. Another portion of the rectified RF is applied to voltage dividers R223 and R224 for metering the Ampl-1 drive at J205.

Collector voltage to Q201 (Ampl-1) is controlled by the Power Control circuit, and is applied through a collector stabilizing network consisting of L224 and R225 and collector feed network L204 and C207. The collector voltage of Q201 is metered through R235 at J205.

The output of Q201 is applied to the base of Class C driver Q202 through a low-pass filter matching network (C209, C210, L205 and L206). Resistors R202, R203 and R204 lower the gain of Q202. Collector voltage to Q202 is coupled through a collector stabilizing network consisting of L225 and R233 and collector feed network L208 and C213.

Collector current for Q202 is metered across tapped manganin resistor R230 at J205 (Driver Current). The reading is taken on the one-Volt scale with the High Sensitivity button pressed, and read as 10 amperes full scale.

Following Q202 is an interstage coupling network (C214 through C221, L209 through L211, R206 and R207.) The output is applied to the base of the class C PA stage, Q203. Supply voltage is coupled through a collector stabilizing network consisting of L226 and R234 and collector feed network C222 and L212.

Collector current for Q203 is metered across tapped manganin resistor R231 at J205. The reading is taken on the one-Volt scale with the High Sensitivity button pressed, and read as 10 amperes full scale.

The PA output is coupled through an output matching network (C224, C225, C226, L213 and L214,) to an M-derived, constant K low-pass filter. C230 through C233 provides ground isolation. The filter output is applied to the antenna through antenna switch K201.

WARNING

The stud mounted RF Power Transistors used in the transmitter contain Beryllium Oxide, a TOXIC substance. If the ceramic, or other encapsulation is opened, crushed, broken or abraded, the dust may be hazardous if inhaled. Use care in replacing transistors of this type.

POWER CONTROL CIRCUIT

When the transmitter is keyed, rectified RF from CR201 is applied to the base of switch Q204, turning it on. Turning on Q204 turns on voltage regulator Q206 which supplies a constant voltage to Power Adjust potentiometer R216.

Q208, Q209 and Q210 operate as an amplifier chain to supply voltage to the collector of Q201 (Ampl-1). The setting of R216 determines the voltage applied to the base of Q208. The higher the voltage at the base of Q208, the harder the amplifiers conduct, supplying more collector voltage to Q201. The lower the voltage at the base of Q208, the less collector voltage is supplied to Q201. Reducing the supply voltage to Q201 reduces the drive to Q202 and Q203, thereby reducing the power output of the PA. The power output can be adjusted by R216 from approximately 15 to 50 Watts.

Temperature protection is provided by Q205, Q207 and thermistor RT201 which is mounted in the PA heatsink. Under normal operating conditions, the circuit is inactive (Q205 is on and Q207 is off). When the heatsink temperature reaches approximately 100°C, the resistance of RT201 decreases. This increases the base voltage applied to Q205, turning it off. Turning off Q205 allows Q207 to turn on, decreasing the voltage at Power Adjust potentiometer R216. This reduces the base voltage to Q208 which causes Q209 and Q210 to conduct less, reducing the collector voltage to Q201 (Ampl-1). This reduces the transmitter output power, keeping the heatsink at a maximum of approximately 100°C. When the heatsink temperature decreases below 100°C, the temperature control circuit turns off, allowing the normal transmitter power output.

MAINTENANCE

DISASSEMBLY

For a more complete mechanical parts breakdown refer to the station manual. To service the transmitter exciter from the front:

1. Turn the two latching knobs (A) counterclockwise to unlatch the Radio Panel Front Door. Refer to Figure 5.
2. Swing the Radio Panel Front Door down as shown.
3. Remove covers.

To service the transmitter Power Amplifier from the rear:

1. Remove the top two screws (B) on the Intermittent or Continuous Duty Power Amplifier. Refer to Figure 6.

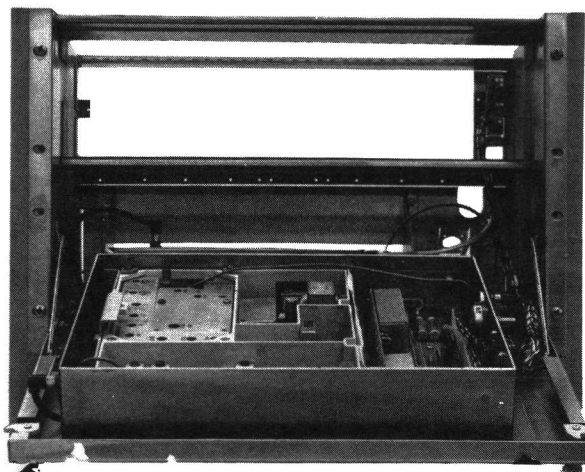


Figure 5 - Access To Exciter - Front View

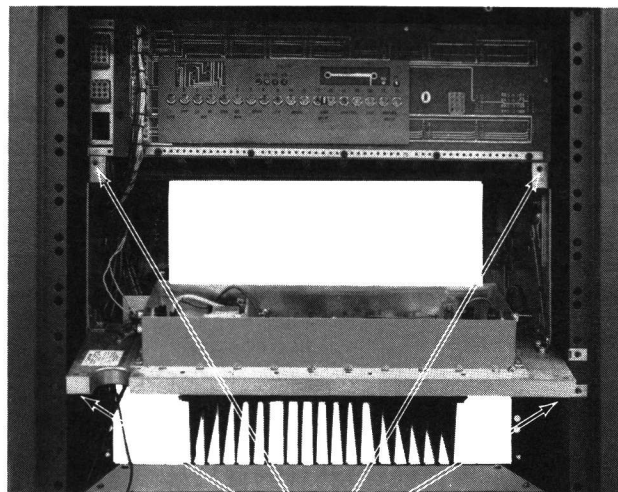


Figure 6 - Access To Power Amplifier - Rear View

2. Swing the Power Amplifier down as shown. Remove the top cover of the Power Amplifier.

NOTE

If the heatsink blower option is present, this blower must be removed before the Power Amplifier can be lowered.

To remove the PA board: Refer to Figure 7.

1. Remove the PA top cover and unplug the exciter/PA cable (B).
2. Unsolder the two feedthrough coils (E) and the thermistor leads (F).
3. For Continuous Duty stations only, remove all heatsink sections from the heat dissipator plate.
4. Remove the PA transistor hold-down nuts and spring washer on the bottom of the PA assembly.
5. Remove the four PA board mounting screws (G), the five screws in the filter casting (H), and the retaining screw in Q210 (J), and lift the board out.

PA TRANSISTOR REPLACEMENT

When replacing a power transistor where more than one are in parallel, make sure all the paralleled transistors are from the same manufacturer for proper operation.

WARNING

The stud mounted RF Power Transistors used in the transmitter contain Beryllium Oxide, a TOXIC substance. If the ceramic or other encapsulation is opened, crushed, broken or abraded, the dust may be hazardous if inhaled. Use care in replacing transistors of this type.

To replace the PA RF transistors:

1. Unsolder one lead at a time with a 50-Watt soldering iron. Use a scribe to hold the lead away from the printed circuit board until the solder cools.
2. Turn the transmitter.

NOTE

If the transmitter has a continuous Duty Power Amplifier a section of Heat Sink may have to be removed to get to the transistor hold-down nuts. Apply a light coat of silicon grease when replacing the removed section of Heat Sink.

3. Hold the body of the transistor to prevent it from turning. Remove the transistor hold-down nut and spring washer through the hole in the heatsink with an 11/32-inch nut-driver. Lift out the old solder from the printed circuit board with a de-soldering tool such as a SOLDA PULLT®. Special care should be taken to prevent damage to the printed circuit board runs.

4. Trim the new transistor leads (if required) to the lead length of the removed transistor. Cut the collector lead at a 45° angle for future identification (see Figure 7). The letter "C" on the top of the transistor indicates the collector.

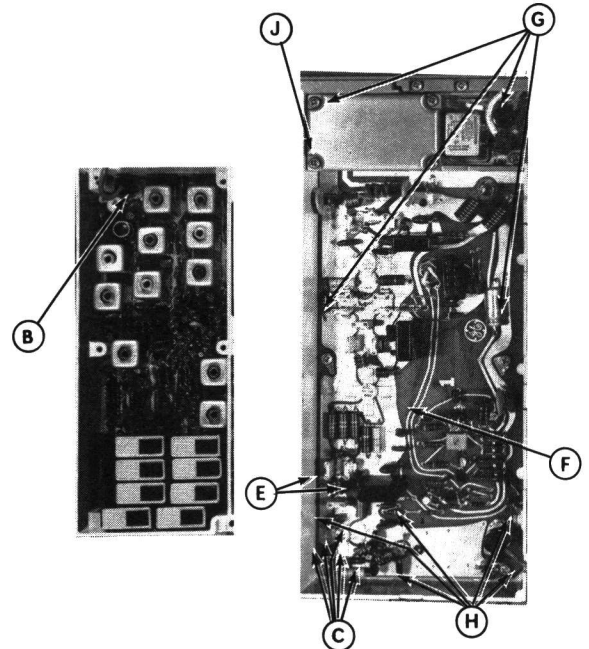


Figure 7 - PA Board Removal

5. Applying a coating of silicon grease around the transistor mounting surface, and place the transistor in the mounting hole. Align the leads as shown in the Outline Diagram. Then hold the body of the transistor and replace the holding-down nut and spring-washer, using moderate torque (8 inch-pounds). A torque wrench must be used for this adjustment since transistor damage can result if too little or too much torque is used.
6. Make sure that the transistor leads are formed as shown in Figure 8 so that the leads can be soldered to the printed circuit pattern, starting from the inner edge of the mounting hole.
7. Solder the leads to the printed circuit pattern. Start at the inner edge of mounting hole and solder the remaining length of transistor lead to the board. Use care not to use excessive heat that causes the printed wire board runs to lift up from the board. Check for shorts and solder bridges before applying power.

CAUTION

Failure to solder the transistor leads as directed may result in the generation of RF loops that could damage the transistor or may cause low power output.

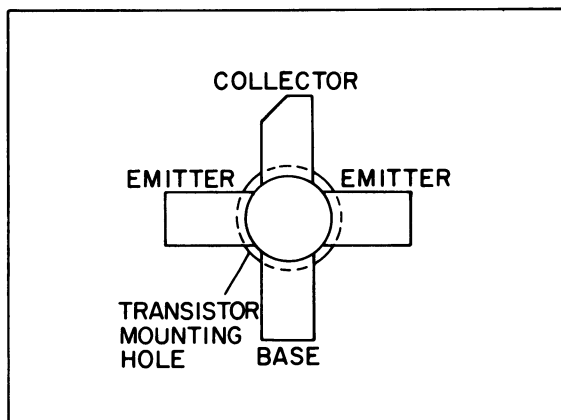


Figure 8 - Lead Identification

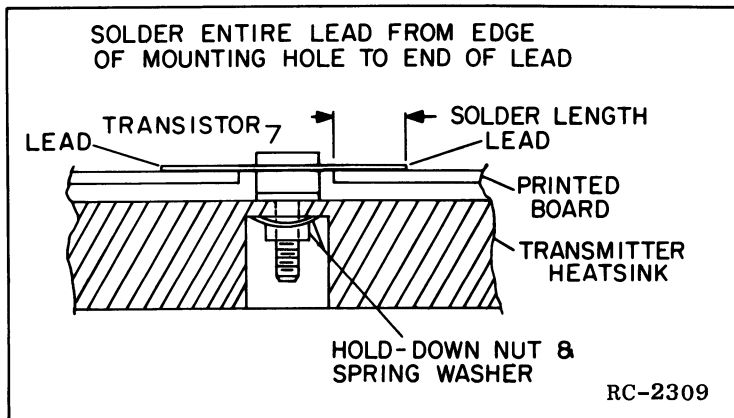


Figure 9 - Lead Forming

**GE Mobile Communications**

General Electric Company
Lynchburg, Virginia 24502

Printed in U.S.A.

MODULATION LEVEL ADJUSTMENT

The MOD ADJUST (R127) was adjusted to the proper setting before shipment and should not normally require readjustment. This setting permits approximately 75% modulation for the average voice level. The audio peaks which would cause overmodulation are clipped by the modulation limiter. The limiter in conjunction with the de-emphasis network, instantaneously limits the slope of the audio wave to the modulator, thereby preventing overmodulation while preserving intelligibility.

TEST EQUIPMENT

- 1. An audio oscillator (GE Model 4EX6A10).
- 2. A frequency modulation monitor.
- 3. An output meter or a VTVM.
- 4. GE Test Set Model 4EX3A11 or 4EX8K12.

PROCEDURE

- 1. Set the station gain control R14 to its fully clockwise position.
- 2. Connect the audio oscillator and the meter through a 0.5 µF (or larger) DC blocking capacitor, across audio input terminals J10 (Green-Hi) and J11 (Black-Lo) on GE Test Set, and connect the Red Test set plug to the system Red metering plug.
- 3. Set the audio generator frequency to 1 kHz.
 - A. In all station combinations except Local Control Intermittent Duty combinations, set the audio generator output to 30 millivolts RMS.
 - B. In Local Control Intermittent Duty station combinations, set the audio generator output to 1.0 volt RMS.
- 4. For transmitters without Channel Guard, set MOD ADJUST R127 for a 4.5 kHz swing with the deviation polarity which gives the highest reading as indicated on the frequency modulation monitor.
- 5. For transmitters with Channel Guard set Channel Guard MOD ADJUST R105 for zero tone deviation. Next, with the 30 millivolts signal at 1000 Hz applied, set MOD ADJUST R104 for 3.75 kHz deviation. Then remove the signal from the audio oscillator and set Channel Guard MOD ADJUST R128 for 0.75 kHz tone deviation.
- 6. For multi-frequency transmitters, set the deviation as described in Steps 4 or 5 on the channel producing the largest amount of deviation.
- 7. Remove the audio oscillator and key the mike. While talking in a normal voice at a distance of four to six inches from the station microphone, adjust station gain control R14 for a deviation of 3 kHz as measured on the deviation monitor.

PA POWER INPUT

For FCC purposes, the PA power input can be determined by measuring the PA supply voltage and PA current, and using the following formula:

P_i = PA voltage x PA current

where:

P_i is the power input in Watts,
PA voltage is measured with Test Set Model 4EX3A11 in Position G on the 15-Volt range (read as 15 Volts full scale), and with the polarity switch in the (-) position. With Test Set Model 4EX8K12, use the B+ position and the 1-Volt range (read as 15 Volts full scale), with the HIGH SENSITIVITY button pressed and the polarity switch in the (-) position.

PA current is measured with the Test Set in Position G in the Test 1 position, and with the HIGH SENSITIVITY button pressed (10 amperes full scale).

Example:

P_i = 12.6 Volts x 5.0 amperes = 63 Watts.

ICOM FREQUENCY ADJUSTMENT

First, check the frequency to determine if any adjustment is required. The frequency should be set with a frequency meter or counter with an absolute accuracy that is 5 to 10 times better than the tolerance to be maintained, and with the entire radio as near as possible to an ambient temperature of 26.5°C (79.8°F).

MASTR II ICOMs should be reset only when the frequency shows deviations in excess of the following limits:

- A. ±0.5 PPM, when the radio is at 26.5°C (79.8°F).
- B. ±2 PPM at any other temperature within the range of -5°C to +55°C (+23°F to +131°F).
- C. The specification limit (±2 PPM or ±5 PPM) at any temperature within the ranges of -40°C to -5°C (-40°F to +23°F) or +55°C to +70°C (+131°F to +158°F).

If an adjustment is required, pry up the cover on the top of the ICOM to expose the trimmer, and use one of the following procedures:

If the radio is at an ambient temperature of 26.5°C (79.8°F), set the oscillator for the correct operating frequency.

If the radio is not at an ambient temperature of 26.5°C, setting errors can be minimized as follows:

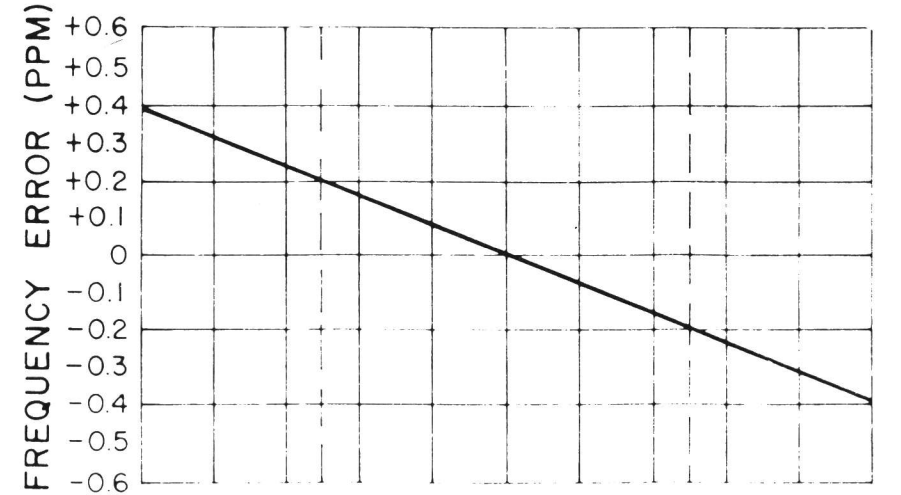
- A. To hold setting error to ±0.6 PPM (which is considered reasonable for 5 PPM ICOMs):
 - 1. Maintain the radio at 26.5°C (±5°C) and set the oscillator to desired frequency, or-
 - 2. Maintain the radio at 26.5°C (±10°C) and offset the oscillator, as a function of actual temperature, by the amount shown in Figure 9.
- B. To hold setting error to ±0.35 PPM (which is considered reasonable for 2 PPM ICOMs): Maintain unit at 26.5°C (±5°C) and offset the oscillator, as a function of actual temperature, by the amount shown in Figure 9.

For example: Assume the ambient temperature of the radio is 18.5°C (65.4°F). At that temperature, the curve shows a correction factor of 0.3 PPM. (At 25 MHz, 1 PPM is 25 Hz. At 50 MHz, 1 PPM is 50 Hz).

With an operating frequency of 50 MHz, set the oscillator for a reading of 15 Hz (0.3 x 50 Hz) higher than the licensed operating frequency. If a negative correction factor is obtained (at temperatures above 26.5°C), set the oscillator for the indicated PPM lower than the licensed operating frequency.

DEGREES FAHRENHEIT

61.8 65.4 69.0 72.6 76.6 79.8 83.4 87.0 90.6 94.2 97.8



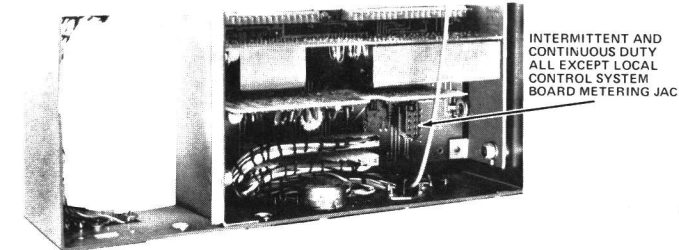
16.5 18.5 20.5 22.5 24.5 26.5 28.5 30.5 32.5 34.5 36.5

-5° LIMIT REF +5° LIMIT

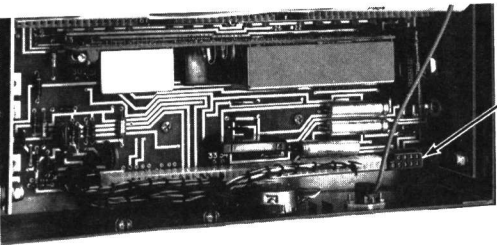
DEGREES CENTIGRADE

RC-2453

Figure 9 - Frequency Characteristics Vs. Temperature



INTERMITTENT AND CONTINUOUS DUTY ALL EXCEPT LOCAL CONTROL SYSTEM BOARD METERING JACK



INTERMITTENT DUTY LOCAL CONTROL SYSTEM BOARD METERING JACK

INTERMITTENT DUTY PA ANTENNA JACK J201

CONTINUOUS DUTY PA ANTENNA JACK J243

EXCITER METERING JACK J103

MOD ADJUST R127

CHANNEL GUARD MOD ADJUST R128

EXCITER ASSEMBLY

POWER ADJUST R216

PA ASSEMBLY

PA METERING JACK J205

TRANSMITTER ALIGNMENT

LBI-4732

EQUIPMENT REQUIRED

- 1. GE Test Set Model 4EX3A11 or Test Kit 4EX8K12.
- 2. A 50-ohm wattmeter connected to antenna jack J906.
- 3. A frequency counter.

PRELIMINARY CHECKS AND ADJUSTMENTS

- 1. Place ICOMs on Exciter Board (crystal frequency = operating frequency ÷ 3).
- 2. For a large change in frequency or a badly mis-aligned transmitter, pre-set the slugs to T101 through T108, and L101, L102 and L103 to the bottom of the coil form.

NOTE

The tuning frequency for multi-frequency transmitters is determined by the operating frequency and the frequency spread between frequencies. Refer to the table below for maximum frequency spread.

- 3. For multi-frequency transmitters with a frequency spread less than that specified in column (1), tune the transmitters to the lowest frequency.

For frequency spread exceeding the limits specified in column (1), tune the transmitter using a center frequency tune up ICOM. Except the maximum frequency spread can be extended to the limits specified in column (3) with 1 dB degradation.

For tuning L101, L102, L103, always tune L101, L102, L103 on the lowest frequency.

Multi-frequency Transmitter Tuning

Transmitter Frequency Range	MAXIMUM FREQUENCY SPREAD		
	(1) without center tuning	with center tuning	with center tuning (1 dB degradation)
25-30 MHz	.080 MHz	.160 MHz	.320 MHz
30-36 MHz	.100 MHz	.200 MHz	.400 MHz
36-42 MHz	.120 MHz	.240 MHz	.470 MHz
42-50 MHz	.140 MHz	.280 MHz	.540 MHz

- 4. Connect the red plug on the GE Test Set to the System Board metering jack, and the black plug to the Exciter metering jack. Set the polarity to +, and set the range to the Test 1 position (1-Volt position for 4EX8K12) for all adjustments. NOTE: With the Test Set connected to the PA metering jack, the voltage reading at position "F" with the HIGH SENSITIVITY button pressed may be converted to driver collector current by reading the current as 10 amperes full scale. The voltage reading at position "G" with the HIGH SENSITIVITY button pressed may be converted to PA collector current by reading the current as 10 amperes full scale.

- 5. All adjustments are made with the transmitter keyed. Unkey the transmitter between steps to avoid unnecessary heating.

STEP	METER POSITION	TUNING CONTROL	METER READING	PROCEDURE
1.	A MOD-1	L101	Maximum	Tune L101 for maximum meter reading.
2.	B	L102 & L103	Maximum	Tune L102 and then L103 for the maximum meter reading.
3.	C MULT-1	T101 & T102	See Procedure	Tune T101 for a dip in meter reading, and then tune T102 for maximum meter reading.
4.	D MULT-2	T103, T102, T101 & T104	See Procedure	Tune T103 for maximum meter reading and re-adjust T102 and T101 for maximum meter reading. Then tune T104 for a dip in meter reading.
5.	F MULT-3	T105, T104, T106 & T107	See Procedure	Tune T105 for maximum meter reading and re-adjust T104 for maximum meter reading. Then tune T106 for a dip in meter reading and T107 for maximum meter reading.
6.	G AMPL-1	T108, T107 & T106	Maximum	Tune T108 for maximum meter reading, and then re-adjust T107 and T106 for maximum meter reading.
7.	D AMPL-1 DRIVE (on PA)	C143, C156	Maximum	Move the black metering plug to the Power Amplifier metering jack and tune C143 and C156 for maximum meter reading.
8.		R216		Set Power Adjust potentiometer R216 on the PA board for the desired power output (from 15 to 50 Watts).

ALIGNMENT PROCEDURE

25—50 MHz, 50-WATT STATION TRANSMITTER
INTERMITTENT & CONTINUOUS DUTY

TEST PROCEDURES

These Test Procedures are designed to assist you in servicing a transmitter that is operating--but not properly. Problems encountered could be low power output, tone and voice deviation, defective audio sensitivity, and modulator adjust control set too high. Once a de-

fect is pin-pointed, refer to the "Service Check" and the additional corrective measures included in the Transmitter Troubleshooting Procedure. Before starting with the Transmitter Test Procedures, be sure the transmitter is tuned and aligned to the proper operating frequency.

TEST EQUIPMENT REQUIRED

for test hookup as shown:

- | | | |
|--|---|--|
| 1. Wattmeter similar to:
Bird # 43
Jones # 711N | 2. VTVM similar to:
Triplet # 850
Heath # IM-21 | 3. Audio Generator similar to:
GE Model 4EX6A10 |
| 4. Deviation Meter (with a .75 kHz scale) similar to: Measurements # 720 | 5. Multimeter similar to:
GE TEST SET MODEL 4EX3A11,
MODEL 4EX8K12 or
20,000 ohms-per-Volt voltmeter | |

POWER MEASUREMENT

TEST PROCEDURE

1. Connect transmitter output from the antenna jack to the wattmeter through a 50-ohm coaxial cable. Make sure the wattmeter is terminated into a 50-ohm load.
2. Key the transmitter and check the wattmeter for the desired power output.

SERVICE CHECK

Check the setting of the Power Adjust Control (R216).

Refer to the QUICK CHECKS on the Transmitter Troubleshooting Procedure.

VOICE DEVIATION AND SYMMETRY

TEST PROCEDURE

1. Connect the test equipment to the transmitter as shown.
2. In radios with Channel Guard, set Channel Guard Mod Adjust R128 for zero tone deviation.
3. Set the audio generator frequency to 1 kHz.
 - A. In all station combinations except Local Control Intermittent Duty combinations, set the audio generator output to 30 millivolts RMS.
 - B. In Local Control Intermittent Duty station combinations, set the audio generator output to 1.0 volt RMS.
4. Key the transmitter and adjust Deviation Meter to carrier frequency.
5. Deviation reading should be ± 4.5 kHz in radios without Channel Guard, and ± 3.75 kHz in radios with Channel Guard.
6. If necessary, adjust MOD ADJUST control R127 for the proper deviation on plus (+) or minus (-) deviation, whichever is greater.

NOTES: MASTR II station transmitters are adjusted for 4.5 kHz deviation at the factory. The factory adjustment will prevent the transmitter from deviating more than 5.0 kHz under the worst conditions of frequency, voltage and temperature.

7. If the deviation reading plus (+) or minus (-) differs by more than 0.5 kHz, recheck Steps 1 and 2 as shown in the Transmitter Alignment Chart.

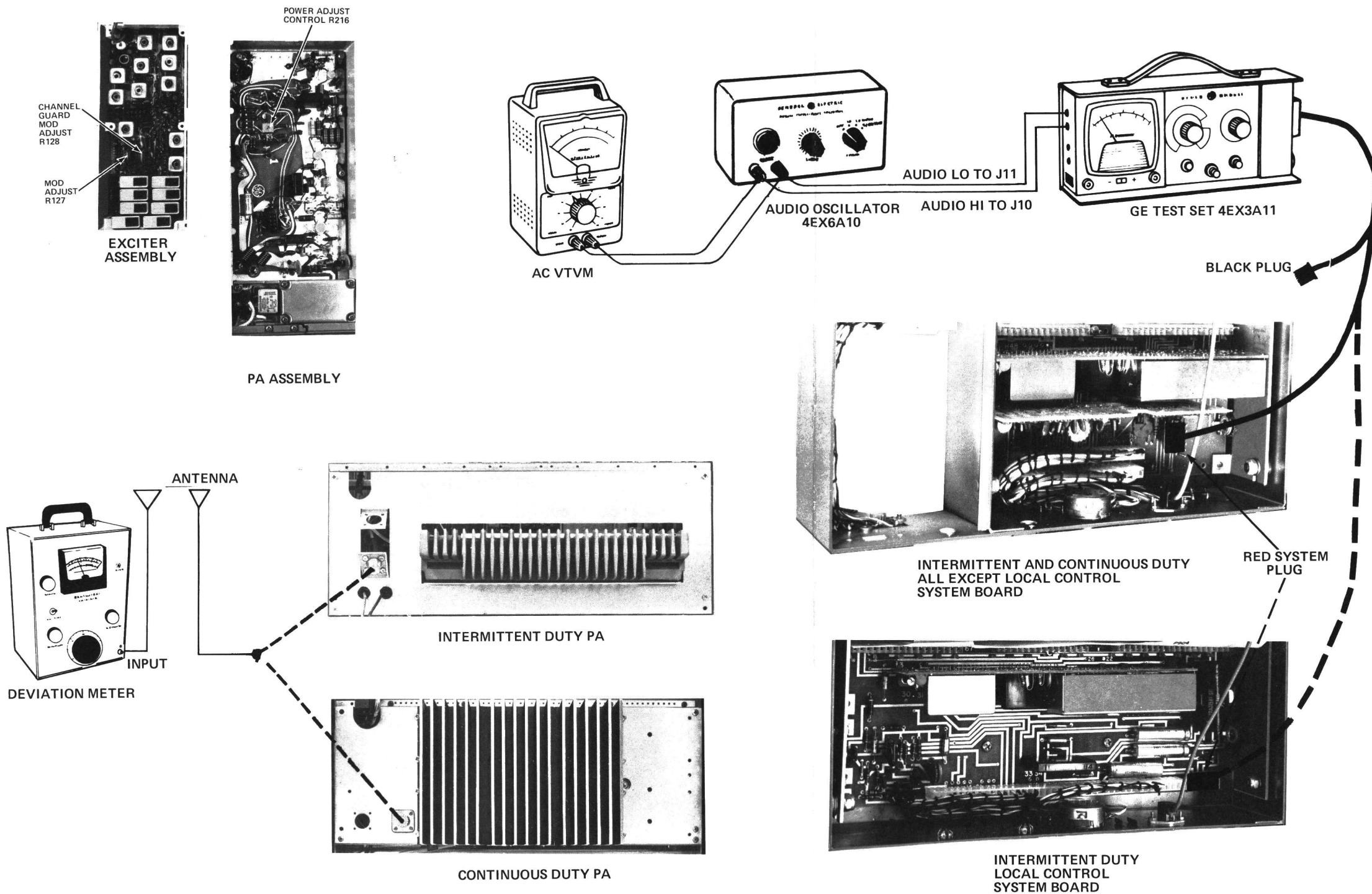
TONE DEVIATION WITH CHANNEL GUARD

TEST PROCEDURE

1. Set up the Deviation Meter and monitor the output of the transmitter.
2. Remove the 1000 Hz signal from the audio generator.
3. Key the transmitter and check for 0.75 kHz deviation. If the reading is low or high, adjust Channel Guard MOD ADJUST R128 for a reading of 0.75 kHz.

NOTES: --

1. On units supplied with Channel Guard, the Phase Modulator Tuning should be adjusted carefully to insure proper performance. (Refer to Step 1 in the Transmitter Alignment Chart).
2. The Tone Deviation Test Procedures should be repeated every time the Tone Frequency is changed.

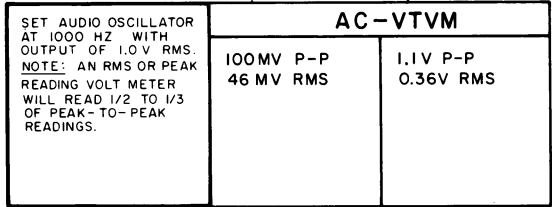


STEP I - QUICK CHECKS

METER POSITION GE TEST SET	PROBABLE DEFECTIVE STAGE		
	HIGH METER READING	LOW METER READING	ZERO METER READING
EXCITER			
A (MOD-1)	Q102, 10-Volt regulator	Q102, CV101, L101, 10-Volt regulator	ICOM, Q101, U101, L101, Q102, CR101, 10-Volt Selector switch ground.
B (MOD-2)	Q104, 10-Volt regulator	Q103, L102, CV103, CV104	Q103, L102, CV102, L103, CV103, CR104, Q104
C (MULT-1)	Q105, Q106 T101	Q105, Q106	Q105, Q106, T101
D (MULT-2)	Q107, T104	T101, T102, T103, Q107	T101, T102, T103, Q107, T104
F (MULT-3)	Q108, T106	T104, T105, Q108	T104, T105, Q108, T106
G (AMPL-1)	Q109, C146, R144	T106, T107, T108, Q109, L108	T106, T107, T108, Q109, L104, L107
POWER AMPLIFIER			
"D" (AMPL-1 DRIVE)		Low Output from Exciter	No output from Exciter, CR201
"C" (AMPL-1 POWER CONTROL VOLT-AGE)	Q210	Q210	No Exciter output, Q210, Q204, CR201
"F" (DRIVER CURRENT)	Q202	Q202, Low Output from Q201	Q202, Q201, Check Pos. D & C
"G" (PA CURRENT)	Q203	Q201, Q202, Q203	Q203, Q202, Q201, Q210

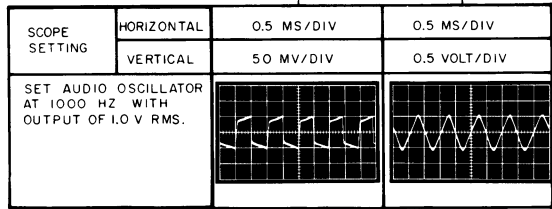
STEP 3
CHECK AUDIO AC VOLTAGES

EQUIPMENT REQUIRED
● AUDIO OSCILLATOR
● AC VTVM



STEP 4
AUDIO & OSC WAVEFORMS

EQUIPMENT REQUIRED
● AUDIO OSCILLATOR
● OSCILLOSCOPE



STEP 2
CHECK TYPICAL DC VOLTAGES

EQUIPMENT REQUIRED
● G.E. TEST MODEL 4EX3A11 OR
● 20,000 OHM-PER-VOLT METER

NOTE: ALL DC READINGS TAKEN WITH THE TRANSMITTER KEYED.

V-DC
TYPICAL MOD-1 READING AT POS. A SHOULD BE: 0.5V

V-DC
TYPICAL MOD-2 READING AT POS. B SHOULD BE: Q.5V

V-DC
TYPICAL MULT-1 READING AT POS. C SHOULD BE: 0.8V

V-DC
TYPICAL MULT-2 READING AT POS. D SHOULD BE: 0.56V

V-DC
TYPICAL MULT-3 READING AT POS. F SHOULD BE: 0.45V

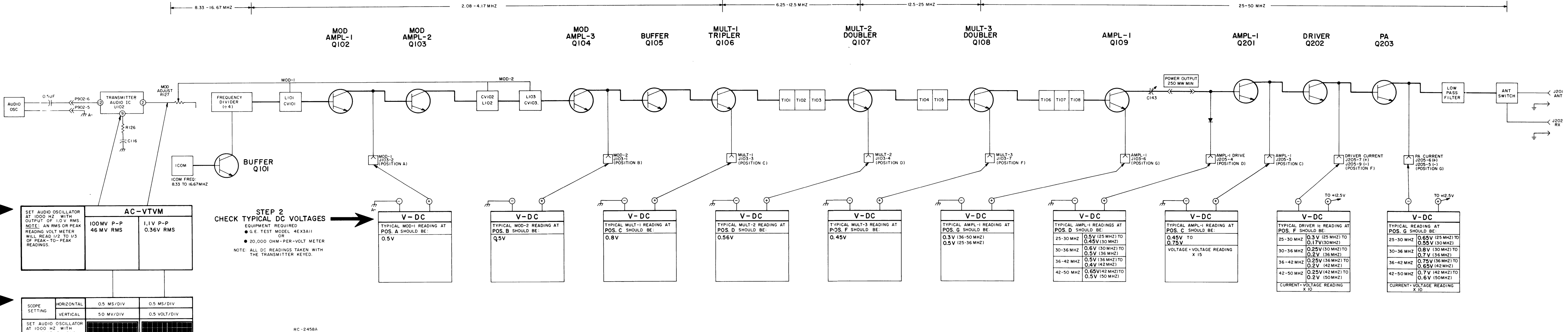
V-DC
TYPICAL AMPL-1 READING AT POS. G SHOULD BE: 0.3V (36-50 MHZ) 0.5V (25-36 MHZ)

V-DC
TYPICAL AMPL-1 READINGS AT POS. D SHOULD BE: 25-30 MHZ 0.5V (25 MHZ) TO 0.45V (30 MHZ) 30-36 MHZ 0.6V (30 MHZ) TO 0.5V (36 MHZ) 36-42 MHZ 0.25V (36 MHZ) TO 0.2V (42 MHZ) 42-50 MHZ 0.65V (42 MHZ) TO 0.5V (50 MHZ)

V-DC
TYPICAL AMPL-1 READING AT POS. C SHOULD BE: 0.45V TO 0.75V VOLTAGE * VOLTAGE READING X 15

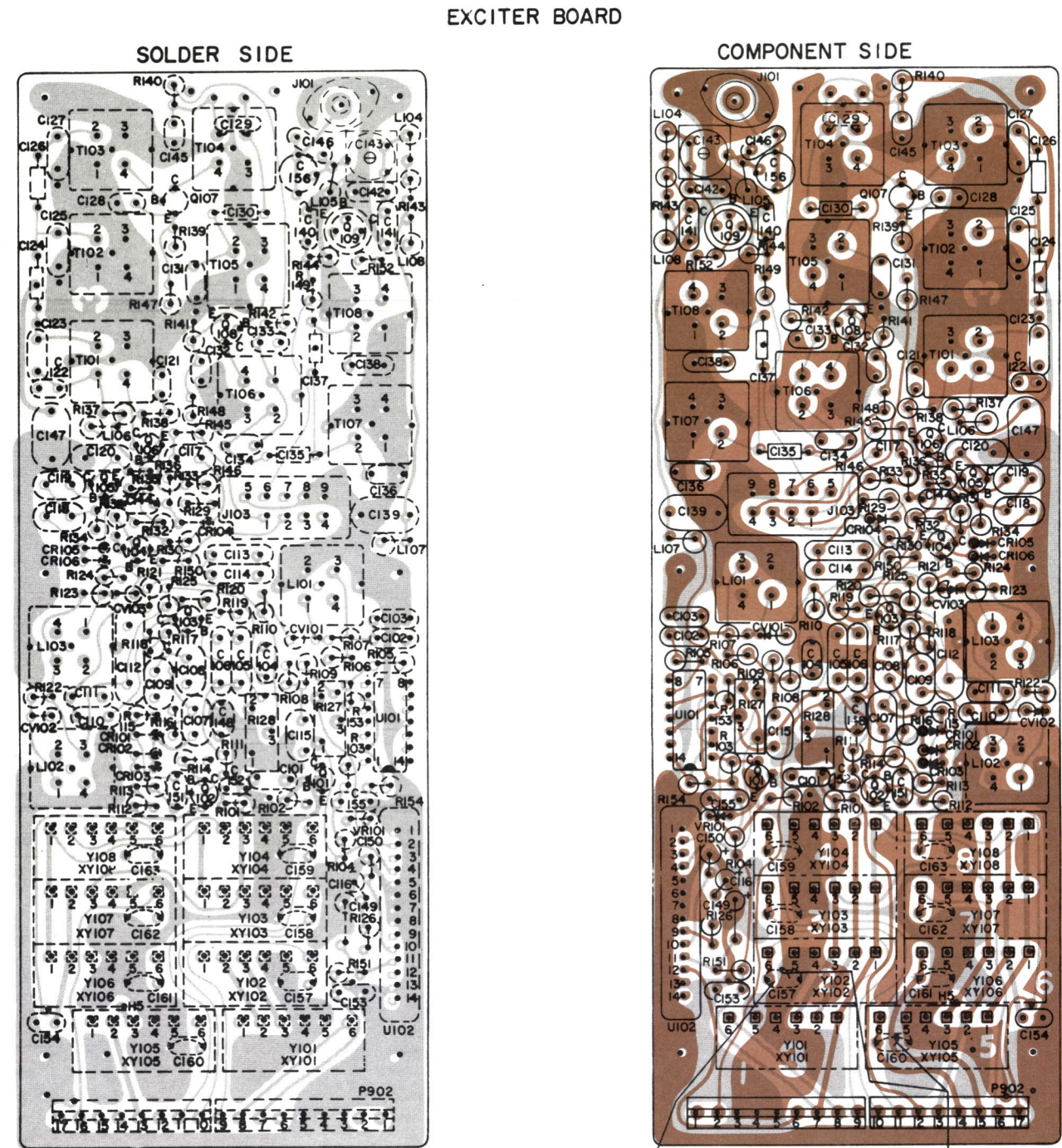
V-DC
TYPICAL DRIVER IC READING AT POS. F SHOULD BE: 25-30 MHZ 0.3V (25 MHZ) TO 0.17V (30 MHZ) 30-36 MHZ 0.25V (30 MHZ) TO 0.2V (36 MHZ) 36-42 MHZ 0.25V (36 MHZ) TO 0.2V (42 MHZ) 42-50 MHZ 0.25V (42 MHZ) TO 0.2V (50 MHZ) CURRENT * VOLTAGE READING X 10

V-DC
TYPICAL READING AT POS. G SHOULD BE: 25-30 MHZ 0.65V (25 MHZ) TO 0.55V (30 MHZ) 30-36 MHZ 0.8V (30 MHZ) TO 0.7V (36 MHZ) 36-42 MHZ 0.75V (36 MHZ) TO 0.65V (42 MHZ) 42-50 MHZ 0.7V (42 MHZ) TO 0.6V (50 MHZ) CURRENT * VOLTAGE READING X 10



TROUBLESHOOTING PROCEDURE

25—50 MHz, 50-WATT STATION TRANSMITTER
INTERMITTENT & CONTINUOUS DUTY



LEAD IDENTIFICATION FOR Q101-Q109

IN EIGHT-FREQUENCY EXCITERS (GROUPS 5-8), CAPACITORS C157-C165 ARE CLIPPED OUT AS REQUIRED TO MEET THE CUSTOMER REQUIREMENTS FOR FREQUENCIES. EXAMPLE: IF CUSTOMER WANTS 100K FOR F1, F2, F5, F7, THEN CAPACITORS C157, C160, AND C162 ARE CLIPPED OUT. C158, C159, C161, C163 ARE LEFT IN.

IN TWO-FREQUENCY EXCITERS (GROUPS 1-4) C157 IS CLIPPED OUT FOR COMBINATIONS WITH 2 TRANSMIT ICONS. DA JUMPERS ARE PRESENT ON FREQUENCY SWITCHING LINES OF OTHER SIX ICON CIRCUITS AS SHOWN.

LEAD IDENTIFICATION FOR Q204-Q209

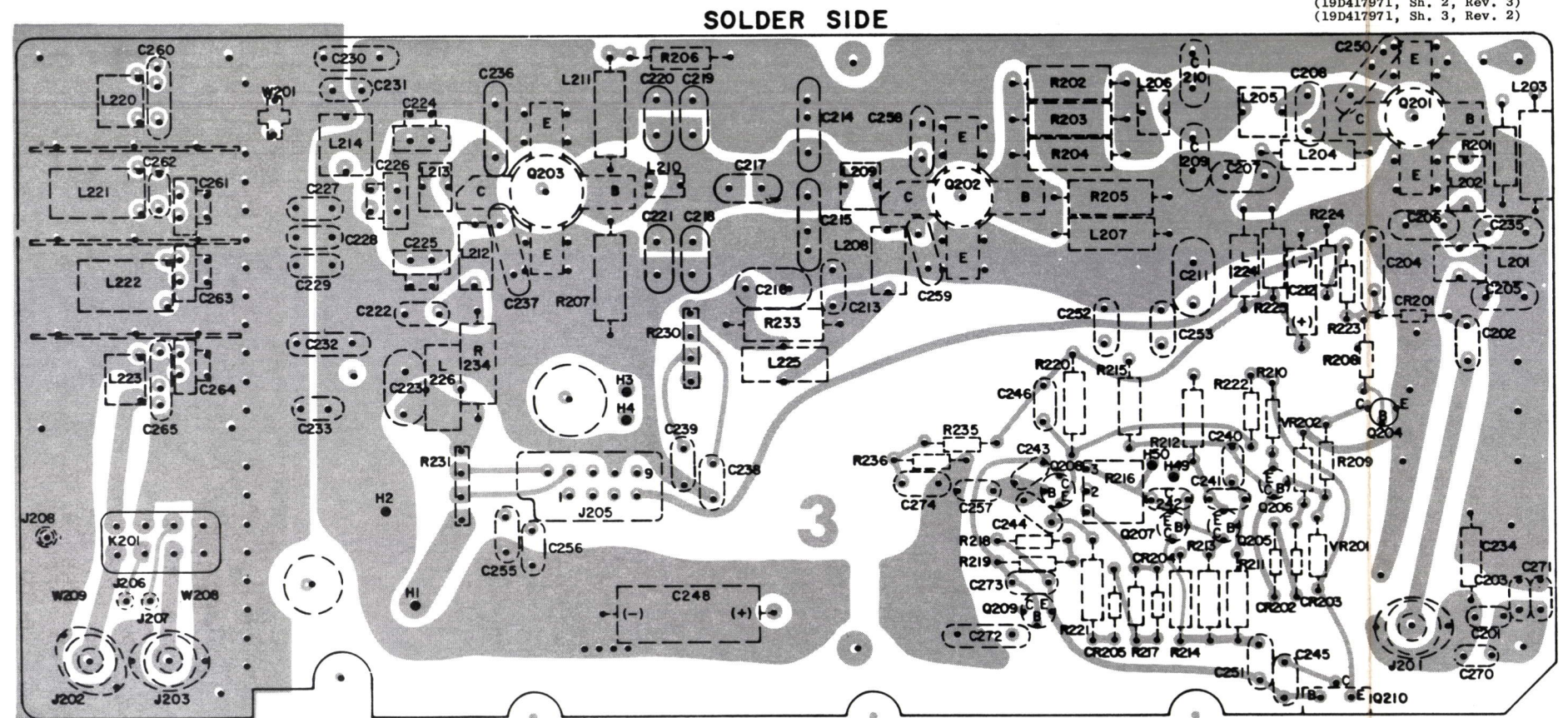
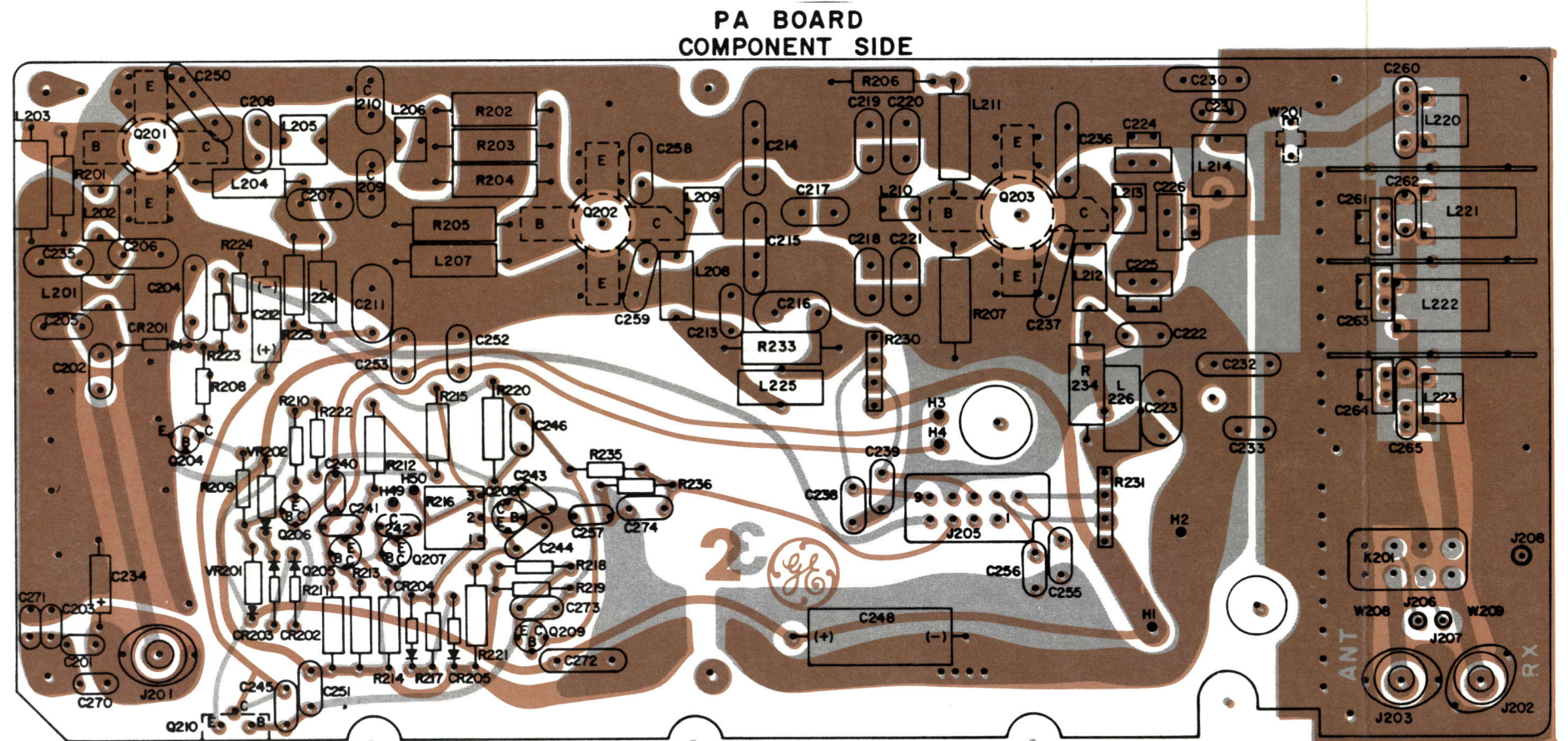
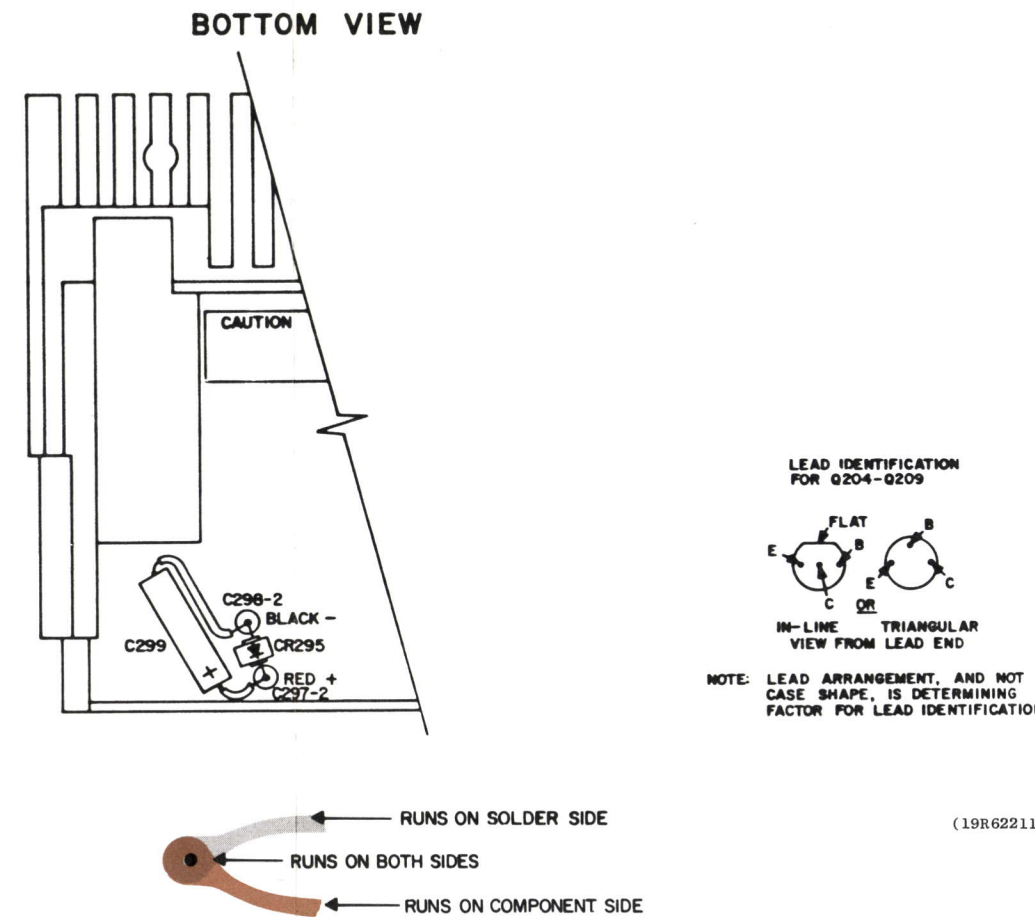
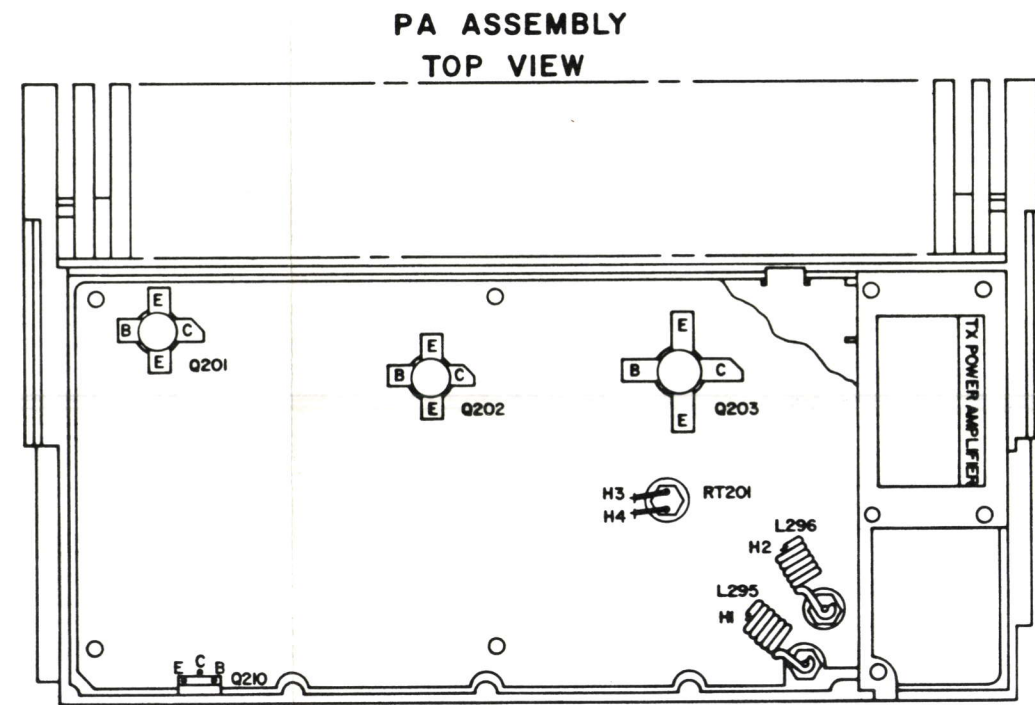
IN-FLAT VIEW FROM LEAD END

OR

IN-FLAT TRIANGULAR VIEW FROM LEAD END

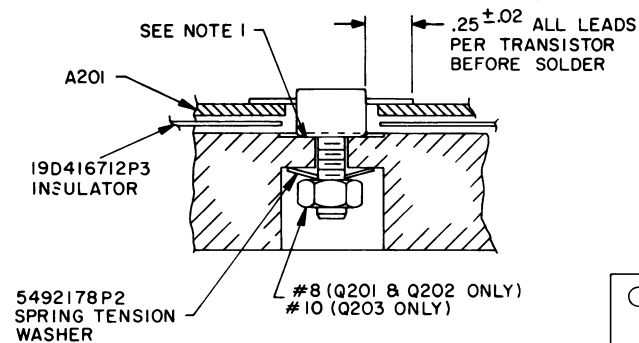
NOTE: LEAD ARRANGEMENT, AND NOT CASE SHAPE, IS DETERMINING FACTOR FOR LEAD IDENTIFICATION.

(19D423234, Rev. 3)

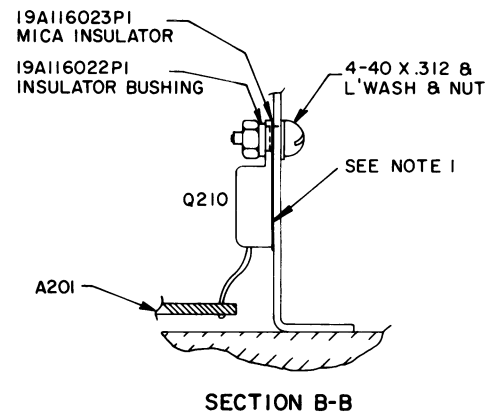


OUTLINE DIAGRAM

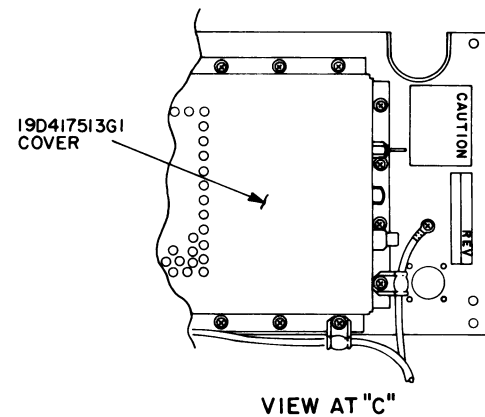
25—50 MHz, 50-WATT STATION TRANSMITTER



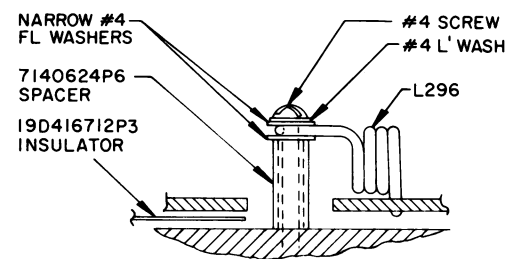
SECTION A-A
TYP MTG FOR
Q201 THRU Q203



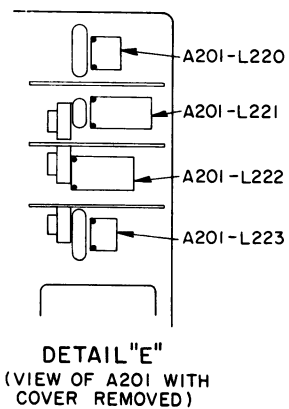
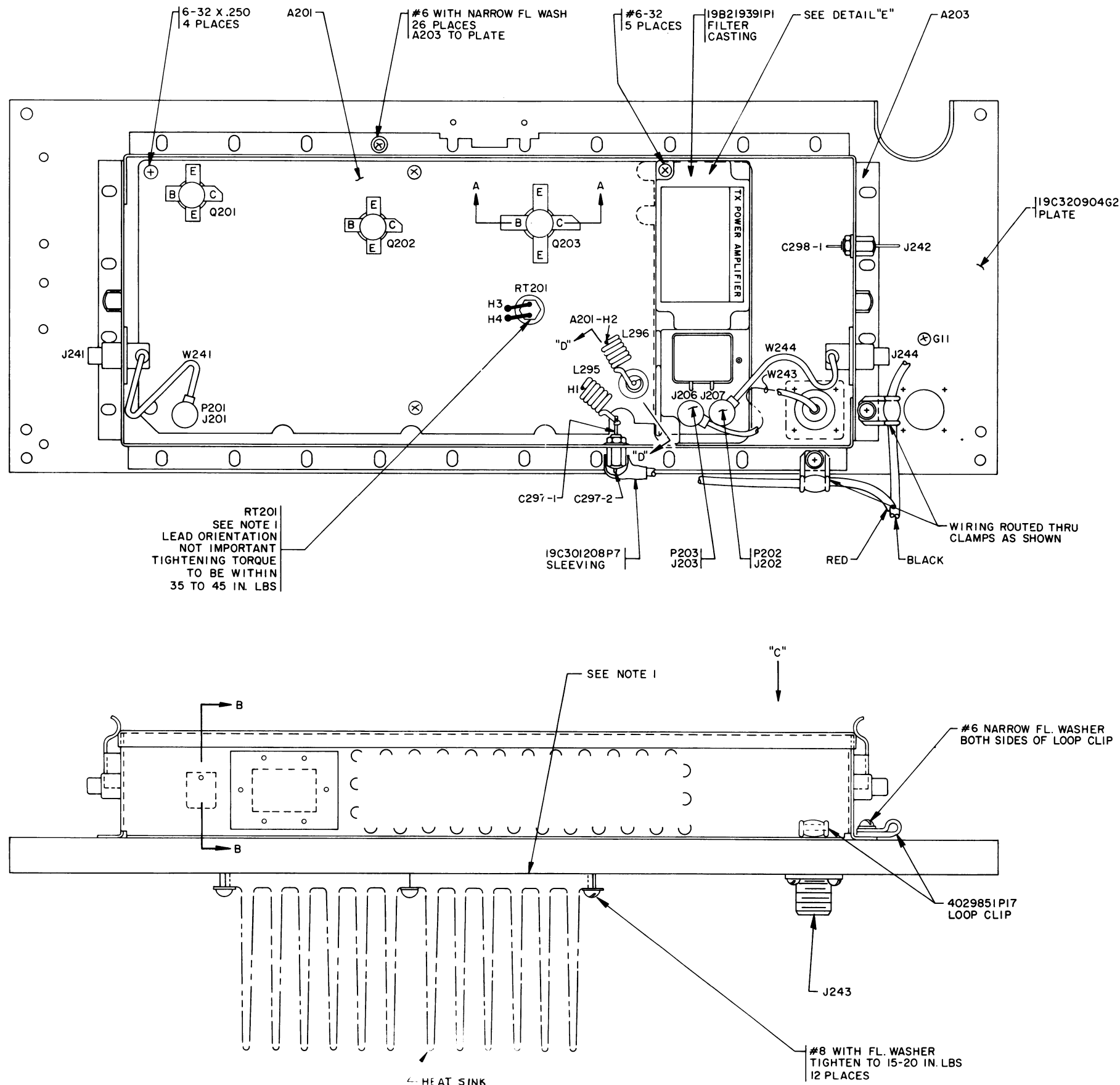
SECTION B-B



VIEW AT "C"

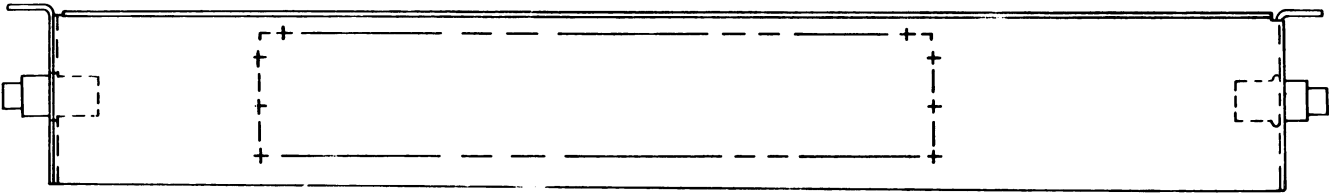


SECTION D-D

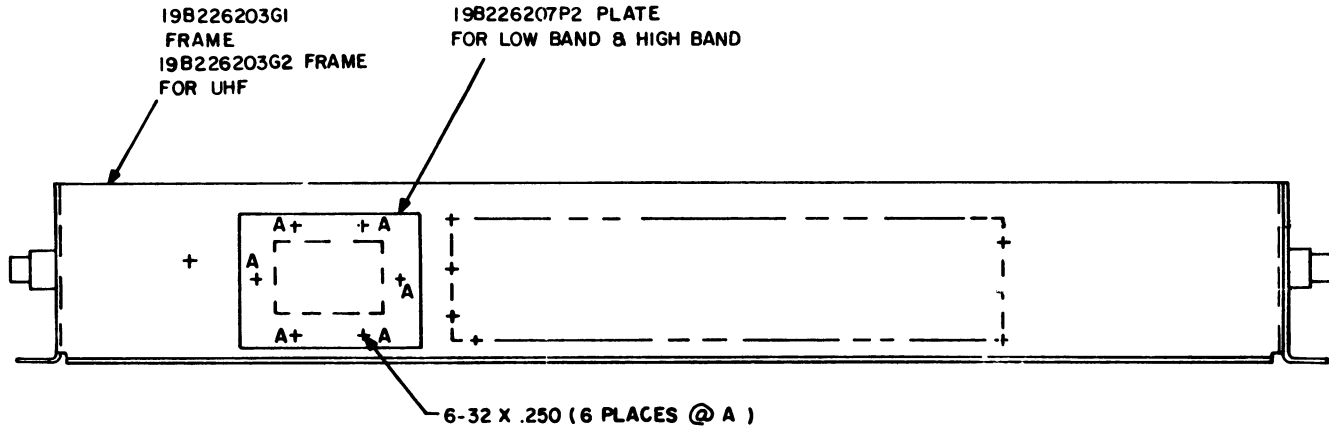
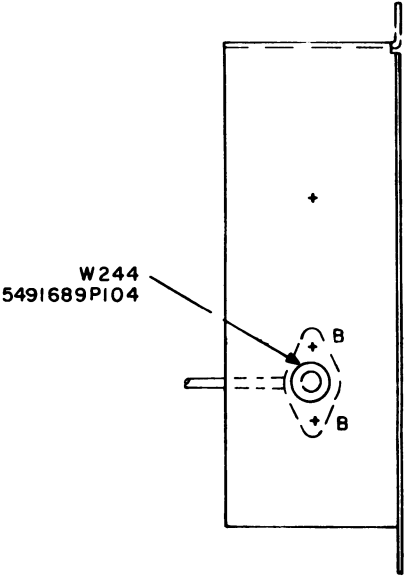
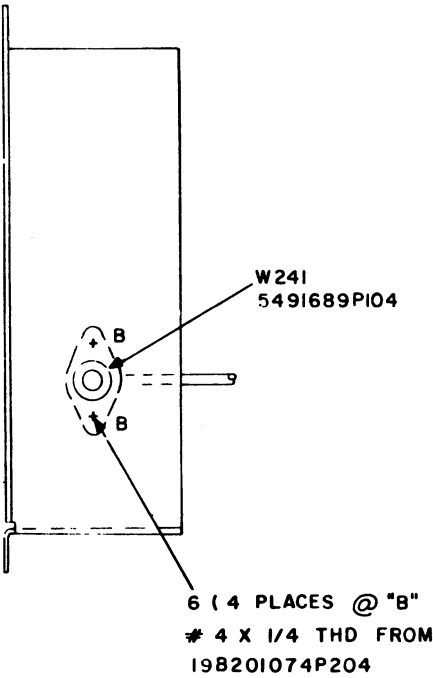
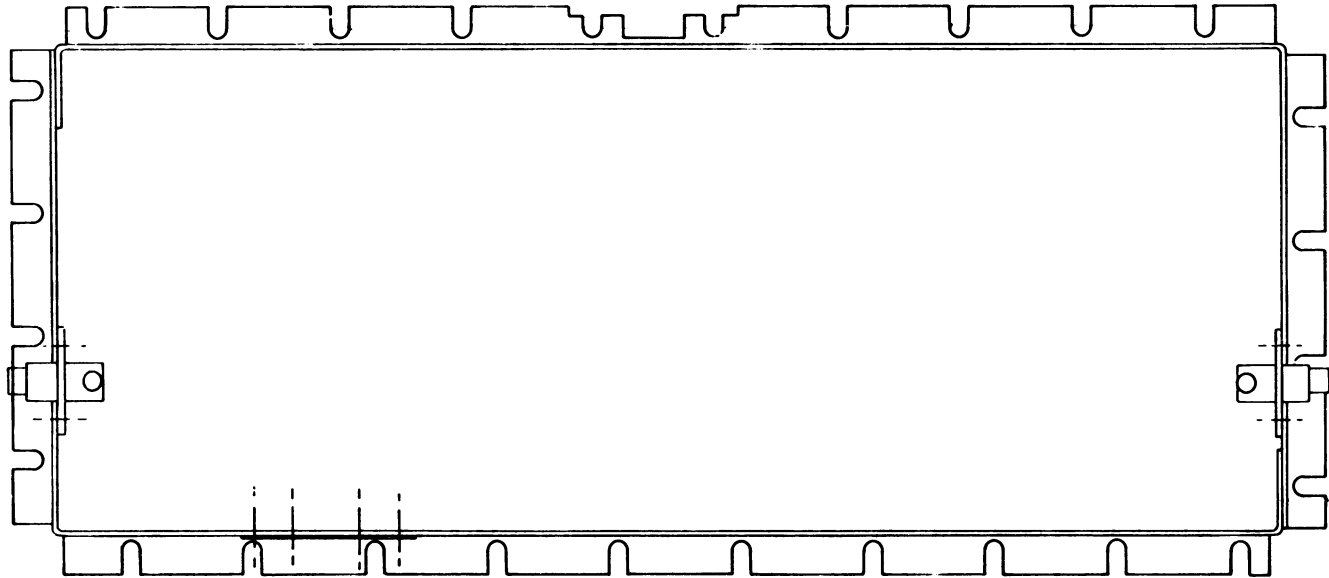


NOTES:
1. APPLY SILICONE GREASE TO BOTH SIDES OF MICA INSULATOR TO MOUNTING SURFACE OF Q201 THRU Q203 & RT201 AND UNPAINTED FLAT SURFACE OF HEAT SINKS.
NO GREASE ALLOWED ON THE THREADED PORTION OF THE MTG STUD.

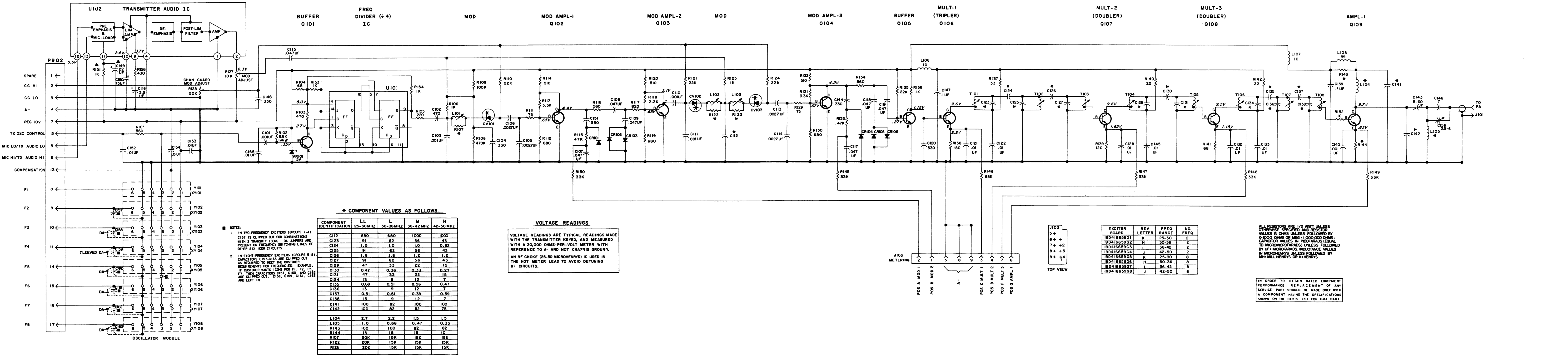
OUTLINE DIAGRAM
25—50 MHz, 50-WATT STATION
TRANSMITTER CONTINUOUS DUTY



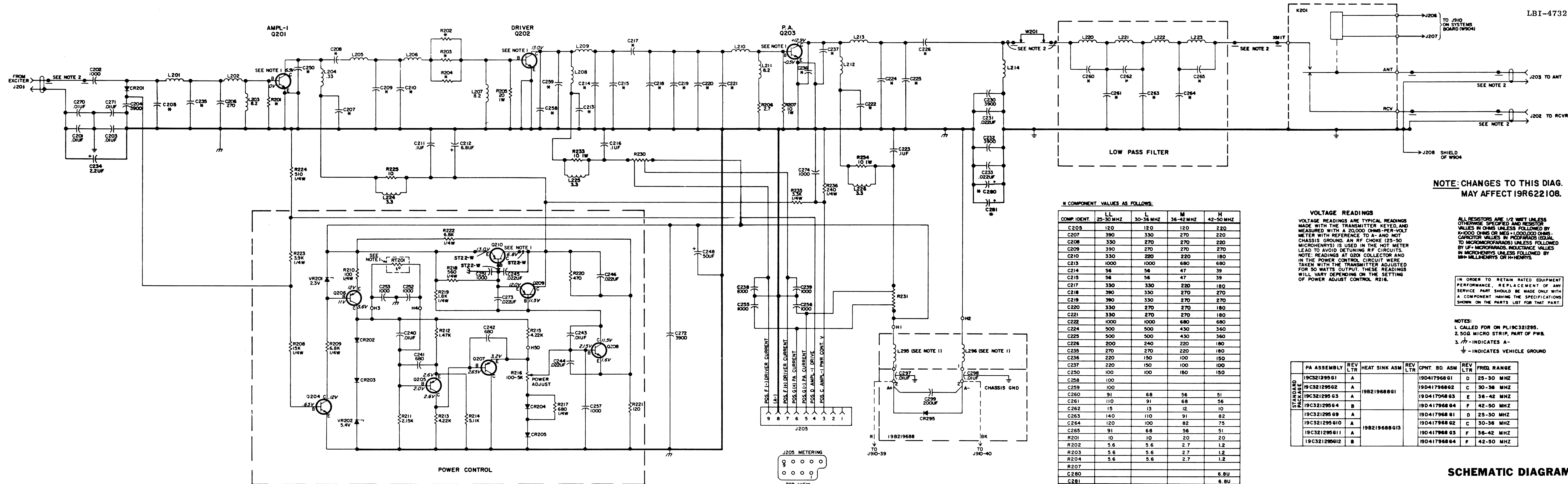
A203



(19D423098, Rev. 1)



SCHEMATIC DIAGRAM
25—50 MHz, EXCITER BOARD 19D416659G1-G8



PARTS LIST

SYMBOL	GE PART NO.	DESCRIPTION
		<div> <div> <div>LBI4897B</div> <div>25-50 MHz, 50 WATT POWER AMPLIFIER 19C321295G1 - G4 19C321295G0 - G12</div> </div> <div> <div>19C321295G1, G9 25-30 MHz (LL) 19C321295G2, G10 30-36 MHz (L) 19C321295G3, G11 36-42 MHz (M) 19C321295G4, G12 42-50 MHz (H) - REV B</div> <div>----- CAPACITORS -----</div> <div>C280 and C281</div> <div>19A134202P15</div> <div>Tantalum: 6.8 uF ±20%, 35 VDCW.</div> <div>----- INDUCTORS -----</div> <div>L280 and L286</div> <div>19A129356P1</div> <div>Coil.</div> <div>----- TRANSISTORS -----</div> <div>Q201</div> <div>19A116965P1</div> <div>Silicon, NPN.</div> <div>Q202</div> <div>19A116839P2</div> <div>Silicon, NPN.</div> <div>Q203</div> <div>19A116839P3</div> <div>Silicon, NPN.</div> <div>Q210</div> <div>19A116375P1</div> <div>Silicon, PNP.</div> <div>----- THERMISTORS -----</div> <div>RT201</div> <div>19A129379G1</div> <div>Thermistor: 40K ohms ±20%, color code white; sim to Carborundum Type NO806J-5.</div> <div> <div>POWER AMPLIFIER BOARD</div> <div>19D417968G1 25-30 MHz 19D417968G2 30-36 MHz 19D417968G3 36-42 MHz 19D417968G4 42-50 MHz - REV F</div> <div>----- CAPACITORS -----</div> <div>C201</div> <div>19A116080P101</div> <div>Polyester: 0.01 uF ±10%, 50 VDCW.</div> <div>C202</div> <div>19A116655P19</div> <div>Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.</div> <div>C203</div> <div>19A116080P101</div> <div>Polyester: 0.01 uF ±10%, 50 VDCW.</div> <div>C204</div> <div>19A116655P23</div> <div>Ceramic disc: 3900 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.</div> <div>C205LL</div> <div>19A700105P36</div> <div>Mica: 120 pF ±5%, 500 VDCW.</div> <div>C205L</div> <div>19A700105P36</div> <div>Mica: 120 pF ±5%, 500 VDCW.</div> <div>C205M</div> <div>19A700105P36</div> <div>Mica: 120 pF ±5%, 500 VDCW.</div> <div>C205H</div> <div>19A700105P44</div> <div>Mica: 220 pF ±5%, 500 VDCW.</div> <div>C206</div> <div>19A700105P46</div> <div>Mica: 270 pF ±5%, 500 VDCW.</div> <div>C207LL</div> <div>7489162P41</div> <div>Silver mica: 390 pF ±5%, 500 VDCW; sim to Sprague Type 118.</div> <div>C207L</div> <div>7489162P39</div> <div>Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.</div> <div>C207M</div> <div>19A700105P46</div> <div>Mica: 270 pF ±5%, 500 VDCW.</div> <div>C207H</div> <div>19A700105P44</div> <div>Mica: 220 pF ±5%, 500 VDCW.</div> <div>C208LL</div> <div>7489162P39</div> <div>Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.</div> <div>C208L and C208M</div> <div>19A700105P46</div> <div>Mica: 270 pF ±5%, 500 VDCW.</div> <div>C208H</div> <div>19A700105P44</div> <div>Mica: 220 pF ±5%, 500 VDCW.</div> <div>C209LL</div> <div>7489162P41</div> <div>Silver mica: 390 pF ±5%, 500 VDCW; sim to Sprague Type 118.</div> <div>C209L</div> <div>19A700105P46</div> <div>Mica: 270 pF ±5%, 500 VDCW.</div> <div>C209M</div> <div>19A700105P46</div> <div>Mica: 270 pF ±5%, 500 VDCW.</div> <div>C209H</div> <div>19A700105P46</div> <div>Mica: 270 pF ±5%, 500 VDCW.</div> </div> </div></div>

*COMPONENTS ADDED, DELETED OR CHANGED BY PRODUCTION CHANGES

SYMBOL	GE PART NO.	DESCRIPTION
C210LL	7489162P39	Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C210L and C210M	19A700105P44	Mica: 220 pF ±5%, 500 VDCW.
C210H	19A700105P41	Mica: 180 pF ±5%, 500 VDCW.
C211	19A116080P107	Polyester: 0.1 uF ±10%, 50 VDCW.
C212	5496267P18	Tantalum: 6.8 uF ±20%, 35 VDCW; sim to Sprague Type 150D.
C213LL and C213L	19A116655P19	Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C213M and C213H	19A116655P17	Ceramic disc: 680 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C214LL and C214L	19A116656P56J0	Ceramic disc: 56 pF ±5%, 500 VDCW; temp coef 0 PPM.
C214M	19A116656P47J0	Ceramic disc: 47 pF ±5%, 500 VDCW; temp coef 0 PPM.
C214H	19A116656P39J0	Ceramic disc: 39 pF ±5%, 500 VDCW, temp coef 0 PPM.
C215LL and C215L	19A116656P56J0	Ceramic disc: 56 pF ±5%, 500 VDCW; temp coef 0 PPM.
C215M	19A116656P47J0	Ceramic disc: 47 pF ±5%, 500 VDCW; temp coef 0 PPM.
C215H	19A116656P39J0	Ceramic disc: 39 pF ±5%, 500 VDCW, temp coef 0 PPM.
C216	19A116080P107	Polyester: 0.1 uF ±10%, 50 VDCW.
C217LL and C217L	7489162P39	Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C217M	19A700105P44	Mica: 220 pF ±5%, 500 VDCW.
C217H	19A700105P41	Mica: 180 pF ±5%, 500 VDCW.
C218LL	7489162P41	Silver mica: 390 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C218L	7489162P39	Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C218M and C218H	19A700105P46	Mica: 270 pF ±5%, 500 VDCW.
C219LL	7489162P41	Silver mica: 390 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C219L	7489162P39	Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C219M and C219H	19A700105P46	Mica: 270 pF ±5%, 500 VDCW.
C220LL	7489162P39	Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C220L and C220M	19A700105P46	Mica: 270 pF ±5%, 500 VDCW.
C220H	19A700105P41	Mica: 180 pF ±5%, 500 VDCW.
C221LL	7489162P39	Silver mica: 330 pF ±5%, 500 VDCW; sim to Sprague Type 118.
C221L and C221M	19A700105P46	Mica: 270 pF ±5%, 500 VDCW.
C221H	19A700105P41	Mica: 180 pF ±5%, 500 VDCW.
C222LL and C222L	19A116655P19	Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C222M and C222H	19A116655P17	Ceramic disc: 680 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C223	19A116080P107	Polyester: 0.1 uF ±10%, 50 VDCW.
C224LL and C224L	19A116679P500J	Silver Mica: 500 pF ±5%, 250 VDCW.
C224M	19A700015P44	Metallized teflon: 430 pF ±5%, 250 VDCW.
C224H	19A700015P42	Metallized teflon: 360 pF ±5%, 250 VDCW.

SYMBOL	GE PART NO.	DESCRIPTION
C225LL and C225L	19A116679P500J	Silver Mica: 500 pF ±5%, 250 VDCW.
C225M	19A700015P44	Metallized teflon: 430 pF ±5%, 250 VDCW.
C225H	19A700015P42	Metallized teflon: 360 pF ±5%, 250 VDCW.
C226LL	19A700015P36	Teflon/Mica: 200 pF ±5%, 250 VDCW.
C226L	19A700015P38	Teflon/Mica: 240 pF ±5%, 250 VDCW.
C226M	19A700015P37	Teflon/Mica: 220 pF ±5%, 250 VDCW.
C226H	19A700015P35	Teflon/Mica: 180 pF ±5%, 250 VDCW.
C230	19A116655P23	Ceramic disc: 3900 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C231	19A116080P103	Polyester: 0.022 uF ±10%, 50 VDCW.
C232	19A116655P23	Ceramic disc: 3900 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C233	19A116080P103	Polyester: 0.022 uF ±10%, 50 VDCW.
C234	5496267P13	Tantalum: 2.2 uF ±20%, 20 VDCW; sim to Sprague Type 150D.
C235LL and C235L	19A700105P46	Mica: 270 pF ±5%, 500 VDCW.
C235M	19A700105P44	Mica: 220 pF ±5%, 500 VDCW.
C235H	19A700105P41	Mica: 180 pF ±5%, 500 VDCW.
C236LL	19A116656P220J4	Ceramic disc: 220 pF ±5%, 500 VDCW; temp coef -470 PPM.
C236L	19A116656P150J1	Ceramic disc: 150 pF ±5%, 500 VDCW, temp coef -150 PPM.
C236M	19A116656P100J1	Ceramic disc: 100 pF ±5%, 500 VDCW, temp coef -150 PPM.
C236H	19A116656P150J1	Ceramic disc: 150 pF ±5%, 500 VDCW, temp coef -150 PPM.
C237LL	19A116656P220J4	Ceramic disc: 220 pF ±5%, 500 VDCW; temp coef -470 PPM.
C237L	19A116656P150J1	Ceramic disc: 150 pF ±5%, 500 VDCW, temp coef -150 PPM.
C237M and C237H	19A116656P100J1	Ceramic disc: 100 pF ±5%, 500 VDCW, temp coef -150 PPM.
C238 and C239	19A116655P19	Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C240	19A116080P101	Polyester: 0.01 uF ±10%, 50 VDCW.
C241 and C242	19A116655P17	Ceramic disc: 680 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C243	19A116080P101	Polyester: 0.01 uF ±10%, 50 VDCW.
C244 thru C246	19A116080P103	Polyester: 0.022 uF ±10%, 50 VDCW.
C248	19A115680P4	Electrolytic: 50 uF +150% -10%, 25 VDCW; sim to Mallory Type TTX.
C250	19A116656P100J1	Ceramic disc: 100 pF ±5%, 500 VDCW, temp coef -150 PPM.
C250M and C250H	19A116656P150J1	Ceramic disc: 150 pF ±5%, 500 VDCW, temp coef -150 PPM.
C251 thru C253	19A116655P19	Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C255 thru C257	19A116655P19	Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C258LL	19A116656P100K4	Ceramic disc: 100 pF ±10%, 500 VDCW, temp coef -470 PPM.
C259LL	19A116656P100K4	Ceramic disc: 100 pF ±10%, 500 VDCW, temp coef -470 PPM.
C260LL*	19A116656P91J2	Ceramic disc: 91 pF ±5%, 500 VDCW; temp coef -220 PPM.

SYMBOL	GE PART NO.	DESCRIPTION
C260L	19A116656P68J1	Ceramic disc: 68 pF ±5%, 500 VDCW, temp coef -150 PPM.
C260M	19A116656P56J1	Ceramic disc: 56 pF ±5%, 500 VDCW; temp coef -150 PPM.
C260H	19A116656P51J1	Ceramic disc: 51 pF ±5%, 500 VDCW; temp coef -150 PPM.
C261LL	19A700015P30	Silver mica: 110 pF ±5%, 250 VDCW.
C261L	19A700015P28	Teflon/Mica: 91 pF ±5%, 250 VDCW.
C261M	19A700015P25	Silver mica: 68 pF ±5%, 250 VDCW.
C261H	19A700015P23	Teflon/Mica: 56 pF ±5%, 250 VDCW.
C262LL	19A116656P15J1	Ceramic disc: 15 pF ±5%, 500 VDCW, temp coef -150 PPM.
C262L	19A116656P13J1	Ceramic disc: 13 pF ±5%, 500 VDCW, temp coef -150 PPM.
C262M	19A116656P12J1	Ceramic disc: 12 pF ±5%, 500 VDCW; temp coef -150 PPM.
C262H	19A116656P10J1	Ceramic disc: 10 pF ±5%, 500 VDCW; temp coef -150 PPM.
C263LL	19A116679P140J	Silver Mica: 140 pF ±5%, 250 VDCW.
C263L	19A700015P30	Silver mica: 110 pF ±5%, 250 VDCW.
C263M	19A700015P28	Teflon/Mica: 91 pF ±5%, 250 VDCW.
C263H	19A700015P27	Silver mica: 82 pF ±5%, 250 VDCW.
C264LL	19A700015P31	Teflon/Mica: 120 pF ±5%, 250 VDCW.
C264L	19A700015P29	Teflon/Mica: 100 pF ±5%, 250 VDCW.
C264M	19A700015P27	Silver mica: 82 pF ±5%, 250 VDCW.
C264H	19A700015P26	Teflon/Mica: 75 pF ±5%, 250 VDCW.
C265LL*	19A116656P91J2	Ceramic disc: 91 pF ±5%, 500 VDCW; temp coef -220 PPM.
		In 19D417968G1 of REV A and earlier:
	19A116656P82J1	Ceramic disc: 82 pF ±5%, 500 VDCW, temp coef -150 PPM.
C265L	19A116656P68J1	Ceramic disc: 68 pF ±5%, 500 VDCW, temp coef -150 PPM.
C265M	19A116656P56J1	Ceramic disc: 56 pF ±5%, 500 VDCW; temp coef -150 PPM.
C265H	19A116656P51J1	Ceramic disc: 51 pF ±5%, 500 VDCW; temp coef -150 PPM.
C270 and C271	19A116080P101	Polyester: 0.01 uF ±10%, 50 VDCW.
C272	19A116655P23	Ceramic disc: 3900 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
C273	19A116080P103	Polyester: 0.022 uF ±10%, 50 VDCW.
C274	19A116655P19	Ceramic disc: 1000 pF ±20%, 1000 VDCW; sim to RMC Type JF Discap.
		----- DIODES AND RECTIFIERS -----
CR201*	19A116052P2	Silicon, fast recovery; sim to Hewlett Packard 5082-2811.
	19A115250P1	Earlier than REV A:
	19A115250P1	Silicon, fast recovery, 225 mA, 50 PIV.
	19A115250P1	Silicon, fast recovery, 225 mA, 50 PIV.
J201 thru J203	19A7000049P2	Connector, receptacle; 500 VDCW maximum; sim to NTFP-1058.
J205	198219374G1	Connector, Includes:
	19C317957P1	Shell.
	19A129360P9	Contact, electrical.
	19A134263P2	Contact, electrical: sim to Selectro 229-1071.
J208	4033513P4	Contact, electrical: sim to Bead Chain L93-3.

SYMBOL	GE PART NO.	DESCRIPTION
K201	19A7000061P1	<div> <div>----- RELAYS -----</div> <div>Hermetic sealed: 180 to 341 ohms coil res, 8-16.3 VDC; sim to GE 38A1V16042, CP Clare HFW-1201558, or Potter-Brumfield HCM6160.</div> </div>
L201LL	19A129347P2	Coil.
L201L	19A129347P1	Coil.
L201M	19A129347P3	Coil.
L201H	19A129347P5	Coil.
L202LL	19A129352P9	Coil.
L202L	19A129354P4	Coil.
L202M	19A129352P8	Coil.
L202H	19A129352P7	Coil.
L203	19A700000P122	Coil, RF: 8.2 uH ±10%; sim to Jeffers 4422-3K.
L204	19A700000P6	Coil, RF: 330 nH ±20%; sim to Jeffers 411-3.
L205LL and L205L	19A129351P3	Coil.
L205M	19A129351P2	Coil.
L205H	19A129351P2	Coil.
L206LL	19A129352P1	Coil.
L206L	19A129352P1	Coil.
L206M	19A129352P3	Coil.
L206H	19A129348P2	Coil.
L207	19A700000P122	Coil, RF: 8.2 uH ±10%; sim to Jeffers 4422-3K.
L208LL and L208L	19A129348P1	Coil.
L208H	19A129348P2	Coil.
L209M and L209H	19A129355P1	Coil.
L209L	19A129352P4	Coil.
L209LL and L209M	19A129355P1	Coil.
L209H	19A129352P2	Coil.
L210LL	19A129358P1	Coil.
L210L	19A129359P1	Coil.
L210M	19A129357P1	Coil.
L210H and R204L	19A129357P2	Coil.
L211	19A700000P122	Coil, RF: 8.2 uH ±10%; sim to Jeffers 4422-3K.
L212LL and L212L	19A129349P1	Coil.
L212M	19A129349P2	Coil.
L212H and L212M	19A129351P1	Coil.
L213L	19A129358P2	Coil.
L213M	19A129355P3	Coil.
L213H	19A129351P4	Coil.
L214LL	19A129353P3	Coil.
L214L	19A129355P5	Coil.
L214M	19A129355P4	Coil.
L214H	19A129352P10	Coil.
L220LL	19A129360P9	Coil.
L220L	19A129360P6	Coil.
L220M	19A129360P4	Coil.
L220H	19A129360P1	Coil.
L221LL	19A129360P10	Coil.

SYMBOL	GE PART NO.	DESCRIPTION
L221L	19A129360P7	Coil.
L221M	19A129360P3	Coil.
L221H	19A129360P2	Coil.
L222LL	19A129360P11	Coil.
L222L	19A129360P8	Coil.
L222M	19A129360P5	Coil.
L222H	19A129360P3	Coil.
L223LL	19A129360P9	Coil.
L223L	19A129360P6	Coil.
L223M	19A129360P4	Coil.
L223H	19A129360P1	Coil.
L224	19A700000P17	Coil, RF: 3.3 uH $\pm 10\%$; sim to Jeffers 4421-1K.
L225 and L226	19A129346G1	Coil.
		----- TRANSISTORS -----
Q204	19A115910P1	Silicon, NPN; sim to Type 2N3904.
Q205 thru Q207	19A115768P1	Silicon, PNP; sim to Type 2N3702.
Q208	19A115910P1	Silicon, NPN; sim to Type 2N3904.
Q209	19A129187P1	Silicon, PNP.
		----- RESISTORS -----
R201LL and R201L	19A700113P15	Composition: 10 ohms $\pm 5\%$, 1/2 w.
R201M and R201H	3R77P200J	Composition: 20 ohms $\pm 5\%$, 1/2 w.
R202LL and R202L	19A700112P9	Composition: 5.6 ohms $\pm 5\%$, 1 w.
R202M	19A700112P1	Composition: 2.7 ohms $\pm 5\%$, 1 w.
R202H	7147161P22	Composition: 1.2 ohms $\pm 5\%$, 1/2 w.
R203LL and R203L	19A700112P9	Composition: 5.6 ohms $\pm 5\%$, 1 w.
R203M	19A700112P1	Composition: 2.7 ohms $\pm 5\%$, 1 w.
R203H	7147161P22	Composition: 1.2 ohms $\pm 5\%$, 1/2 w.
R204LL and R204L	19A700112P9	Composition: 5.6 ohms $\pm 5\%$, 1 w.
R204M	19A700112P1	Composition: 2.7 ohms $\pm 5\%$, 1 w.
R204H	7147161P22	Composition: 1.2 ohms $\pm 5\%$, 1/2 w.
R205	3R78P200J	Composition: 20 ohms $\pm 5\%$, 500 VDCW, 1 w.
R206	19A700113P1	Composition: 2.7 ohms $\pm 5\%$, 1/2 w.
R207	19A700112P15	Composition: 10 ohms $\pm 5\%$, 1 w.
R208	19A700106P91	Composition: 15K ohms $\pm 5\%$, 1/4 w.
R209	19A700106P83	Composition: 6.8K ohms $\pm 5\%$, 1/4 w.
R210	19A700106P39	Composition: 100 ohms $\pm 5\%$, 1/4 w.
R211	19A116278P233	Metal film: 2150 ohms $\pm 5\%$, 1/2 w.
R212	19A116278P21	Metal film: 1470 ohms $\pm 5\%$, 1/2 w.
R213	19A116278P261	Metal film: 4220 ohms $\pm 5\%$, 1/2 w.
R214	19A116278P269	Metal film: 5110 ohms $\pm 5\%$, 1/2 w.
R215	19A116278P261	Metal film: 4220 ohms $\pm 5\%$, 1/2 w.
R216	19A116559P102	Variable cermet: 5000 ohms $\pm 20\%$, 1/2 w; sim to CTS Series 360.
R217	19A700106P59	Composition: 680 ohms $\pm 5\%$, 1/4 w.
R218	19A700106P57	Composition: 560 ohms $\pm 5\%$, 1/4 w.
R219	19A700106P69	Composition: 1.8K ohms $\pm 5\%$, 1/4 w.
R220	19A700113P55	Composition: 470 ohms $\pm 5\%$, 1/2 w.